## CONTENTS

## Selection Guide

$\qquad$Selection Guide2, 3
■Filters
Resistor Tunable Filter SV Series/ SR Series/ SRA Series ..... 4 to 10
BCD Resistor
RD-404 ..... 11
Resistor Tunable FilterHR Series12 to 15
Resistor Tunable FilterRT Series16 to 17Voltage Tunable FilterProgrammable FilterProgrammable FilterProgrammable Filter
18 to 19
VT Series20 to 22
DT-408 Series ..... 23 to 26
DT-208 Series ..... 27
Programmable Filter
DT-5FL/6FL ..... 28, 29
Programmable FilterDT-8FL30, 31
Fixed Frequency FilterDV Series32 to 35
Fixed Frequency FilterCF Series36 to 38
Band-elimination FilterSD-1BE39
200B/S Band Pass Filter40
Low Pass Filter for Wide Band Speech Signals SF-8FLC-1 ..... 41
Amplifiers
Low Noise FET AmplifierCA-251F442, 43
Low Noise FET Differential Amplifier CA-451F4 ..... 44, 45
Low Noise AmplifierLow Noise Differential AmplifierCA-261F246, 47
Differential AmplifierProgrammable Gain AmplifierBinary Latch Adapter
CA-461F2 ..... 48, 49
CA-406L2 ..... 50, 51
CA-206L2 ..... 52, 53
CA-903N ..... 54
High Speed Inverting Amplifier CA-102R3 ..... 55
Low Noise Amplifier
SA Series ..... 56, 57
Oscillators
Resistor Tunable Oscillator CG-402 ..... 58, 59
Resistor Tunable Oscillator CG-202 ..... 60, 61
CG-102/302 ..... 62 to 64
Oscillator Adapter
Random Binary Generator ..... 65 to 68
CG-742N ..... 69 to 71
CD-552R3/552R4 ..... 72 to 76
Phase Detector
CD-951V4 ..... 77 to 79
CD-505R2 ..... 80 to 84
Dimensions
Dimensional Outline Drawing ..... 85 to 87

## SELECTION GUIDE

## Filters

LP: Low pass BW: Band width

HP: High pass BP: Band pass
SIP: Single-inline package

BE: Band elimination DIP: Dual-inline package

| Model | Type | Order | Rolloff | Attenuation characteristic | Cut-off (center) frequency setting |  | Dimensions | Refer to |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | Range* ${ }^{\text {1 }}$ | Control type |  |  |
| SR-4FL | LP | 4 | 42dB/oct equivalent | Elliptic | 40 Hz to 100 kHz | Resistor tunable | 20-pin SIP | p6, p8 to 10 |
| SRA-4FL1 | LP | 4 | $42 \mathrm{~dB} /$ oct equivalent | Elliptic | 40 Hz to 1.6 kHz | Resistor tunable | 20-pin SIP | p7 to 10 |
| SV-4FL* | LP | 4 | $42 \mathrm{~dB} /$ oct equivalent | Elliptic | 10 Hz to 100 kHz | Resistor tunable | 15-pin SIP | p4 to 5 |
| SR-4FH | HP | 4 | $42 \mathrm{~dB} /$ oct equivalent | Elliptic | 40 Hz to 5 kHz | Resistor tunable | 20-pin SIP | p6, p8 to 10 |
| SRA-4FH1 | HP | 4 | $42 \mathrm{~dB} /$ oct equivalent | Elliptic | 40 Hz to 1.6 kHz | Resistor tunable | 20-pin SIP | p7 to 10 |
| SR-4BL | LP | 4 | $24 \mathrm{~dB} /$ oct | Butterworth | 40 Hz to 100 kHz | Resistor tunable | 20-pin SIP | p6, p8 to 10 |
| SRA-4BL1 | LP | 4 | $24 \mathrm{~dB} / \mathrm{cot}$ | Butterworth | 40 Hz to 1.6 kHz | Resistor tunable | 20-pin SIP | p7 to 10 |
| SV-4BL* | LP | 4 | $24 \mathrm{~dB} /$ oct | Butterworth | 10 Hz to 100 kHz | Resistor tunable | 15-pin SIP | p4 to 5 |
| SR-4BH | HP | 4 | $24 \mathrm{~dB} /$ oct | Butterworth | 40 Hz to 5 kHz | Resistor tunable | 20-pin SIP | p6, p8 to 10 |
| SRA-4BH1 | HP | 4 | $24 \mathrm{~dB} / \mathrm{oct}$ | Butterworth | 40 Hz to 1.6 kHz | Resistor tunable | 20-pin SIP | p7 to p10 |
| SR-1BP | BP | 2(1-pole pair) | 6dB/oct BW | Butterworth | 40 Hz to 10kHz | Resistor tunable | 20-pin SIP | p6, p8 to 10 |
| SR-2BP | BP | 4(2-pole pair) | $12 \mathrm{~dB} /$ oct BW | Butterworth | 40 Hz to 10 kHz | Resistor tunable | 20-pin SIP | p6, p8 to 10 |
| SRA-2BP1 | BP | 4(2-pole pair) | 12dB/oct BW | Butterworth | 40 Hz to 1.6 kHz | Resistor tunable | 20-pin SIP | p7 to 10 |
| SR-2BE | BE | 4(2-pole pair) | Max. attenuation: 60 dB | Butterworth | 40 Hz to 10 kHz | Resistor tunable | 20-pin SIP | p6, p8 to 10 |
| SR-2BLH | LP, HP | 2 | 12dB/oct | Butterworth | 40 Hz to 100 kHz | Resistor tunable | 20-pin SIP | p6, p8 to 10 |

* Supply voltage: $5 \mathrm{~V}, 3.3 \mathrm{~V}$ single power for SV Series

| RD-404 | Logic frequency setting is available with the combination use of SR and SRA filters. (Filter characteristics of SR and SRA filters applied) |  |  |  | 10 Hz to 16.9 kHz | Digital tunable | 20-pin SIP | p11 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| HR-4FL | LP | 4 | 42dB/oct equivalent | Elliptic | 10 Hz to 100 kHz | Resistor tunable | 24-pin DIP | p12 to 15 |
| HR-4FH | HP | 4 | $42 \mathrm{~dB} /$ oct equivalent | Elliptic | 10 Hz to 50 kHz | Resistor tunable | 24-pin DIP | p12 to 15 |
| HR-4BL | LP | 4 | $24 \mathrm{~dB} /$ oct | Butterworth | 10 Hz to 100 kHz | Resistor tunable | 24-pin DIP | p12 to 15 |
| HR-4BH | HP | 4 | $24 \mathrm{~dB} /$ oct | Butterworth | 10 Hz to 50 kHz | Resistor tunable | 24-pin DIP | p12 to 15 |
| HR-2BP | BP | 4(2-pole pair) | 12dB/oct BW | Butterworth | 10 Hz to 50 kHz | Resistor tunable | 24-pin DIP | p12 to 15 |
| RT-8FLA | LP | 8 | $135 \mathrm{~dB} /$ oct equivalent | Elliptic | 10 Hz to 20 kHz | Resistor tunable | 40-pin DIP | p16, p17 |
| RT-8FLB | LP | 8 | $100 \mathrm{~dB} /$ oct equivalent | Elliptic | 10 Hz to 20kHz | Resistor tunable | 40-pin DIP | p16, p17 |
| RT-3BP | BP | 6 (3-pole pair) | 1/3oct BW | Butterworth | 10 Hz to 20kHz | Resistor tunable | 40-pin DIP | p16, p17 |
| VT-4BLA | LP | 4 | $24 \mathrm{~dB} / \mathrm{oct}$ | Butterworth | 100 Hz to 100 kHz | Voltage tunable | 40-pin DIP | p18, p19 |
| VT-4BHA | HP | 4 | $24 \mathrm{~dB} /$ oct | Butterworth | 20 Hz to 20kHz | Voltage tunable | 40-pin DIP | p18, p19 |
| VT-2BP | BP | 4(2-pole pair) | 12dB/oct BW | Butterworth | 20 Hz to 20kHz | Voltage tunable | 40-pin DIP | p18, p19 |
| DT-212D | $\begin{aligned} & \text { LP, HP, } \\ & \text { BP } \end{aligned}$ | 2(1-pole pair) | $\begin{aligned} & \hline 12 \mathrm{~dB} / \text { oct (HP/LP) } \\ & 6 \mathrm{~dB} / \text { oct (BP) } \end{aligned}$ | Universal | 1 Hz to 159.9 kHz | Digital tunable | 40-pin DIP | p20 to 22 |
| DT-408D | $\begin{aligned} & \text { LP, HP, } \\ & \text { BP } \end{aligned}$ | 2(1-pole pair) | $\begin{array}{\|l\|} \hline 12 \mathrm{~dB} / \text { oct } \times 2(\mathrm{HP} / \mathrm{LP}) \\ 6 \mathrm{~dB} / \text { oct } \times 2(\mathrm{BP}) \\ \hline \end{array}$ | Universal | 1 kHz to 159 kHz | Digital tunable | 40-pin DIP | p23 to 26 |
| DT-208D | $\begin{aligned} & \text { LP, HP, } \\ & \text { BP } \end{aligned}$ | 2(1-pole pair) | $\begin{aligned} & \text { 12dB/oct (HP/LP) } \\ & 6 \mathrm{~dB} / \text { oct (BP) } \\ & \hline \end{aligned}$ | Universal | 10 kHz to 1.59 MHz | Digital tunable | 40-pin DIP | p27 |
| DT-5FL | LP | 5 | 60dB/oct equivalent | Elliptic | 10 Hz to 20kHz | Digital tunable | 40-pin DIP | p28, p29 |
| DT-6FL | LP | 6 | 80dB/oct equivalent | Elliptic | 10 Hz to 20kHz | Digital tunable | 40-pin DIP | p28, p29 |
| DT-8FL | LP | 8 | $130 \mathrm{~dB} /$ oct equivalent | Elliptic | 20 Hz to 100 kHz | Digital tunable | 60-pin DIP | p30, p31 |
| DV Series*2 | LP, HP, <br> BP, BE | 2(1-pole pair) to 8 <br> (4-pole pair) | $18 \mathrm{~dB} /$ oct to $200 \mathrm{~dB} /$ oct(LP) $18 \mathrm{~dB} /$ oct to $75 \mathrm{~dB} /$ oct(HP) 12dB/oct BW to 36dB/oct BW(BP) Max. attenuation: 26 to 72 dB (BE) | Butterworth, Chebyshev, Elliptic, Universal | 0.01 Hz to 20 kHz | Frequency fixed |  | p32 to 35 |
| CF Series*2 | $\begin{aligned} & \text { LP, HP, } \\ & \text { BP, BE } \end{aligned}$ | $\begin{aligned} & \text { 2(1-pole pair) } \\ & \text { to } 8 \\ & \text { (4-pole pair) } \end{aligned}$ | $\begin{aligned} & \text { 18dB/oct to } 300 \mathrm{~dB} / \text { oct(LP/HP) } \\ & \text { 12dB/oct BW to } 36 \mathrm{~dB} / \text { /oct BW(BP) } \\ & \text { Max. attenuation: } 26 \text { to } 72 \mathrm{~dB}(\mathrm{BE}) \end{aligned}$ | Butterworth, Chebyshev, Elliptic, Universal | 1 Hz to 100 kHz | Frequency fixed | $\begin{aligned} & \text { 28-pin DIP, } \\ & \text { DIP 40-pin } \end{aligned}$ | p36 to 38 |
| SD-1BE | BE | 2(1-pole pair) | Max. attenuation: 24 dB | Butterworth | $50 \mathrm{~Hz} / 60 \mathrm{~Hz}$ | Digital tunable | 20-pin SIP | p39 |
| CF-4FPA | BP | 8(4-pole pair) | Min.: 15dB ( $\pm 200 \mathrm{~Hz}$ ) <br> Min.: 45dB (300Hz) | Elliptic | 800 Hz to 2800 Hz | Frequency fixed | 40-pin DIP | p40 |
| SF-8FLC-1 | LP | 8 | Max.: -25dB (8kHz), <br> Max.: -50dB (9kHz), <br> Max.: $-70 \mathrm{~dB}(14 \mathrm{kHz})$ |  | 7 kHz | Frequency fixed | 20-pin SIP | p41 |

[^0]
## Amplifiers

| Model | Input configuration | Gain | Frequency | Impedance | Voltage noise (typ.) | Current noise (typ.) | Dimensions | Refer to |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CA-251F4 | Single-end FET | $\times 100$ fixed | DC to 10 MHz | $1 \mathrm{M} \Omega$ | $1.4 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ | $150 f \mathrm{~A} / \sqrt{\mathrm{Hz}}$ | 20-pin shielded SIP | p42, p43 |
| CA-261F2 | Single-end bipolar | $\times 100$ fixed | DC to 200kHz | $100 \mathrm{k} \Omega$ | $0.8 \mathrm{nV} / \mathrm{/Hz}$ | $1.5 \mathrm{pA} / \sqrt{\mathrm{Hz}}$ | 20-pin shielded SIP | p46, p47 |
| CA-206L2 | Single-end FET | $\times 1$ to 100 (variable) | DC to 500 kHz | $1 \mathrm{M} \Omega$ | $7 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ | - | 20-pin SIP | p52, p53 |
| CA-451F4 | Differential FET | $\times 100$ fixed | DC to 10 MHz | 2G $\Omega$ | $2.5 \mathrm{nV} / \mathrm{/Hz}$ | $100 \mathrm{fA} / \sqrt{\mathrm{Hz}}$ | 20-pin shielded SIP | p44, p45 |
| CA-461F2 | Differential bipolar | $\times 100$ fixed | DC to 200kHz | $100 \mathrm{k} \Omega$ | $1.5 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ | $2.5 \mathrm{pA} / \sqrt{\mathrm{Hz}}$ | 20-pin shielded SIP | p48, p49 |
| CA-406L2 | Differential FET | $\times 1$ to 100 (variable) | DC to 200kHz | 30G $\Omega$ | $27 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ | - | 20-pin SIP | p50, p51 |
| CA-102R3 | Inverting amplifier | Connected with 2 external resistors | DC to 10MHz | - | - | - | 12-pin SIP | p55 |
| CA-903N | Adapter to enable CA-206L2/406L2 setting in binary code (endowed with latching functions) |  |  |  |  |  |  | p54 |

- Low Noise Amplifier

| SA-220F5 | Single-end FET | 46 dB | 300 Hz to 100 MHz | $1 \mathrm{M} \Omega$ | $0.5 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ | $200 \mathrm{fA} / \sqrt{\mathrm{Hz}}$ | $68 \times 43 \times 28 \mathrm{~mm}^{*}$ | p 56 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| SA-230F5 | Single-end | 46 dB | 400 Hz to 140 MHz | $50 \Omega$ | $0.25 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ | $5.0 \mathrm{pA} / \sqrt{\mathrm{Hz}}$ | $68 \times 43 \times 17.6 \mathrm{~mm}^{*}$ | p 56 |
| SA-430F5 | Differential | 46 dB | 400 Hz to 110 MHz | $50 \Omega$ | $0.35 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ | $7.0 \mathrm{pA} / \sqrt{\mathrm{Hz}}$ | $68 \times 43 \times 28 \mathrm{~mm}^{*}$ | p 56 |
| SA-200F3 | Single-end | 40 dB | DC to 800 kHz | $1 \mathrm{k} / 10 \mathrm{k} /$ | $0.5 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ | $2.2 \mathrm{pA} / \sqrt{\mathrm{Hz}}$ | $68 \times 43 \times 17.6 \mathrm{~mm}{ }^{*}$ | p 57 |
| $100 \mathrm{k} \Omega$ |  |  |  |  |  |  |  |  |

* Excluding protruding sections


## Oscillators

| Model | Output waveform | Frequency range | Output voltage | Frequency setting | Dimensions | Refer to |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CG-102R1 | Sinewave | 20 Hz to 20kHz | 2.5Vrms (variable) | 2 external resistors connected | 24-pin DIP | p62 to 64 |
| CG-102R2 | Sinewave | 1 kHz to 100 kHz | 2.5 Vrms (variable) | 2 external resistors connected | 24-pin DIP | p62 to 64 |
| CG-202R3 | Sinewave | 100 kHz to 1 MHz | 2.5Vrms (variable) | 2 external resistors connected | 24-pin DIP | p60, p61 |
| CG-302R1 | Sinewave | 20 Hz to 20kHz | 2.5Vrms (variable) | 2 external resistors connected | 20-pin SIP | p62 to 64 |
| CG-302R2 | Sinewave | 1 kHz to 100kHz | 2.5Vrms (variable) | 2 external resistors connected | 20-pin SIP | p62 to 64 |
| CG-402R1 | Sinewave | 20 Hz to 20kHz | 2.5 Vrms (variable) | 2 external resistors connected | 12-pin SIP | p58, p59 |
| CG-402R2 | Sinewave | 1 kHz to 100kHz | 2.5 Vrms (variable) | 2 external resistors connected | 12-pin SIP | p58, p59 |
| OP-102 + DT-212 | Sinewave | 1 Hz to 159.9 kHz | 2.5 Vrms (variable) | BCD: 3 digits | 20-pin SIP | p65 to 67 |
| CG-742N | Random binary | - | $\pm 5 \mathrm{~V}$ | 1 external resistor connected/ external clock | 40-pin DIP | p69 to 71 |
| CG-742N + LPF | White noise | - | - | 1 external resistor connected/ external clock | 40-pin DIP | p69 to 71 |

## Phase Detectors

| Model | Frequency range | Input amplifier | Detection system | LPF | Gain | Reference signal | Phase shifter | Dimensions | Others | Refer to |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CD-552R3 | $\begin{aligned} & 1 \mathrm{kHz} \text { to } 200 \mathrm{kHz}, \\ & \times 1 \end{aligned}$ | Single-end | Square-wave multiplication | 1-pole to 1kHz | $\times 1$ to 10 | C-MOS(0/5V) | 0/90 ${ }^{\circ}$ | 20-pin shielded SIP | $2 f$ detection available | p72 to 76 |
| CD-552R4 | $\begin{aligned} & 10 \mathrm{kHz} \text { to } 2 \mathrm{MkHz}, \\ & \times 1 \end{aligned}$ | Single-end | Square-wave multiplication | 1-pole to 10 kHz | $\times 1$ to 10 | C-MOS(0/5V) | 0/90 ${ }^{\circ}$ | 20-pin shielded SIP | 2f detection available | p72 to 76 |
| CD-505R2 | 10 Hz to 10 kHz | Differential, $\times 1$ (band pass embedded) | Square-wave multiplication | 1/2-pole to 1 kHz | $\times 1$ | C-MOS(0/5V) | $\begin{aligned} & 90^{\circ} \pm 45^{\circ} \\ & \text { continuous } \\ & \text { variable } \end{aligned}$ | 40-pin DIP | Post amplifier available as a phase shifter or signal amplifier | p80 to 84 |

## Voltage Controlled Phase Shifter (for reference signal)

| Model | Frequency range | Amount of phase shift | I/O voltage | Refer to |
| :---: | :---: | :---: | :---: | :---: |
| CD-951V4 | 1 kHz to 2 MHz | $0 \% / 180^{\circ}$ switchable, <br> $\pm 90^{\circ}$ continuous variable | C-MOS $(0 / 5 \mathrm{~V})$ | p77 to 79 |

## SV-4BL1 SV-4BL2 SV-4FL1 SV-4FL2

SV series filters are resistor tunable low-pass filters that are powered by 5 V or 3.3 V of single supply voltage. The setting of cutoff frequency is facilitated with the external resistors (4 pcs.). Butterworth and Elliptic are incorporated into filter characteristics, and the filters fall into two types (Type 1 and Type 2) according to the frequency range. The downsizing of filters has been achieved to actualize a 15 -pin single-inline package (SIP).

SV4BL1/2 : 4-pole Butterworth slow pass
SV4FL1/2 : 4-pole elliptic low pass

| Model | SV-4BL1 | SV-4FL1 | SV-4BL2 | SV-4FL2 |
| :--- | :--- | :---: | :---: | :---: |
| Filter characteristics | Butterworth low pass | Elliptic low pass | Butterworth low pass | Elliptic low pass |
| Order | 4-pole |  |  |  |

- Absolute maximum ratings

| Supply voltage | 6 V |
| :--- | :--- |
| Input voltage | Supply voltage or less |

$\boldsymbol{\nabla}$ Cut-off frequency (fc)

| Range $^{* 1}$ | 10 Hz to 10 kHz | 100 Hz to 100 kHz |
| :--- | :--- | :--- |
| Accuracy $^{* 2}$ | $\pm 3 \%$ |  |
| Setting method | Connected with external resistors (4 pcs.) |  |

$\boldsymbol{\nabla}$ Pass-band characteristic

| Gain ${ }^{3}$ | $0 \pm 0.3 \mathrm{~dB}$ |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Ripple | - | $0.28 \mathrm{dBp}-\mathrm{p}$ typ. | - | $0.28 \mathrm{dBp}-\mathrm{p}$ typ. |

## - Attenuation characteristics

| Rolloff | $24 \mathrm{~dB} /$ oct | $42 \mathrm{~dB} / o c t$ equivalent | $24 \mathrm{~dB} / o c t$ | $42 \mathrm{~dB} / o c t$ equivalent |
| :--- | :---: | :---: | :---: | :---: |
| Attenuation <br> characteristics (2fc) | 24 dB typ. | 55 dB typ. | 24 dB typ. | 55 dB typ. |
| Minimum attenuation | - | 46 dB typ. | - | 46 dB typ. |

High frequency attenuation (up to 1 MHz )

Min. 60dB

## VInput characteristics

| Input impedance | Min. $50 \mathrm{k} \Omega$ |
| :--- | :--- |
| Maximum input voltage | 5 V |
| Minimum input voltage | 0 V |
| $\mathrm{VOutput} \mathrm{characteristics}^{\text {Output impedance }}$ |  |
| Maximum output voltage | Min. 4.9 V |
| Minimum output voltage | Max. 100 mV |
| Load resistance | Min. $10 \mathrm{k} \Omega$ |
| Voltage noise | Max. $100 \mu \mathrm{Vrms}$ |
| Distortion ${ }^{* 4}$ | $0.01 \%$ typ. |
| Offset voltage ${ }^{* 5}$ | $\pm 30 \mathrm{mV}$ typ. |
| Offset drift | $30 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ typ. |
| Mid-potential <br> output accuracy ${ }^{*} 6$ | $\pm 1 \%$ |

## $\nabla$ Others

| Supply voltage | $5 \mathrm{~V}(3 \mathrm{~V}$ to 5.5 V$)$ |
| :--- | :--- |
| Quiescent current | 10 mA typ. |
| Available temperature/ <br> humidity range | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
| Storage temperature/ <br> humidity range | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions | $39 \times 15 \times 5.5 \mathrm{~mm}(15$ pin), Model S15 |
| Note: The following specifications are applied unless otherwise specified:RF= $31.8 \mathrm{k} \Omega$, Power: 5 V, Mid-potential: 2.5 V, Load: $10 \mathrm{k} \Omega$, Ambient temp.: $23 \pm 55^{\circ} \mathrm{C}$ <br> *1: As to SV series, expansion of the lower cut-off (center) frequency with the external capacitors is disabled. *2: -3 dB derived with reference to fc/10 *3: Gain in fc/10 <br> *4: Distortion in fc/10 *5: Drift from mid-potential (adjustable with a trimming resistor) *6: Mid-potential output is a supply voltage $/ 2$. |  |

## Characteristics



SV-4FL1/2


## Connection diagram



* This circuit is used for normal connection. Pin 14 needs to be dis connected if no offset calibration is required.

Example 3 (when input and output are AC-coupled)


* This circuit is used to DC-interrupt the prior and subsequent circuits. Pin 14 needs to be disconnected if no offset calibration is required.

Example 2 (when the mid-potential is externally input)


* Noise superimposed in the mid-potential exerts effects on noise characteristics if the mid-potential is externally input.
* If the mid-potential is assigned to the prior and subsequent circuits, this circuit is used. Pin 14 needs to be disconnected if no offset calibration is required.


## - Calculation of coupling capacitor

Input: $\quad \mathrm{C} 1[\mu \mathrm{~F}]=\frac{3.18}{\text { fch }[\mathrm{Hz}]}$
Output: $\quad \mathrm{C} 2[\mu \mathrm{~F}]=\frac{159}{\operatorname{Load}[\mathrm{k} \Omega] \cdot \text { fch }[\mathrm{Hz}]}$
fch: Coupling frequency ( -3 dB )
-6 dB is gained in coupling frequency (fch) if the coupling frequencies for input and output are equal.
fch: Set at $1 / 10$ or less of cut-off frequency (fc)

## Cut-off frequency setting

- Equation of external resistor RF

Type 1 $\quad \mathrm{R}_{\mathrm{F}}[\mathrm{k} \Omega]=\frac{15.9 \times 10^{3}}{\mathrm{fc}[\mathrm{Hz}]}$
Type $2 \quad \operatorname{RF}[\mathrm{k} \Omega]=\frac{159 \times 10^{3}}{\mathrm{fc}[\mathrm{Hz}]}$

Note: Resistance error results in cut-off frequency error and a deterioration of filter characteristics.
Be sure to use a resistor with tolerance of $1 \%$.
RF: $1.6 \mathrm{k} \Omega$ to $1.6 \mathrm{M} \Omega$

Resistor Tunable Filter


## SR-4BL/4FL SR-4BH/4FH SR-2BLH SR-1BP/2BP SR-2BE

SR series filters are ultrasmall resistor tunable filters in single-inline package (SIP). An easy setting of cutoff (center) frequency is assured with the external resistors. The abundance of filter types extends the range of choices.

SR-4BL1/2/3: 4-pole Butterworth low pass
SR-4FL1/2/3: 4-pole elliptic low pass
SR-4BH1/2: 4-pole Butterworth high pass
SR-4FH1/2: 4-pole elliptic high pass
SR-2BLH1/2/3: 2-pole Butterworth low/high pass
SR-1BP1/2: 1-pole pair band pass
SR-2BP1/2: 2-pole pair band pass
SR-2BE1/2: 2-pole pair band elimination

| Model | SR-4BL | SR-4FL | SR-4BH | SR-4FH | SR-2BLH | SR-1BP | SR-2BP |  |  |  |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Filter characteristics | Butterworth <br> low pass | Elliptic <br> low pass | Butterworth <br> high pass | Elliptic <br> high pass | Butterworth <br> low/high pass | Butterworth <br> band pass | Butterworth <br> band pass |  |  |  |  |  |  |
| Butterworth <br> band elimination |  |  |  |  |  |  |  |  |  |  |  |  |  |
| Order | 2-pole |  |  |  |  |  |  |  |  | 2-pole | 1-pole pair | 2-pole pair | 2-pole pair |

$\nabla$ Absolute maximum ratings
Supply voltage $( \pm$ Vs $) \quad \pm 18 \mathrm{~V}$

| Input voltage | $\pm \mathrm{Vs}$ |
| :--- | :--- |

$\boldsymbol{\nabla}$ Cut-off (fc, -3 dB )/center (fo) frequency characteristics

| Range | 40 Hz to $1.6 \mathrm{kHz*}$ |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 400 Hz to 20kHz*1 | 400 Hz to 5kHz*1 | 400Hz to 20kHz*\| | 400 Hz to $10 \mathrm{kHz}{ }^{* 1}$ |  |  |
|  | 5 kHz to 100kHz** | - | 5 kHz to $100 \mathrm{kHz}^{*+1}$ | - |  |  |
| Accuracy* ${ }^{\text {2 }}$ | Max. $\pm 3 \%$ |  |  |  |  |  |
| Setting method | Connected with external resistors (4 pcs.) |  | \|Connected with external resistors (2 pcs.)| |  | Connected with external resistors (4 pcs.) |  |
| VPass-band characteristic |  |  |  |  |  |  |
| Gain ${ }^{\text {3 }}$ | $0 \pm 0.3 \mathrm{~dB}$ | $0 \pm 1 \mathrm{~dB}$ | $0 \pm 0.3 \mathrm{~dB}$ | $0 \pm 1 \mathrm{~dB}$ |  | $0 \pm 0.3 \mathrm{~dB}$ |
| Ripple | - $\quad 0.28 \mathrm{dBp}-\mathrm{P}$ (typ) | - $\quad 0.28 \mathrm{dBp}-\mathrm{p}$ (typ) | - |  |  |  |
| Upper limit frequency (small signal) ${ }^{*}$ | - | $50 \mathrm{kHz}( \pm 1 \mathrm{~dB})$ | $\begin{gathered} 100 \mathrm{kHz} \\ ( \pm 1 \mathrm{~dB}, \mathrm{HPF})^{{ }^{5}} \end{gathered}$ | - |  | $50 \mathrm{kHz}( \pm 1 \mathrm{~dB})$ |

- Attenuation characteristics

$\operatorname{FInput}$ characteristics

| Input impedance |  | Min. $50 \mathrm{k} \Omega$ |
| :--- | :--- | :--- |
| Maximum input |  | $\leq 10 \mathrm{kHz}$ |
| voltage (linear) | $\pm 10 \mathrm{~V}$ |  |
|  | $\leq 50 \mathrm{kHz}$ | $\pm 5 \mathrm{~V}, \pm 10 \mathrm{~V}$ for 4BL3/4FL3/2BLH3 filters |

Output characteristics

| Output imp | pedance | Max. 100 |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Maximum output voltage |  | $\pm 10 \mathrm{~V}$ (Max.100kHz for 4BL3/4FL3/2BLH3 filters, Max.10kHz for other filters) |  |  |  |  |  |  |
| Load resistance |  | Min. 10k . |  |  |  |  |  |  |
| Voltage noise |  | Max. $140 \mu \mathrm{Vrms}$ ( 10 Hz to 500kHz) |  |  |  |  |  |  |
| DC offset | Voltage | Max. $\pm 30 \mathrm{mV}$ |  |  |  |  |  |  |
|  | Adjustment | Enabled |  |  |  |  |  |  |
|  | Drift | $30 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) | $15 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) |  |  |  |  | $30 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) |
| Distortion ${ }^{\text {*3 }}$ (typ) |  | 0.01\% | 0.1\% |  | 0.01\% (LPF) | 0.01\% |  |  |
| Slew rate (typ) |  | - | 2V/ $/ \mathrm{s}^{* 6}$ |  |  | - |  | 2V/ $\mu \mathrm{s}$ |
| $\nabla$ Others |  |  |  |  |  |  |  |  |
| Supply voltage |  | $\pm 15 \mathrm{~V}$ ( $\pm 5$ to $\pm 18 \mathrm{~V}$ ) |  |  |  |  |  |  |
| Quiescent current (typ) |  | $\pm 12 \mathrm{~mA}$ (Types 1\&2) $\pm 16 \mathrm{~mA}$ (Types 1\&2) <br> $\pm 27 \mathrm{~mA}$ (Type 3) $\pm 36 \mathrm{~mA}$ (Type 3) | $\pm 8 \mathrm{~mA}$ | $\pm 16 \mathrm{~mA}$ | $\begin{gathered} \pm 8 \mathrm{~mA} \text { (Types 1\&2) } \\ \pm 18 \mathrm{~mA}(\text { Type } 3) \\ \hline \end{gathered}$ | $\pm 8 \mathrm{~mA}$ | $\pm 12 \mathrm{~mA}$ | $\pm 20 \mathrm{~mA}$ |

## Temperature/ Operation <br> humidity range Storage <br> Dimensions

$-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \%$ RH
$-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \%$ RH
$51.5 \times 14 \mathrm{~mm}$, S20 type, 5.5 mm in thickness for Type 3 and 2BE filter, 4 mm in thickness for other filters
Note: The following specifications are applied unless otherwise specified: $\mathrm{Rf}=31.8 \mathrm{k} \Omega, 23 \pm 5^{\circ} \mathrm{C}, \pm 15 \mathrm{~V}$
*1: As to SR series, expansion of the lower cut-off (center) frequency with the external capacitors ( 2 or 4 pcs.) is enabled. *2: Gain in frequency (*3): 0dB *3: 4FL, 4BL: fc/ $10,4 \mathrm{FH}: 10 \mathrm{fc}(\mathrm{fc} \leq 3 \mathrm{kHz}), 3.3 \mathrm{fc}(\mathrm{fc}>3 \mathrm{kHz}$ ), $4 \mathrm{BH}: 3.3 \mathrm{fc}, 2 \mathrm{BLH}: \mathrm{LPF} \rightarrow \mathrm{fc} / 10, \mathrm{HPF} \rightarrow 10 \mathrm{fc}(\mathrm{Types} 1 \& 2)$, 3.3 fc (Type 3) *4: As to 1 BP filter, $\mathrm{Q}=10,20,30,40$, or 50 is available if a designated pin is connected with GND. Range: $1.81 \leq \mathrm{Q} \leq 50$ if connected with the external resistors *5: Type 3 : $1 \mathrm{MHz}+0, \mathrm{Max}$. -3 dB (HPF) *6: SR-2BLH3 (only): $10 \mathrm{~V} / \mu \mathrm{s}$

Resistor Tunable Filter


## SRA-4BL1 SRA-4BH1 SRA-4FL1 SRA-4FH1 SRA-2BP1

SRA series filters are power-thrifty resistor tunable filters actualizing the reduction in quiescent current to 1 to 2 mA that is $1 / 10$ of the current SR series filters. SRA series filters maintain pin-compatible with SR series filters (see P8) and become capable of operation at $\mathrm{min} . \pm 2.5 \mathrm{~V}$ of supply voltage that allows low power consumption as necessary.
Butterworth and elliptic low pass and high pass, and Butterworth band pass are incorporated into filter characteristics. An easy setting of cutoff (center) frequency is assured with the external resistors as with SR series filters, which enables a low-pass expansion with the external capacitors.

| Model | SRA-4BL1 | SRA-4FL1 | SRA-4BH1 | SRA-4FH1 | SRA-2BP1 |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Filter characteristics | Butterworth <br> low pass | Elliptic <br> low pass | Butterworth <br> high pass | Elliptic <br> high pass | Butterworth <br> band pass |
| Order | 4-pole | 4-pole | 4-pole | 4-pole | 2-pole pair |

$\boldsymbol{\nabla}$ Absolute maximum ratings

| Supply voltage ( $\pm \mathbf{V s}$ ) | $\pm 18 \mathrm{~V}$ |
| :--- | :--- |
| Input voltage | $\pm \mathrm{Vs}$ |

$\boldsymbol{\nabla}$ Cut-off (fc, -3dB)/center frequency characteristics

| Range $^{* 1}$ | 40 Hz to 1.6 kHz |
| :--- | :--- |
| Accuracy $^{* 2}$ | $\pm 3 \%$ |
| Setting method | Connected with external resistors (4 pcs.) |

$\boldsymbol{\nabla}$ Pass-band characteristic

| Gain $^{* 3}$ | $0 \pm 0.3 \mathrm{~dB}$ | $0 \pm 1 \mathrm{~dB}$ | $0 \pm 0.3 \mathrm{~dB}$ | 0 |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Ripple | - | $0.28 \mathrm{dBp}-\mathrm{p}$ | - | $0.28 \mathrm{dBp}-\mathrm{p}(\mathrm{typ})$ | - |
| Upper limit frequency <br> (small signal) ${ }^{2}$ | - |  | $50 \mathrm{kHz}( \pm 1 \mathrm{~dB})$ | - |  |

$\boldsymbol{\nabla}$ Attenuation characteristics

| Rolloff (typ) | 24dB/oct | 42dB/oct equivalent | 24dB/oct | 42dB/oct equivalent | 12dB/octBW |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Q (typ) | - | 5 |  |  |  |
| Attenuation characteristics (1/2fc or 2fc) (typ) | 24 dB | 55dB | 24 dB | 55dB | 35dB |
| Minimum attenuation (typ) | - | 46dB | - | 46dB | - |
| High frequency attenuation (up to 1 MHz ) | 70dB |  | - |  | 70dB |

$\nabla$ Input characteristics

| Input impedance | $\operatorname{Min} .50 \mathrm{k} \Omega$ |
| :--- | :--- |
| Maximum input voltage | $\pm 10 \mathrm{~V}$ |

VOutput characteristics

| Output impedance | Max. 100 |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Maximum output voltage | $\pm 10 \mathrm{~V}$ |  |  |  |
| Load resistance | Min. 10k $\Omega$ |  |  |  |
| Voltage noise | Max. 140 HVrms | Max. $200 \mu \mathrm{Vrms}$ | Max. $240 \mu \mathrm{Vrms}$ | Max. $140 \mu \mathrm{Vrms}$ |
| DC offset $\quad$ Voltage | Max. $\pm 30 \mathrm{mV}$ |  |  |  |
| Adjustment | Enabled |  |  |  |
| Drift | $30 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ | $15 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ |  |  |
| Distortion ${ }^{\text {+3 }}$ (typ) | 0.01\% | 0.1\% |  | 0.01\% |
| Slew rate (typ) | - | 10V/us |  | - |

## $\nabla$ Others

| Supply voltage |  | $\pm 2.5 \mathrm{~V}$ to $\pm 18 \mathrm{~V}$ |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Quiescent current (typ) |  | $\pm 1.5 \mathrm{~mA}$ | $\pm 2 \mathrm{~mA}$ | $\pm 1 \mathrm{~mA}$ | $\pm 2 \mathrm{~mA}$ | $\pm 1.5 \mathrm{~mA}$ |
| Temperature/ | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |  |  |  |  |
| humidity range | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |  |  |  |  |
| Dimensions |  | $51.5 \times 14 \times 4 \mathrm{~mm}$, S20 type |  |  |  |  |

Note: The following specifications are applied unless otherwise specified: $\mathrm{RF}=31.8 \mathrm{k} \Omega, 23 \pm 5^{\circ} \mathrm{C}, \pm 15 \mathrm{~V}$ (some items may fail to meet specifications if used at other supply voltage) *1: Expansion of the lower cut-off (center) frequency with the external capacitors is enabled. *2: Gain in frequency (*3): 0dB *3: 4FL, 4BL: fc/10 4FH: 10fc 4 BH : 3.3fc

## Basic connection diagram



Offset voltage adjustment


C : $0.1 \mu \mathrm{~F}$ (cer)
C : $0.1 \mu \mathrm{~F}$ (cer)

## Block diagram



SR/SRA-4BH


## Cut-off (center) frequency setting

- Equation of external resistor RF

Type $1 \quad \mathrm{R}_{\mathrm{F}}=\frac{15.9 \times 10^{3}}{\mathrm{fc} \text { or fo }}[\mathrm{k} \Omega]$
Types $2 \& 3 \mathrm{R}_{\mathrm{F}}=\frac{159 \times 10^{3}}{\text { fc or fo }}[k \Omega]$

- Equation of external resistor RF for expansion of lower cut-off (center) frequency
An external capacitor ( $\mathrm{CF}^{\prime}$ ) is used.
Type $1 \quad \mathrm{RF}_{\mathrm{F}}=\frac{159}{\left(\mathrm{C}^{\prime}+0.01\right) \times(\mathrm{fc} \text { or fo) })}[\mathrm{k} \Omega]$
Types 2\&3 $\mathrm{RF}_{\mathrm{F}}=\frac{159}{\left(\mathrm{C}_{\mathrm{F}}{ }^{\prime}+0.001\right) \times(\mathrm{fc} \text { or fo })}[\mathrm{k} \Omega]$
Note: Units: fc or fo in $\mathrm{Hz}, \mathrm{CF}^{\prime}$ in $\mu \mathrm{F}$
RF: 8 k to $400 \mathrm{k} \Omega$ ( 10 k to $400 \mathrm{k} \Omega$ for SRA series), 1.5 k to
$40 \mathrm{k} \Omega$ for Type 3 filters
Be sure to use a resistor and capacitor with tolerance of $1 \%$.


## Characteristics



4FL


4FH


1BP/2BP


1BP/2BP(Magnified view)


2BE


■Multichannel Filter 3315


This 3315 is capable of storing up to 8 SR/SRA filters that is utilized as a fixed frequency-allocated multichannel filter.
Filter characteristics vary with type of filters to be stored.

| Available filters | : All SR filters and SRA filters |
| :---: | :---: |
| Number of channels | : Max. 8 |
| fc/fo setting | : Fixed resistors (2 or 4 pcs.) are soldered to the discrete platform (accessory) and connected to the socket. |
| Supply voltage | : AC100V, $\pm 10 \%, 48 \mathrm{~Hz}$ to 62 Hz |
| Dimensions | $\text { : } 215(\mathrm{~W}) \times 88(\mathrm{H}) \times 300(\mathrm{D}) \mathrm{mm}$ |

## Application

## 8-pole low pass/ elliptic



Rv: 10 to $50 \mathrm{k} \Omega$
C : $0.1 \mu \mathrm{~F}$ (cer)

- Cut-off frequency setting (ripple: 0.53 dB )

External resistor (RF1 to RF8) is derived from the following equation.

$$
\begin{array}{ll}
R_{F 1}=R_{F 2}=R_{F 3}=R_{F 4}=R_{F} \\
R_{F 5}=1.801 R_{F} & R_{F 6}=1.221 R_{F} \\
R_{F 7}=1.797 R_{F} & R_{F 8}=0.4788 R_{F}
\end{array}
$$

Type $1 \mathrm{R}_{\mathrm{F}}=\frac{15.9 \times 10^{3}}{\mathrm{fc}}(\mathrm{k} \Omega)$
Type $2 \mathrm{R}_{\mathrm{F}}=\frac{159 \times 10^{3}}{\mathrm{fc}}(\mathrm{k} \Omega)$

- Equation of external resistor for expansion of lower cut-off frequency
Type $1 \quad \mathrm{R}_{\mathrm{F}}=\frac{159}{(\mathrm{CF}+0.01) \times \mathrm{fc}}(\mathrm{k} \Omega)$
Type $2 \mathrm{R}_{\mathrm{F}}=\frac{159}{(\mathrm{CF}+0.001) \times \mathrm{fc}}(\mathrm{k} \Omega)$

Note: Units: fc in $\mathrm{Hz}, \mathrm{CF}$ in $\mu \mathrm{F}$
*SRA series carry Type 1 filters only.



Normalized frequency [ffic]

4-pole low pass/ Bessel


Rv: 10 to $50 \mathrm{k} \Omega$
C $: 0.1 \mu \mathrm{~F}$ (cer)


Phase characteristics

## - Cut-off frequency setting

External resistor (RF1 to RF4) is derived from the following equation.

$$
\begin{array}{ll}
R_{F 1}=0.673 \times R_{F} & R_{F 2}=0.712 \times R_{F} \\
R_{F 3}=0.384 \times R_{F} & R_{F 4}=1.014 \times R_{F}
\end{array}
$$

Type $1 \quad \mathrm{R}_{\mathrm{F}}=\frac{15.9 \times 10^{3}}{\mathrm{fc}}(\mathrm{k} \Omega)$
Types 2\&3 $\quad \mathrm{R}_{\mathrm{F}}=\frac{159 \times 10^{3}}{\mathrm{fc}}(\mathrm{k} \Omega)$

- Equation of external resistor for expansion of lower cut-off frequency

Type 1

$$
\mathrm{R}_{\mathrm{F}}=\frac{159}{(\mathrm{CF}+0.01) \times \mathrm{fc}}(\mathrm{k} \Omega)
$$

Types 2\&3 $\quad \mathrm{R}_{\mathrm{F}}=\frac{159}{(\mathrm{CF}+0.001) \times \mathrm{fc}}(\mathrm{k} \Omega)$
Note : Units: fc in $\mathrm{Hz}, \mathrm{CF}$ in $\mu \mathrm{F}$
*SRA series carry Type 1 filters only.


## BCD Resistor



- Absolute maximum ratings

| Supply voltage $( \pm \mathrm{Vs})$ | $\pm 18 \mathrm{~V}$ |
| :--- | :--- |
| Input voltage | $\pm \mathrm{Vs}$ |
| Control voltage | $+5.5 \mathrm{~V},-0.5 \mathrm{~V}$ |

$\boldsymbol{\nabla}$ Frequency setting mode

| Mode | BCD 1 digit (0 to 15) |
| :--- | :--- |
|  | BCD 1 digit + 1 (1 to 16, specified pin short |
|  | (6)-(8), (10)-(11), (13)-(15), (17)-(18) $)$ |

FFrequency setting range
RD-404D1 (or RD-404D2) resistor + SR/SRA filters

| SR/SRA type |  | Type 1 |  | Type 2 |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RD mode |  | BCD | BCD+1 | BCD | BCD+1 |
| RD-404D1 | Min. | $0 \mathrm{~Hz}^{*}$ | 10 Hz | OHz* | 100 Hz |
|  | Max. | 150 Hz | 160 Hz | 1.5 kHz | 1.6 kHz |
|  | Resolution | 10 Hz | 10 Hz | 100 Hz | 100 Hz |
| RD-404D2 | Min. | $0 \mathrm{~Hz}^{*}$ | 100 Hz | $0 \mathrm{~Hz}^{*}$ | 1 kHz |
|  | Max. | 1.5 kHz | 1.6 kHz | 15kHz | 16kHz |
|  | Resolution | 100 Hz | 100 Hz | 1 kHz | 1 kHz |

*A voltage of 13 V DC is present in filter output if OHz is selected.
Parallel connection of RD-404D1/2 resistors + SR/SRA filters

| SR/SRA type |  | Type 1 |  | Type 2 |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
| RD mode | 404D2 | $B C D$ | $B C D+1$ | $B C D$ | $B C D+1$ |
|  | 404D1 | $B C D$ | $B C D$ | $B C D$ | $B C D$ |
| Min. | 0 Hz | 100 Hz | 0 Hz | 1 kHz |  |
| Max. | 1.59 kHz | 1.69 kHz | 15.9 kHz | 16.9 kHz |  |
| Resolution | 10 Hz | 10 Hz | 100 Hz | 100 Hz |  |

$\nabla$ Frequency setting accuracy
Accuracy
$\pm 1 \%$ (for RD-404D only)
$\boldsymbol{\nabla}$ Control characteristics

| Code | BCD: 1 digit $(8,4,2,1)$ |
| :--- | :--- |
| Logic and level | $0 \mathrm{~V}: \mathrm{ON}$ <br> +5 V or open: OFF |
| Level input process <br> (internal) | Pulled up to +5 V at $100 \mathrm{k} \Omega$ |

## $\nabla$ Others

| Supply voltage |  | $\pm 15 \mathrm{~V}( \pm 5 \mathrm{~V}$ to $\pm 18 \mathrm{~V})$ |
| :---: | :---: | :---: |
| Quiescent current (typ) |  | +6.2mA, -1.2 mA (typ) |
| Temperature/ humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \%$ RH |
| Dimensions |  | $51.5 \times 14 \times 4.0 \mathrm{~mm}$, S20 type |

Note: The following specifications are applied unless otherwise specified:
$23 \pm 5^{\circ} \mathrm{C}, \mathrm{Vs}= \pm 15 \mathrm{~V}$
SRA series carry Type 1 filters only.

* Potential effects on characteristics including gain and rolloff may be concerned depending on the type of SR/SRA filters to be combined with. (especially if connected in parallel)


## RD-404D1/2

RD-404D is a logic control resistor designed for SR/SRA series resistor tunable filters. The setting of cutoff (center) frequency under digital signals is enabled if RD-404D resistor is used in combination with SR/SRA series.

## HR-4BL HR-4FL HR-4BH HR-4FH HR-2BP

HR series filters are resistor tunable filters that not only realize a wide operating temperature range but ensure high reliability through the adoption of the hermetic seal method and ceramic packaging. An easy setting of cutoff (center) frequency is assured with four external resistors of the same resistance.
4-pole Butterworth and elliptic low pass and high pass, and 2-pole Butterworth band pass are incorporated into filter characteristics.
The setting range of cutoff (center) frequency falls into two types: Type $1(10 \mathrm{~Hz}$ to 1.6 kHz ) and Type $2(100 \mathrm{~Hz}$ to $100 \mathrm{kHz}(50 \mathrm{kHz})$ ).

The operating temperature range is selectable, $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ (most of industrial request) or $-55^{\circ} \mathrm{C}$ to $125^{\circ} \mathrm{C}$ (MIL-STD).
Screening meets MIL-STD and special reliable tests are available on request.

(1) Filter characteristics

4BL: 4-pole Butterworth low pass filter 4FL: 4-pole Elliptic low pass filter 4BH: 4-pole Butterworth high pass filter 4FH: 4-pole Elliptic high pass filter 2BP: 2-pole pair Butterworth band pass filter
(2)Cutoff (center) frequency setting range

1: 10 Hz to 1.6 kHz
2: 100 Hz to $100 \mathrm{kHz}(50 \mathrm{kHz})$
(3) Operating temperature range

E: $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$
M: $-55^{\circ} \mathrm{C}$ to $125^{\circ} \mathrm{C}$

## Screening

| Item | Applicable standard | Product reliability level |  |
| :---: | :---: | :---: | :---: |
|  | MIL-STD-883 | MG-B2* | MG, EG |
| Internal visual | 2017 | $\bigcirc$ | $\bigcirc$ |
| Stabilization bake | 1008 Condition C | $\bigcirc$ | - |
| Temperature cycling | 1010 Condition C | $\bigcirc$ | $\bigcirc$ |
| Constant acceleration | 2001 Condition A, in Y1 direction | $\bigcirc$ | - |
| Pre burn-in | According to specifications $23^{\circ} \mathrm{C}$ | $\bigcirc$ | - |
| Burn-in | $101585^{\circ} \mathrm{C} 160 \mathrm{H}$ | $\bigcirc$ | (48 hrs) |
| Final electrical test | Tests at normal, maximum, and minimum operating temperatures according to specifications | $\bigcirc$ | $\text { ( } 23^{\circ} \mathrm{C} \text { only) }$ |
| Seal | 1014 Fine \& Gross | $\bigcirc$ | $\bigcirc$ |
| External visual | 2009 | $\bigcirc$ | $\bigcirc$ |

* Screened if an order for 10 or more filters is received.


## Basic connection diagram


$\boldsymbol{F}$ Absolute maximum ratings

| Supply voltage $( \pm \mathbf{V s})$ |  | $\pm 18 \mathrm{~V}$ |
| :--- | :--- | :--- |
| Input voltage |  | $\pm \mathrm{Vs}$ |
| Load | $2 \mathrm{k} \Omega$ |  |
| Temperature $/$ <br> range | Operation | $\mathrm{HR}-\mathrm{XXXX}-\mathrm{EG}:-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}, \mathrm{HR}-\mathrm{XXXX}-\mathrm{MG}:-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |
|  | Storage | $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |


| Model | HR-4BL1/2 | HR-4FL1/2 | HR-4BH1/2 | HR-4FH1/2 | HR-2BP1/2 |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Filter characteristics | 4-pole Butterworth <br> low pass | 4-pole Elliptic <br> low pass | 4-pole Butterworth <br> high pass | 4-pole Elliptic <br> high pass | 2-pole pair Butterworth <br> band pass |

Cut-off (fc, -3dB)/center (fo) frequency characteristics


- Attenuation characteristics

| Rolloff | $24 \mathrm{~dB} / \mathrm{oct}$ | $42 \mathrm{~dB} /$ oct equivalent | $24 \mathrm{~dB} / o c t$ | $42 \mathrm{~dB} / \mathrm{oct}$ equivalent | $12 \mathrm{~dB} / \mathrm{oct} \mathrm{BW}$ |
| :--- | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{Q}$ | $-\quad-$ | - | - | $5 \pm 5 \%$ |  |
| Attenuation <br> characteristics*2 | 24 dB typ | 55 dB typ | 24 dB typ | 55 dB typ | 35 dB typ |
| Minimum attenuation | - | 46 dB typ | - | 46 dB typ |  |
| High frequency attenuation <br> (up to 1MHz) | Min. 70 dB | Min. 60 dB | - | - | Min. 60 dB |

FInput characteristics

| Input voltage range | $\pm 10 \mathrm{~V}$ |
| :--- | :--- |
| Input impedance | Min. $50 \mathrm{k} \Omega$ |

VOutput characteristics

## Output voltage range $\pm 10 \mathrm{~V}$

| Output impedance |  | Max. $100 \Omega$ |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Load resistance Offset voltage ${ }^{* 3}$ |  | Min. 10k $\Omega$ |  |  |  |  |
|  |  | Max. $\pm 30 \mathrm{mV}$ |  |  |  |  |
| Offset drift |  | $5 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ typ | $16 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ typ | $10 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ typ | $5 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ typ | $5 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ typ |
| Noise | Type 1 | 40 $\mu \mathrm{Vrms} \mathrm{typ}$ | $90 \mu \mathrm{Vrms} \mathrm{typ}$ | $120 \mu \mathrm{Vrms}$ typ | $190 \mu \mathrm{Vrms}$ typ | $50 \mu \mathrm{Vrms}$ typ |
|  | Type 2 | $35 \mu \mathrm{Vrms} \mathrm{typ}$ | $60 \mu \mathrm{Vrms} \mathrm{typ}$ | $100 \mu \mathrm{Vrms}$ typ | 140 $\mu \mathrm{V}$ rms typ | $45 \mu \mathrm{Vrms}$ typ |
| Distortion | Type 1 | 0.004\% typ | 0.01\% typ | 0.02\% typ | 0.02\% typ | 0.004\% typ |
|  | Type 2 | 0.003\% typ | 0.005\% typ | 0.02\% typ | 0.02\% typ | 0.002\% typ |
| Slew rate | Type 1 | - | - | 10V/ $\mu$ s typ | 10V/us typ | - |
|  | Type 2 | - | - | 25V/us typ | 25V/us typ | - |

VOthers

| Supply voltage |  | $\pm 15 \mathrm{~V}$ |
| :--- | :--- | :--- |
| Supply voltage <br> range | Type 1 | $\pm 1.5 \mathrm{~V}$ to $\pm 18 \mathrm{~V}$ |
|  | Type 2 | $\pm 5 \mathrm{~V}$ to $\pm 18 \mathrm{~V}$ |


|  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Quiescent | Type 1 | $\pm 1.5 \mathrm{~mA}$ typ | $\pm 2 \mathrm{~mA}$ typ | $\pm 1 \mathrm{~mA}$ typ | $\pm 2 \mathrm{~mA}$ typ |
| current | Type 2 | $\pm 15 \mathrm{~mA}$ typ | $\pm 20 \mathrm{~mA}$ typ | $\pm 10 \mathrm{~mA}$ typ | $\pm 1.5 \mathrm{~mA}$ typ |
| Dimensions | $20 \times 33 \times 7 \mathrm{~mm}$ (lead excluded) (24-pin DIP), KC type | $\pm 20 \mathrm{~mA}$ typ | $\pm 15 \mathrm{~mA} \mathrm{typ}$ |  |  |

*1: Expansion of the lower cut-off (center) frequency with the external capacitors ( 4 pcs .) is enabled. *2: Attenuation for low pass and band pass: 2 fc , for high pass: $1 / 2 \mathrm{fc}$
*3: Zero adjustment available

## Multichannel Filter 3314



Rear panel

This 3314 is capable of storing up to 4 HR filters that is utilized as a desktoptype fixed frequency filter.

| Available filters | All HR filters |
| :--- | :--- |
| Number of channels | Max. 4 |
|  | Continuous connection of $\mathrm{CH} 1 / 2$ with $\mathrm{CH} 3 / 4$ avail- <br> able |
| fc/fo setting | Fixed resistors (2 or 4 pcs.) are soldered to the <br> discrete platform (accessory) and connected to <br> the socket. |
| Supply voltage | AC100V $\pm 10 \%, 48 \mathrm{~Hz}$ to 62 Hz <br> Dimensions |
|  | $225(\mathrm{~W}) \times 67(\mathrm{H}) \times 250(\mathrm{D}) \mathrm{mm}$ <br> (protrusion not included) |

## Block diagram



## HR-2BP



## Cut-off (center) frequency setting

- Equation of external resistor RF

Type $1 \quad \mathrm{R}_{\mathrm{F} 1}=\mathrm{R}_{\mathrm{F} 2}=\mathrm{R}_{\mathrm{F} 3}=\mathrm{R}_{\mathrm{F} 4}=\mathrm{R}_{\mathrm{F}}$

$$
\begin{aligned}
& \text { RF }=\frac{15.9 \times 10^{3}}{\mathrm{fc} \text { or fo }[\mathrm{Hz}]}[\mathrm{k} \Omega] \\
& \text { Type 2 } \quad \mathrm{R}_{\mathrm{F} 1}=\mathrm{R}_{\mathrm{F} 2}=\mathrm{R}_{\mathrm{F} 3}=\mathrm{R}_{\mathrm{F} 4}=\mathrm{R}_{\mathrm{F}} \\
& \\
& \mathrm{R}_{\mathrm{F}}=\frac{159 \times 10^{3}}{\mathrm{fc} \text { or fo }[\mathrm{Hz}]}[\mathrm{k} \Omega]
\end{aligned}
$$

- Equation of external resistor Rf for expansion of the lower frequency with the use of a capacitor ( $C_{F}$ )

Type 1

$$
\mathrm{R}_{\mathrm{F} 1}=\mathrm{R}_{\mathrm{F} 2}=\mathrm{R}_{\mathrm{F} 3}=\mathrm{R}_{\mathrm{F} 4}=\mathrm{R}_{\mathrm{F}}
$$

$$
\mathrm{R}_{\mathrm{F}}=\frac{159}{\left(\mathrm{C}_{\mathrm{F}}[\mu \mathrm{~F}]+0.01\right) \times \mathrm{fc} \text { or fo }[\mathrm{Hz}]}[\mathrm{k} \Omega]
$$

Type 2

$$
\begin{aligned}
& R_{F 1}=R_{F 2}=R_{F 3}=R_{F 4}=R_{F} \\
& R_{F}=\frac{159}{\left(\mathrm{C}_{F}[\mu \mathrm{~F}]+0.001\right) \times \text { fc or fo }[\mathrm{Hz}]}[\mathrm{k} \Omega
\end{aligned}
$$

## Characteristics

## Amplitude



4FH


4FL


2BP


4BH



## Temperature

Cut-off frequency drift (Type 1: fc $=500 \mathrm{~Hz}$, Type 2: fc $=5 \mathrm{kHz}$ )
4BL1






2BP2


FOffset voltage drift (Type 1: $\mathrm{fc}=500 \mathrm{~Hz}$, Type 2: $\mathrm{fc}=5 \mathrm{kHz}$ )



4FL2









## RT-8FLA1/2 RT-8FLB1/2 RT-3BP1/2

RT series filters are resistor tunable filters that allocate cutoff (center) frequencies with the external resistors (6 or 8 pcs.). RT-8FLA/8FLB low pass filters possess steep attenuation characteristics, which are suited to be used as anti-aliasing filters. RT-3BP 1/ 3 -octave-band pass filter is in conformity with IEC-225 standards.

135 dB /oct or equivalent: 8-pole Elliptic low pass RT-8FLA $100 \mathrm{~dB} /$ oct or equivalent: 8-pole Elliptic low pass RT-8FLB
$1 / 3$ oct bandwidth $(Q=4.32)$ : 3-pole pair band pass RT-3BP
$\boldsymbol{\nabla}$ Absolute maximum ratings

| Supply voltage ( $\pm$ Vs) |  | $\pm 18 \mathrm{~V}$ |  |  |
| :---: | :---: | :---: | :---: | :---: |
| Input voltage |  | $\pm \mathrm{Vs}$ |  |  |
| VFilter characteristics |  |  |  |  |
| Filter characteristics |  | RT-8FLA/8FLB: 8-pole Elliptic LPF 3BP: 3-pole pair BPF |  |  |
| $\nabla \mathrm{fc}, \mathrm{fo}$ |  |  |  |  |
| Setting |  | Connected with external resistors of the same resistance RT-8FLA/8FLB: 8 pcs. <br> 3BP: 6 pcs |  |  |
| Range | Type 1 | 10 Hz to 2kHz |  |  |
|  | Type 2 | 100 Hz to 20 kHz |  |  |
| External resistors | Type 1 | $R_{F}(k \Omega)=15.9 \times 10^{3} / \mathrm{fc}$ or fo $(\mathrm{Hz})$ |  |  |
|  | Type 2 | $\mathrm{R}_{\mathrm{F}}(\mathrm{k} \Omega)=159 \times 10^{3} / \mathrm{fc}$ or fo (Hz) |  |  |
| Setting accuracy |  | Max. $\pm 2 \%$ (errors of external resistors excluded) |  |  |
| VPass-band characteristics |  |  |  |  |
| Model |  | RT-8FLA1/2 | RT-8FLB1/2 | RT-3BP1/2 |
| Gain |  | $0 \mathrm{~dB} \pm 0.1 \mathrm{~dB}(\max )$ |  | $0 \mathrm{~dB} \pm 1 \mathrm{~dB}$ (max) |
| Adjusted RF |  | - |  | $0 \mathrm{~dB} \pm 0.1 \mathrm{~dB}$ (typ) |
| Ripple (p-p) |  | 0.15 dB (typ) | 0.15dB(typ) | - |
| ( $\leq 0.9 \mathrm{fc}$ ) |  | 0.3dB(max) | 0.3dB(max) | - |
| Adjusted RF |  | 0.1 dB (typ) | 0.1 dB (typ) | - |
| Distortion |  | 0.005\%(typ)at 1kHz |  |  |

- Attenuation characteristics

| Model | RT-8FLA1/2 | RT-8FLB1/2 | RT-3BP1/2 |
| :--- | :---: | :---: | :---: |
| Rolloff | 135dB/oct equiv. | 100dB/oct equiv. | - |
| Q | - | - | $4.32(B W 1 / 30 c t)$ |
| Attenuation <br> characteristics | $86 \mathrm{~dB}($ typ $) 1.56 \mathrm{fc}$ | 92dB(typ)2.0fc | 18dB/octBW |
| Minimum attenuation | 86 dB (typ) | 106 dB (typ) | - |
| High frequency <br> attenuation <br> 10fc (fo) to 1MHz | Min. 80dB | Min. 86dB | Min. 80dB |
| VInput characteristics |  |  |  |


| Input impedance | $\operatorname{Min} .50 \mathrm{k} \Omega$ |
| :--- | :--- |
| Maximum input voltage (linear) | $\pm 10 \mathrm{~V}$ |
| VOutput characteristics |  |


| Output impedance | Max. $100 \Omega$ |  |
| :---: | :---: | :---: |
| Maximum output voltage | $\pm 10 \mathrm{~V}$ |  |
| Voltage noise (input shor) | Max. $140 \mu \mathrm{Vrms}$ (BW10Hz to 500kHz) |  |
| Offset voltage | $\pm 10 \mathrm{mV}$ (typ) adjustable |  |
| $\nabla$ Others |  |  |
| Supply voltage |  | $\pm 15 \mathrm{~V}( \pm 5$ to $\pm 18 \mathrm{~V})$ |
| Quiescent current |  | $\begin{aligned} & \text { 8FLA, 8FLB : } \pm 40 \mathrm{~mA} \text { (typ) } \\ & 3 \text { BP }: \pm 25 \mathrm{~mA} \text { (typ) } \end{aligned}$ |
| Temperature/ humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10 \%$ to $80 \% \mathrm{RH}$ |
| Dimensions |  | $54.4 \times 33.7 \times 6.5 \mathrm{~mm}$, Type H |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}, \mathrm{Vs}= \pm 15 \mathrm{~V}$

## Basic connection diagram



## Equation of external resistor RF

$$
\begin{array}{rlr}
\text { Type 1 } & \mathrm{RF}_{\mathrm{F}}=\frac{15.9 \times 10^{3}}{\mathrm{fc} \text { or fo }}(\mathrm{k} \Omega) & \text { Type } 2 \\
\mathrm{RF}=\frac{159 \times 10^{3}}{\mathrm{fc} \text { or fo }}(\mathrm{k} \Omega) \\
& \mathrm{RF}_{\mathrm{F}}=\frac{159}{(\mathrm{CEXT}+0.01) \times \mathrm{fc} \text { or } \mathrm{fo}}(\mathrm{k} \Omega) & \mathrm{RF}_{\mathrm{F}}=\frac{159}{\left(\mathrm{C}_{\mathrm{EXT}}+0.001\right) \times \mathrm{fc} \text { or fo }}(\mathrm{k} \Omega)
\end{array}
$$

Note: Units: fc or fo in Hz, Cext in $\mu \mathrm{F}$
Note: Cext is required only for expansion of the lower cut-off/center frequency (fc/fo).

## Block diagram



## Characteristics





## ■Multichannel filter 3316



This 3316 is capable of storing up to 8 RT filters that is utilized as a fixed frequency-allocated multichannel filter.
Filter characteristics vary with type of filters to be stored.

| Available filters | All RT filters |
| :--- | :--- |
| Number of channels | Max. 8 |
| fc/fo setting | Fixed resistors ( 6 or 8 pcs.) are soldered to <br> the discrete platform (accessory) and |
|  | connected to the socket. |
| Supply voltage | AC100V $, \pm 10 \%, 48 \mathrm{~Hz}$ to 62 Hz <br> $215(\mathrm{~W}) \times 88(\mathrm{H}) \times 300(\mathrm{D}) \mathrm{mm}$ <br> Dimensions <br> (protrusion not included) |

Voltage Tunable Filter


## VT-4BLA, VT-4BHA, VT-2BPA

VT-A series filters are capable of controlling frequencies with external voltage and fall into the following three types: $24 \mathrm{~dB} /$ oct low pass filter (VT-4BLA), 24dB/oct high pass filter (VT-4BHA), and 2-pole pair band pass filter ( $Q=5$; VT-2BPA).
Frequency rises to a maximum as the external control voltage is at the maximum of +10 V , which allows the low pass filters to obtain 100 kHz and high/band pass filters to obtain 20 kHz . The frequency control range has been increased by a factor of 1000 for low/high pass filters and of 100 for band pass filters. The addition of an external capacitor is to vary frequencies to lower.



The phase linear filter is completed with two additional resistors and four capacitors as shown below.

$$
\begin{aligned}
& \mathrm{C}_{1}=\frac{17.453}{\text { Max. set frequency }(\mathrm{Hz})}-0.00025(\mu \mathrm{~F}) \\
& \mathrm{C}_{2}=\frac{15.567}{\text { Max. set frequency }(\mathrm{Hz})}-0.00125(\mu \mathrm{~F})
\end{aligned}
$$

Note: Max. set frequency: 62.2 kHz


Note 1: Do not connect an unused pin with other pins. Note 2: A black circle ( $\boldsymbol{O}$ ) on the case top denotes Pin 1.

Offset voltage adjustment


With the use of constants provided in the figure, $\pm 50 \mathrm{mV}$ of DC offset voltage can be regulated upon voltage output.

## Control voltage (Vc)

Frequency characteristics of the frequency control circuit are expressed in a flat response between DC and 10 kHz . It enables cut-off frequency to vary at several tens $\mu \mathrm{s}$, which has beneficial effects on dynamic change in frequencies. If noise sources are superposed in control voltage, however, it triggers
potential fluctuations in set frequencies. Small control voltage is susceptible to noise, which may result in the instability of set cut-off frequency. Thorough elimination of noise sources from control voltage is required to regain stable frequency.


## ■Multichannel filter 3334



This 3334 is utilized as a 2 -channel desktop-type voltage control filter. Voltage control is performed through a 10 -rotating potentiometer or external voltage.

| Available filters | All VT-A filters |
| :--- | :--- |
| Number of channels | Max. 2 |
| $\mathrm{fc} /$ fo setting | Set with a 10-rotating potentiometer that is |
|  | located on the panel or external control <br>  <br>  <br>  <br> voltage. <br> Simensions <br>  <br>  <br>  <br>  <br>  <br>  <br>  <br> (protrusion not included) |

Programmable Filter


## DT-212D, DT-212DC1, DT-212DC2

DT-212 series filters are regarded as universal filters capable of controlling frequencies with digital signal. The following three types of outputs are to be obtained simultaneously: low pass filter with $12 \mathrm{~dB} /$ oct of rolloff, high pass filter with $12 \mathrm{~dB} /$ oct of rolloff, and band pass filter with $6 \mathrm{~dB} /$ oct of bandwidth. DT-212 series filters facilitate the settings of gain and $Q$ through the adoption of the external resistors, besides the configuration of filters possessing various characteristics and high-order filters.
Frequency is controlled by BCD 3 digits (12 lines). The frequency range falls into three types: 1 Hz to 1.599 kHz (DT-212DC1), 100 Hz to 159.9 kHz (DT-212DC2), and a range to be designated with the external capacitors (DT-212D).
$\boldsymbol{\nabla}$ Filter characteristics

| Type |  | Low pass, high pass, band pass |
| :---: | :---: | :---: |
| Order |  | 2 (1-pole pair) |
| Rolloff |  | 12dB/oct low pass, high pass 6dB/oct•BW band pass |
| Characteristics |  | Configuration of any high-order filters available. (with external resistors) |
| Frequency setting range (fc) |  | DT-212DC1: 1 Hz to 1.599 kHz <br> DT-212DC2 : 100 Hz to 159.9 kHz <br> DT-212D :Range specified with the external capacitors |
| Q | Range | $1 / 3$ to $1 \times 10^{6 / f c}$ |
|  | Setting | Set with external resistors. |


| Noise | Low pass : $35 \mu \mathrm{Vrms}(\mathrm{typ})$ <br> High pass : $100 \mu \mathrm{Vrms}(t y p)$ <br> Band pass: $30 \mu \mathrm{Vrms}(\mathrm{typ})$ <br> (in the 10 Hz to 500 kHz bandwidth) |
| :---: | :---: |
| Offset voltage | $\pm 20 \mathrm{mV}$ (typ) <br> Adjustable with an external trimmer potentiometer. |
| Offset voltage drift | $5 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) |
| FCut-off frequency control characteristics |  |
| Code | BCD: 3 digits, positive logic ( +5 V ) |
| Input circuit | CMOS4000 series, pulled down to GND (internal) at $100 \mathrm{k} \Omega$ |
| Accuracy | $\pm 0.1 \%$ (typ)(212D), $\pm 0.5 \%$ (typ)(212DC1/2) |


| Impedance | Specified with a gain external resistor. <br> $(10 \mathrm{k} \Omega / \mathrm{gain})$ |
| :--- | :--- |
| Maximum voltage | $\pm 10 \mathrm{~V} / \mathrm{gain}$ |
| Maximum voltage | Same as supply voltage |

-Built-in operational amplifier

| Input bias current | $200 \mathrm{nA}(\mathrm{typ})$ |
| :--- | :--- |
| f t | $10 \mathrm{MHz}(\mathrm{typ})$ |
| Slew rate | $8 \mathrm{~V} / \mathrm{\mu s}(\mathrm{typ})$ |
| VOthers |  |


| Impedance | Max. $5 \Omega$ |
| :--- | :--- |
| Maximum voltage | $\pm 10 \mathrm{~V}(\leq 100 \mathrm{kHz})$ |
| Load resistance | Min. $2 \mathrm{k} \Omega$ |
| Pass-band gain ${ }^{+1}$ | Gained with external resistors. |
| Distortion $^{+2}$ | $0.002 \%$ (typ) |

Note: The following specifications are applied unless otherwise specified:
Supply voltage: $\pm 15 \mathrm{~V}$ and +5 V , Gain: $1, \mathrm{Q}=0.707$, Ambient temp.: $23 \pm 5^{\circ} \mathrm{C}$

## Block diagram



| Supply voltage |  | $\pm 15 \mathrm{~V} \pm 10 \%+5 \mathrm{~V} \pm 10 \%$ |
| :--- | :--- | :--- |
| Quiescent current |  | typ $:+15 \mathrm{~mA} /-18 \mathrm{~mA},+2.2 \mathrm{~mA}$ <br> max: $+23 \mathrm{~mA} /-27 \mathrm{~mA},+3.3 \mathrm{~mA}$ |
| Temperature/ | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
| humidity range | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions | $54.4 \times 33.7 \times 9.4 \mathrm{~mm}$, Type HA |  |

*1: Low pass outputs are DC-coupled. High frequency characteristics of high pass outputs: Max. 500kHz
*2: Measurement point: fc/2 (low pass), 2fc (high pass), fo (band pass)

## Pinout diagram



Note *1: Do not connect an unused pin with other pins.
*2: Only external capacitors (CEXT) are available.
*3: A black circle ( $\bigcirc$ ) on the case top denotes Pin 1.

## Basic connection diagram 2-pole low pass/high pass filters



Equation of gain $\mathrm{GLP}_{\mathrm{LP}}=\mathrm{GHP}=\frac{10}{\mathrm{R}_{\mathrm{G}}}(\mathrm{I} / \mathrm{O}$ phase inversion $)$
Equation of Q

$$
\begin{aligned}
& \mathrm{Q}=\frac{\mathrm{R}_{\mathrm{G}}}{\mathrm{R}_{\mathrm{Q}}} \frac{\mathrm{R}_{\mathrm{Q}}+10}{2 \mathrm{R}_{\mathrm{G}}+10} \\
& \mathrm{RQ}=\frac{10 \mathrm{R}_{\mathrm{G}}}{\left(2 \mathrm{R}_{\mathrm{G}}+10\right) \mathrm{Q}-\mathrm{R}_{\mathrm{G}}}(\mathrm{k} \Omega)
\end{aligned}
$$

Units: $\mathrm{R}_{\mathrm{G}}$ and $\mathrm{R}_{\mathrm{Q}}$ in $\mathrm{k} \Omega$
E.g.: Determine RG and Re of Butterworth and Bessel characteristics. (Gain $=2$, a $12 \mathrm{~dB} /$ oct low pass filter assigned)
$\mathrm{R}_{\mathrm{G}}=\frac{10}{\mathrm{GLP}}=5 \mathrm{k} \Omega$
$R \mathrm{Q}=\frac{50}{20 \mathrm{Q}-5}$
$=5.469 \mathrm{k} \Omega(\mathrm{Q}=0.70711$, Butterworth $)$
$=7.637 \mathrm{k} \Omega(\mathrm{Q}=0.57735$, Bessel $)$


## Basic connection diagram 1-pole pair band pass filters



Equation of gain $\quad \mathrm{GBP}=\frac{10}{\mathrm{R}_{\mathrm{G}}}(\mathrm{I} / \mathrm{O}$ phase inversion $)$
Equation of Q

$$
\mathrm{Q}=0.5+\frac{5}{\mathrm{R}_{\mathrm{G}}}+\frac{5}{\mathrm{RQ}_{\mathrm{Q}}}
$$

$$
\mathrm{R} \mathrm{Q}=\frac{10}{2 \mathrm{Q}-1-\mathrm{GBP}}(\mathrm{k} \Omega)
$$

Units: $\mathrm{R}_{\mathrm{G}}$ and $\mathrm{R}_{\mathrm{Q}}$ in $\mathrm{k} \Omega$
E.g.: Determine Rg and Re when Q is set at 2,5 , and 10 .

$$
\text { (Gain }=5 \text {, a 1-pole pair band pass filter assigned })
$$

$$
\begin{aligned}
& \mathrm{R}_{\mathrm{G}}=\frac{10}{\mathrm{GBP}_{\mathrm{BP}}}=2 \mathrm{k} \Omega \\
& \begin{aligned}
\mathrm{R}_{\mathrm{G}} & =\frac{10}{2 \mathrm{Q}-1-5} \\
& =-5 \mathrm{k} \Omega(\mathrm{Q}=2)^{*} \\
& =2.5 \mathrm{k} \Omega(\mathrm{Q}=5) \\
& =0.71 \mathrm{k} \Omega(\mathrm{Q}=10)
\end{aligned}
\end{aligned}
$$

* The following specifications should be satisfied:
$Q \geq 3$ is obtained if a gain is " 5 ", and the maximum gain is " 3 " if $Q$ is set at 2 .


## PROGRAMMABLE FILTER

## Frequency setting

DT-212 series filters allow frequency setting through external contacts or digital signal. The frequency setting (BCD: 3 digits) is completed by assigning weights to the relevant input pins, as shown below. Internal logic reaches "Hi" if +5 V is placed to the input pin (bit) and "Lo" if the input pin is set at 0 V or open. The sum of bit weights (Hi) denotes frequency, and the frequency (fc) - sum (N) relationship is represented in the following equations.

$$
\begin{array}{ll}
\text { DT-212DC1 } & \mathrm{fc}=\mathrm{N}(\mathrm{~Hz}) \\
\text { DT-212DC2 } & \mathrm{fc}=100 \mathrm{~N}(\mathrm{~Hz}) \\
\text { DT-212D } & \mathrm{fc}=\frac{\mathrm{N}}{20 \cdot \text { CEXT }}(\mathrm{Hz}) \\
& \quad(\mathrm{CEXT}: \mu \mathrm{F})
\end{array}
$$

DT-212DC1 built-in capacitor: 50000 pF
DT-212DC2 built-in capacitor: 500 pF


Operation in TTL level requires a voltage of +3.5 or more and a power of +5 or less when Hi level is placed. If the voltage does not attain +3.5 V , connect a proper pull-up resistor to TTL output.

## Offset voltage adjustment

- When low pass or high pass output is used

- When band pass output is used



## Supply power and GND connection

DT-212 series filters are powered by $\pm 15 \mathrm{~V}$ and +5 V , and also allow a power of +5 V to be diverted from +15 V .

- When only $\pm 15 \mathrm{~V}$ is supplied

A power of +5 V is derived from the connection shown in the following diagram. The Hi level of the logic input signal should be +5.3 V at the maximum due to fluctuations in Zener voltage.
The quiescent current for $\pm 15 \mathrm{~V}$ obtains 22 mA (typ) after an increase of 7 mA .


## - When $\pm 15 \mathrm{~V}$ and +5 V are supplied

The connection of Pins (36) and (37) requires caution to prevent the return current from flowing into the analog circuit from +5 V of logic power. Pins (36) and (37) are to be connected on the power side as shown below.
Be sure to use a power of +5 V that is small in ripple and pulse noise as with $\pm 15 \mathrm{~V}$. The method with the use of only $\pm 15 \mathrm{~V}$ is adopted if a proper power of +5 V fails to be obtained.



## DT-408D DT-408DC2

DT-408 series filters are universal filters embedded with 2-stage 2-pole state variable filters. These filters facilitate the settings of gain and $Q$ through the adoption of the external resistors, besides the configuration of filters possessing various characteristics and high-order filters.
Frequency is controlled by BCD 2 digits ( 8 lines). The frequency range falls into the following two types: a range to be designated with the external capacitors (DT-408D) and 1 kHz to 159 kHz (DT-408DC2).
DT-408 series filters are Type HB $(54.4 \times 33.7 \times 8.0 \mathrm{~mm})$ with 40 -pin DIP.

## FFilter characteristics

| Type |  | Low pass, high pass, band pass |
| :---: | :---: | :---: |
| Order |  | 2 (1-pole pair) $\times 2$ |
| Rolloff |  | 12dB/oct low pass, high pass 6dB/oct • BW band pass |
| Characteristics |  | Configuration of any high-order filters available. Max. 4-pole filters per unit |
| Frequency setting range (fc) |  | DT-408D: Range specified with the external capacitors DT-408DC2 : 1 kHz to 159 kHz |
| Q | Range | Range: $1 / 3$ to $1 \times 10^{6} / \mathrm{fc}$ |
|  | Setting | Set with external resistors. |


| Noise | Low pass $: 15 \mu \mathrm{Vrms}$ (typ) <br> High pass $: 70 \mu \mathrm{Vrms}$ (typ) <br> Band pass : $30 \mu \mathrm{Vrms}$ (typ) <br> (fc $=80 \mathrm{kHz}$, in the 10 Hz to 500 kHz bandwidth) |
| :--- | :--- |
| Offset voltage | $\pm 20 \mathrm{mV}$ (typ) <br> Zero adjustment available with an external <br> trimmer potentiometer |
| Offset drift | $10 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) |
| $\boldsymbol{V C u t - o f f ~ f r e q u e n c y ~ c o n t r o l ~ c h a r a c t e r i s t i c s ~}$ |  |

$\boldsymbol{V}$ Input characteristics

| Impedance | Specified with a gain external resistor. (10k $\Omega /$ gain) |
| :--- | :--- |
| Maximum voltage | $\pm 10 \mathrm{~V} /$ gain |
| Absolute maximum <br> voltage | Same as supply voltage |

## FOutput characteristics

| Impedance | Max. $5 \Omega$ |
| :--- | :--- |
| Maximum voltage | $\pm 10 \mathrm{~V}(\leq 100 \mathrm{kHz})$ |
| Load resistance | Min. $2 \mathrm{k} \Omega$ |
| Pass-band gain | Gained with external resistors. |
| Distortion | $0.003 \%$ (typ) |

[^1]Supply voltage: $\pm 15 \mathrm{~V}$, Gain: $1, Q=0.7071$, Ambient temp.: $23 \pm 5^{\circ} \mathrm{C}$

| Code | BCD $: 2$ digits, negative logic |
| :--- | :--- |
| Input circuit | Pulled up to +5 V at $100 \mathrm{k} \Omega$ |
| Accuracy | DT-408D $: \pm 0.1 \%$ (typ) |
|  | DT-408DC2 $: \pm 0.5 \%$ (typ) |

VBuilt-in operational amplifier

| Bias current |  | 200nA (typ) |
| :---: | :---: | :---: |
| ft |  | 10MHz (typ) |
| Slew rate |  | 8V/ $/$ s (typ) |
| $\nabla$ Others |  |  |
| Supply voltage |  | $\pm 15 \mathrm{~V} \pm 10 \%$ |
| Quiescent current |  | $\pm 50 \mathrm{~mA}$ (typ) |
| Temperature/ | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
| humidity range | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions |  | $54.4 \times 33.7 \times 8.0 \mathrm{~mm}$, Type HB |

## - Pinout diagram

## Block diagram



## Basic connection diagram 2-channel 2-pole low pass/high pass filters



Low-pass characteristics


Equation of gain $G_{L P}=G_{H P}=\frac{10}{R_{G}}$
Equation of $\mathrm{Q} \quad \mathrm{RQ}_{\mathrm{Q}}=\frac{10 \mathrm{R}_{\mathrm{C}}}{\left(2 \mathrm{R}_{\mathrm{G}}+10\right) \mathrm{Q}-\mathrm{R}_{\mathrm{G}}}$
Units: $\mathrm{R}_{\mathrm{G}}$ and $\mathrm{R}_{\mathrm{Q}}$ in $\mathrm{k} \Omega$

High-pass characteristics


## Basic connection diagram 2-channel 1-pole pair band pass filters



Equation of gain $\quad \mathrm{GBP}_{\mathrm{BP}}=\frac{10}{\mathrm{R}_{\mathrm{G}}}$
Equation of $Q$
$\mathrm{R}_{\mathrm{Q}}=\frac{10}{2 \mathrm{Q}-1-\mathrm{GBP}_{\mathrm{BP}}}$
Units: $\mathrm{R}_{\mathrm{G}}$ and $\mathrm{R}_{\mathrm{Q}}$ in $\mathrm{k} \Omega$

## Frequency setting

DT-408 series filters allow cut-off (center) frequency setting through external contacts or digital signal. The frequency setting (BCD: 2 digits) is completed by assigning weights to the relevant input pins. Internal logic reaches "Lo" if +0 V is placed to the input pin and "Hi" if the input pin is set at +5 V or open. The sum of bit weights (Lo) denotes frequency.

The frequency (fc) - sum ( N ) relationship is represented in the following equations.

$$
\text { DT-408DC2 } \quad \mathrm{fc}=\mathrm{N}[\mathrm{kHz}]
$$

DT-408D $\quad \mathrm{fc}=\frac{\mathrm{N}}{2 \cdot \mathrm{CEXT}}[\mathrm{Hz}]$
Units: CExt in $\mu \mathrm{F}$

## Application 1-channel 4-pole Butterworth low pass filters

## $\mathrm{fc}=1 \mathrm{kHz}$ to 159 kHz




Application 1-channel 4-pole Butterworth high pass filters


Programmable Filter


## DT-208D DT-208DC3

DT-208 series filters are regarded as universal filters capable of controlling frequencies with digital signal (Max. set frequency: 1.59 MHz ).
The following three types of outputs are to be obtained simultaneously: low pass filter with $12 \mathrm{~dB} /$ oct of rolloff, high pass filter with $12 \mathrm{~dB} /$ oct of rolloff, and band pass filter with $6 \mathrm{~dB} /$ oct of bandwidth. DT-208 series filters facilitate the settings of gain and $Q$ through the adoption of the external resistors, besides the configuration of filters possessing various characteristics and high-order filters.
Frequency is controlled by BCD 2 digits ( 8 lines). The frequency range falls into the following two types: 10 kHz to 1.59 MHz (DT-208DC with a built-in capacitor) and a range to be designated with the external capacitors (DT-208D).

FFilter characteristics

| Type | Low pass, high pass, band pass |
| :--- | :--- |
| Order | 2 (1-pole pair) |
| Rolloff | $12 \mathrm{~dB} /$ oct low pass, high pass <br> 6dB/oct $\cdot$ BW band pass |
| Characteristics | Configuration of any high-order filters available. <br> (Used with high-speed inverter CA-102R3. <br>  |
| Established with external resistors.) |  |

$\nabla$ Input characteristics

| Impedance | Specified with a gain external resistor. (2k $\Omega /$ gain) |
| :--- | :--- |

Maximum voltage $\pm 10 \mathrm{~V} /$ gain
Absolute maximum Same as supply voltage
voltage
VOutput characteristics

| Impedance | Max. $5 \Omega$ |
| :--- | :--- |
| Maximum voltage | $\pm 10 \mathrm{~V}(\leq 1 \mathrm{MHz})$ |
| Load resistance | Min. $2 \mathrm{k} \Omega$ |
| Pass-band gain | Gained with external resistors. |
| Distortion | $0.02 \%$ (typ) |
| Noise | $60 \mu \mathrm{Vrms}$ (typ) <br> Low pass output (in the 10 Hz to 500 kHz bandwidth) |
| Offset voltage | $\pm 30 \mathrm{mV}$ (typ) Adjustable with an external trimmer potentiometer. |

FCut-off frequency control characteristics

| Code | BCD: 2 digits, negative logic |  |
| :---: | :---: | :---: |
| Input circuit | CMOS input, pulled up to +5 V at $100 \mathrm{k} \Omega$ (internal) |  |
| Accuracy | DT-208D : $\pm 0.25 \%$ (typ), DT-208DC3 : 0.5\% (typ) |  |
| $\nabla$ Others |  |  |
| Supply voltage |  | $\pm 15 \mathrm{~V}( \pm 14$ to $\pm 16)$ |
| Quiescent current |  | $\pm 50 \mathrm{~mA}$ (typ) |
| Temperature/ humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions |  | $54.4 \times 33.7 \times 9.4 \mathrm{~mm}$, Type HA |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}, \mathrm{Vs}= \pm 15 \mathrm{~V}$

## Block diagram



## Basic connection diagram

- DT-208 low pass/high pass filters


$$
\text { Gain }=\frac{2 \times 10^{3}}{\mathrm{R}_{\mathrm{G}}[\Omega]} \quad \mathrm{Re}_{\mathrm{Q}}[\mathrm{~W}]=\frac{\mathrm{R}_{\mathrm{BP}}[\Omega]}{3 \mathrm{Q}-1}(\mathrm{I} / \mathrm{O} \text { phase inversion })
$$

- DT-208 band pass filters


Gain $=\frac{\operatorname{R}_{\mathrm{BP}}[\Omega]}{\operatorname{R}_{\mathrm{G}}[\Omega]} \quad \operatorname{RQ}[\Omega]=\frac{\operatorname{R}_{\mathrm{BP}}[\Omega]}{2(\mathrm{Q}-1)}(\mathrm{I} / \mathrm{O}$ phase inversion $)$

## Determination of $C_{F}$ (DT-208D)

DT-208D filters possess no frequency-determining capacitor, which requires the installation of external CF.
The sum of bit weights when logic is controlled to "Lo" is expressed in "N".
$\mathrm{fc}[\mathrm{Hz}]=\frac{\mathrm{N}}{4 \times 10^{5} \times \mathrm{C}_{\mathrm{F}}[\mathrm{F}]}$
$\mathrm{C}_{\mathrm{F}}[\mathrm{F}]=2.5 \times 10^{-6} \times \frac{\mathrm{N}}{\mathrm{fc}[\mathrm{Hz}]}$
E.g.: When logic $(\mathrm{N})$ is 100 :

To obtain 1 MHz in $\mathrm{fc}, 250 \mathrm{pF}\left(\mathrm{C}_{\mathrm{F}}\right)$ is pre-assigned to DT-208D
(DT-208DC3 has an internal CF of 250 pF ).
The configuration of the Elliptic filters and band elimination filters with the use of DT-208 series filters requires the combination use of High-Speed Inverting Amplifier CA-102R3. See Page 55 for further information on CA-102R3.


Absolute maximum ratings

| Supply voltage ( $\pm \mathrm{Vs}$ ) | $\pm 16 \mathrm{~V}$ |
| :--- | :--- |
| Input voltage | $\pm \mathrm{Vs}$ |
| Control voltage | $\pm 5.5 \mathrm{~V}-0.5 \mathrm{~V}$ |

## $\boldsymbol{\nabla}$ Filter characteristics

| DT-5FL1/2 | 5-pole Elliptic LPF |
| :--- | :--- |
| DT-6FL1/2 | 6-pole Elliptic LPF |

$\mathbf{\nabla}$ Cut-off frequency (fc)

-Pass-band characteristics

| Gain | $0 \mathrm{~dB} \pm 0.3 \mathrm{~dB}(0.05 \mathrm{fc})$ |
| :--- | :--- |
| Ripple | $0.13 \mathrm{dBp}-\mathrm{p}$ (design center value) |
| Distortion | $0.05 \%$ (typ) |

## Basic connection diagram



* INH is set at " 0 " (open or +5 V ) to allocate cut-off frequency according to 3-bit binary signal (A, B, C). All frequency-determining resistors become opened upon setting of " 1 " ( 0 V ) at INH, which enables the setting of cut-off frequency with the external resistors. Contact us for the calculation and connection of the external resistors.


## DT-5FL1/2 DT-6FL1/2

DT-5FL/6FL series filters are low pass filters possessing steep attenuation characteristics, which are intended for anti-aliasing at A/D conversion. These filters allow cut-off frequency to be shifted at 8 positions with digital signal, which are suitable for frequent shift in sampling frequency.

$$
\begin{array}{ll}
60 \mathrm{~dB} / \text { oct or equivalent: } 5 \text {-pole elliptic low pass } & \text { DT-5FL1/2 } \\
\text { 80dB/oct or equivalent: } 6 \text {-pole elliptic low pass } & \text { DT-6FL1/2 }
\end{array}
$$



## Characteristics

DT-5FL


DT-6FL


## Multichannel filter 3344



This case outfitted with the power supply is capable of storing up to 8 DT-5FL/ 6FL filters.

| Number of channels | Max. 8 |
| :--- | :--- |
| Fc control | 8-channel batch control with a push switch on <br> the front panel (remote control available) |
| Supply voltage | AC100V, $\pm 10 \%, 48 \mathrm{~Hz}$ to 62 Hz <br> $215(\mathrm{~W}) \times 88(\mathrm{H}) \times 300$ <br> Dimensions mm <br> (protrusion not included) |
|  |  |

## PROGRAMMABLE LOWPASS FILTER



## DT-8FL1/2

DT-8FL series filters are designed as anti-aliasing filters possessing 8-pole elliptic characteristics
These filters are allocated with cut-off frequencies of 20 Hz to 20 kHz and of 100 Hz to 100 kHz that can be shifted at 10 positions in accordance with 4-bit external signal (1-, 2-, 5-sequence).
DT-8FL series filters are in 60 -pin dual-inline package (DIP) and powered by $\pm 8 \mathrm{~V}$.
$\boldsymbol{\nabla}$ Absolute maximum ratings

| Supply voltage $( \pm \mathrm{Vs})$ | $\pm 10 \mathrm{~V}$ |
| :--- | :--- |
| Input voltage | $\pm \mathrm{Vs}$ |
| Control voltage | $+8.5 \mathrm{~V},-0.5 \mathrm{~V}$ |

$\boldsymbol{\nabla}$ Filter characteristics

| Filter characteristics | 8 -pole elliptic LPF |
| :--- | :--- |

## VCut-off frequency (fc)

| Cut-off frequency <br> range* | Type 1: 20 Hz to 20 kHz <br> Type 2: 100 Hz to 100 kHz <br> $1-, 2-, 5-$-sequence |
| :--- | :--- |
| Setting | 4-bit binary code, negative logic |

$\boldsymbol{V}$ Pass-band characteristics

| Gain | $0 \mathrm{~dB} \pm 0.1 \mathrm{~dB}$ (at 0.05 fc$)$ |
| :--- | :--- |
| Ripple | $0.1 \mathrm{dBp}-\mathrm{p}$ (typ) (at DC to fc) |
| Distortion | Max. $0.013 \%$ (at $0.5 \mathrm{fc}, 1 \mathrm{Vrms})$ |

- Attenuation characteristics

| Rolloff | $130 \mathrm{~dB} /$ oct equivalent |
| :--- | :--- |
| Attenuation <br> characteristics | 82 dB (typ) (at 1.56 fc to 1 MHz ) |

VInput characteristics

| Input impedance | Min. $10 \mathrm{k} \Omega, 20 \mathrm{k} \Omega$ (typ) |
| :--- | :--- |
| Maximum input <br> voltage (linear) | $\pm 5 \mathrm{~V}$ |

## VOutput characteristics

| Output impedance |  | Max. 100 ${ }^{\text {, }} 50 \Omega$ (typ) |
| :---: | :---: | :---: |
| Maximum output voltage |  | $\pm 5 \mathrm{~V}$ |
| Voltage noise |  | Type 1: $60 \mu \mathrm{Vrms}$ (typ) <br> Type 2: $80 \mu \mathrm{Vrms}$ (typ) <br> (BW: 10Hz to 500 kHz ) |
| Offset voltage |  | $\pm 10 \mathrm{mV}$ (typ) adjustable |
| Load resistance |  | Min. $2 \mathrm{k} \Omega$ |
| $\nabla$ Others |  |  |
| Supply voltage |  | $\pm 8 \mathrm{~V} \pm 10 \%$ |
| Quiescent current |  | Type 1: $\pm 30 \mathrm{~mA}$ (typ) <br> Type 2: $\pm 72 m A$ (typ) |
| Temperature/ humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions |  | $76.7 \times 47.2 \times 8.0 \mathrm{~mm}$, Type ID |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}, \mathrm{Vs}= \pm 15 \mathrm{~V}$
${ }^{*} \mathrm{fc}=\mathrm{A}$ point passing 0 dB

## Basic connection diagram



## Control

| Control |  |  |  | Cut-off frequency [Hz] |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\mathrm{D}}$ | $\overline{\mathrm{C}}$ | $\overline{\mathrm{B}}$ | $\overline{\mathrm{A}}$ | DT-8FL1 | DT-8FL2 |
| 0 | 1 | 0 | 0 | 20 k | 100 k |
| 0 | 1 | 0 | 1 | 10 k | 50 k |
| 0 | 1 | 1 | 0 | 5 k | 20 k |
| 0 | 1 | 1 | 1 | 2 k | 10 k |
| 1 | 0 | 0 | 0 | 1 k | 5 k |
| 1 | 0 | 0 | 1 | 500 | 2 k |
| 1 | 0 | 1 | 0 | 200 | 1 k |
| 1 | 0 | 1 | 1 | 100 | 500 |
| 1 | 1 | 0 | 0 | 50 | 200 |
| 1 | 1 | 0 | 1 | 20 | 100 |

1: 0 V or GND
$0:+8 \mathrm{~V}$ or open
The control terminal is pulled up to +8 V at $100 \mathrm{k} \Omega$ for internal processing.

## Block diagram



## Characteristics



Group delay


Phase


Phase matching of cut-off frequency


## FIXED FREQUENCY FILTER

Fixed Frequency Filter


## DV series

DV series filters are semi-custom-designed fixed frequency filters that allow customers to select desirable attenuation characteristics from our various existing characteristics. These filters can be customized to your specifications including the cut-off frequency (fc), center frequency (fo), and selectivity (Q).

## DV series model and order specifications


E.g.: DV-3BL-DC denotes a 3-pole Butterworth DC-coupled low pass filter.

■Order specifications (Model and the following items are required for an order for customization.)

| Filter type | Specifications | Remarks |
| :---: | :---: | :---: |
| High pass filter Low pass filter | - Cut-off frequency | $-3 \mathrm{~dB}$ |
| Narrow band pass filter | - Center frequency <br> - Q | $\mathrm{Q}=\frac{\text { Center frequency }}{3 \mathrm{~dB} \text { bandwidth }}$ |
| Wide band pass filter | - Upper limit frequency (fch) <br> - Lower limit frequency (fcl) | -3dB each <br> Note that fcylfcllimits are imposed. |
| Band elimination filter | Center frequency |  |

-Partial modification to standard filters and customized filters
Partial modification to standard filters is available as listed below. Custom making on filters is also available by special order. Contact us for further information.

- Supply voltage of +24 V is modified to $\pm 15 \mathrm{~V}$.
- Wide-band pass filter is rendered with different attenuation characteristics between high pass and low pass.
- Q of band elimination filters is set at any number other than 5 .


## - High pass filters

| Model | DV-3BH | DV-4BH | DV-5BH | DV-6BH | DV-8FH |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Order | 3 | 4 | 5 | 6 | 8 |
| Rolloff | 18dB/oct | 24dB/oct | $30 \mathrm{~dB} / \mathrm{cct}$ | $36 \mathrm{~dB} / \mathrm{cot}$ | $75 \mathrm{~dB} / \mathrm{oct}$ |
| Attenuation characteristics | Butterworth |  |  |  | NF polar* ${ }^{1}$ |
| Cut-off frequency range | 5 Hz to 20kHz |  |  |  |  |
| Cut-off frequency accuracy ( $25^{\circ} \mathrm{C}$ ) | $\begin{aligned} & \pm 2 \%(100 \mathrm{~Hz} \leq \mathrm{fc}<10 \mathrm{kHz}), \pm 3 \%(20 \mathrm{~Hz} \leq \mathrm{fc}<100 \mathrm{~Hz}, 10 \mathrm{kHz}<\mathrm{fc}<20 \mathrm{kHz}) \\ & \pm 5 \%(5 \mathrm{~Hz} \leq \mathrm{fc}<20 \mathrm{~Hz}) \end{aligned}$ |  |  |  |  |
| Maximum input voltage (Vrms) | 3.0 |  |  |  | $\begin{aligned} & \hline 2.5 \text { (fc } \leq 3 \mathrm{kHz}) \\ & 2.0(3 \mathrm{kHz}<\mathrm{fc}) \\ & \hline \end{aligned}$ |
| Input impedance | Min. 50k $\Omega$ |  |  |  |  |
| Output impedance | Max. $100 \Omega$ |  |  |  |  |
| Load impedance | Min. 10k $\Omega$ |  |  |  |  |
| Pass-band gain | $0 \mathrm{~dB} \pm 0.5 \mathrm{~dB}$ |  |  |  | $0 \mathrm{~dB} \pm 1 \mathrm{~dB}{ }^{1}$ |
| Distortion (2Vrms) | Max. 0.5\% |  |  |  |  |
| Noise | Max. $140 \mu \mathrm{Vrms}$ ( 10 Hz to 500 kHz BW ) |  |  |  |  |
| Supply voltage | $\pm 24 \mathrm{~V}$ |  |  |  |  |
| Quiescent current (typ) | $10 \mathrm{~mA}(\mathrm{fc} \leq 5 \mathrm{kHz})$ <br> $12 \mathrm{~mA}(5 \mathrm{kHz}<\mathrm{fc})$ |  | $12 m A(f c \leq 3 k H z)$ $15 \mathrm{~mA}(3 \mathrm{kHz}<\mathrm{fc})$ | $15 \mathrm{~mA}(\mathrm{fc} \leq 3 \mathrm{kHz})$ 25 mA ( $3 \mathrm{kHz}<\mathrm{fc}$ ) |  |
| Operating temperature | Range: 0 to $50^{\circ} \mathrm{C}$ |  |  |  |  |


Note 1: Dimensions are determined with protrusions such as a connect pin excluded.
Note 2: Type B is applicable to US sockets, and all DV series filters are to be made with Type B.

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$
*1. Response hill rolloff: Min. 55dB Pass-band ripple: Max. $\pm 1 \mathrm{~dB}$

## Characteristics



## Low pass filters

VAC-coupled filters (DV-7FL excluded)


| Model ${ }^{+1}$ | $\begin{aligned} & \hline \text { 3BL-DC } \\ & \text { 3LL-DC } \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \text { 4BL-DC } \\ & \text { 4LL-DC } \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline \text { 5BL-DC } \\ & \text { 5LL-DC } \end{aligned}$ | 6BL-DC <br> 6LL-DC | 8FL-DC | 7FL |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Cut-off frequency range | 1 Hz to 20kHz |  |  |  | 5 Hz to 20kHz |  |
| Attenuation characteristics | BL: Butterworth LL: Bessel |  |  |  | NF polar |  |
| Maximum input voltage (Vrms) | 7.0 (fc $\leq 10 \mathrm{kHz}) \quad 3.0$ ( $10 \mathrm{kHz}<\mathrm{fc} \leq 20 \mathrm{kHz}$ ) |  |  |  | 2.5 | See above |
| Supply voltage | $\pm 15 \mathrm{~V}$ |  |  |  |  |  |
| Quiescent current (typ) | $\pm 12 \mathrm{~mA}$ |  |  |  | $\pm 18 \mathrm{~mA}$ |  |
| Offset voltage | $\pm 5 \mathrm{mV}\left(23 \pm 5^{\circ} \mathrm{C}\right), 100 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) |  |  |  |  |  |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$
*1: Be sure to assign "DV-" to the beginning of a model name for order. (E.g.: 3BL-DC $\rightarrow$ DV-3BL-DC) Other specifications are in conformity with AC-coupled filters.

## Characteristics




| Type | Dimensions (mm) |
| :---: | :---: |
| L | $30.8 \times 53.7 \times 18.4$ |
| ML | $40.8 \times 70.8 \times 20.2$ |
| B | $53.0 \times 53.0 \times 100.0$ |

Note 1: Dimensions are determined with protrusions such as a connect pin excluded.
Note 2: Type B is applicable to US sockets, and all DV series filters are to be made with Type B.

## FIXED FREQUENCY FILTER

## Band pass filters

VNarrow band pass filters (Specifications of model, center frequency (fo), and selectivity (Q) are required for order.)

| Model | DV-2BP | DV-3BP | DV-4BP | DV-5BP | DV-6BP |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Order | 4 (2-pole pair) | 6 (3-pole pair) | 8 (4-pole pair) | 10 (5-pole pair) | 12 (6-pole pair) |
| Rolloff | 12dB/oct BW | 18dB/oct BW | 24dB/oct BW | 30dB/oct BW | 36dB/oct BW |
| Center frequency range | 40 Hz to 20 kHz |  | 40 Hz to 10 kHz |  |  |
| Center frequency accuracy | $\pm 1 \%\left(25 \pm 5^{\circ} \mathrm{C}\right), \pm 2 \%\left(0\right.$ to $\left.50^{\circ} \mathrm{C}\right)$ |  |  |  |  |
| Q | 1 to 10 (Error: $\pm 10 \%$ ) |  |  |  |  |
| Maximum input voltage | 7Vrms |  |  |  |  |
| Input impedance | Min. 50k $\Omega$ |  |  |  |  |
| Output impedance | Max. $100 \Omega$ |  |  |  |  |
| Load impedance | Min. $10 \mathrm{k} \Omega$ |  |  |  |  |
| Pass-band gain | $0 \mathrm{~dB} \pm 1 \mathrm{~dB}$ |  |  |  |  |
| Distortion | Max. 0.1\% (1Vrms) |  |  |  |  |
| Noise | Max. $140 \mu \mathrm{Vrms}$ ( 10 Hz to 500 kHz BW) |  |  |  |  |
| Supply voltage | $\pm 15 \mathrm{~V}$ |  |  |  |  |
| Quiescent current (typ) | $\pm 12 \mathrm{~mA}$ | $\pm 20 \mathrm{~mA}$ | $\pm 24 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ | $\pm 40 \mathrm{~mA}$ |
| Operating temperature | Range: 0 to $+50^{\circ} \mathrm{C}$ |  |  |  |  |
| Type | Type L | Type ML |  | Type NL |  |

IEC (IEC-225)-compliant $1 / 3 /$ oct, $1 / 2 /$ oct, and $1 /$ oct filters adhere to $4.3,2.9,1.4$ of selectivity ( $Q$ ) each in 3BP type.
FWide band pass filters (Specifications of model, lower limit frequency (fcl), and upper limit frequency (fcн) are required for order.)

| Model | DV-3BW | DV-4BW | DV-5BW | DV-6BW | DV-8FW |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Order | 6 (3-pole pair) | 8 (4-pole pair) | 10 (5-pole pair) | 12 (6-pole pair) | 16 (8-pole pair) |
| Rolloff | $18 \mathrm{~dB} / \mathrm{oct}$ | $24 \mathrm{~dB} /$ oct | $30 \mathrm{~dB} / \mathrm{oct}$ | 36dB/oct | $75 \mathrm{~dB} / \mathrm{oct}$ |
| Attenuation characteristics | Butterworth |  |  |  | NF polar ${ }^{2}$ |
| Cut-off frequency range | 5 Hz to 20 kHz |  |  |  |  |
| Minimum bandwidth*1 | 4.0 | 3.0 | 2.5 | 2.0 | 2.0 |
| Center frequency accuracy |  |  |  |  |  |
| Maximum input voltage (Vrms) | 3.0 |  |  |  | $\begin{aligned} & 2.5\left(\mathrm{f}_{\mathrm{CH}} \leq 3 \mathrm{kHz}\right), \\ & 2.0\left(\mathrm{f}_{\mathrm{CH}}<3 \mathrm{kHz}\right) \end{aligned}$ |
| Input impedance | Min. 50k $\Omega$ |  |  |  |  |
| Output impedance | Max. $100 \Omega$ |  |  |  |  |
| Load impedance | Min. 10k $\Omega$ |  |  |  |  |
| Pass-band gain | Max. $0 \mathrm{~dB} \pm 1 \mathrm{~dB}$ |  |  |  | 0dB (+0dB, -4dB) |
| Distortion | Max. 0.5\% (2Vrms) |  |  |  |  |
| Noise | Max. $140 \mu \mathrm{Vrms}(10 \mathrm{~Hz}$ to 500 kHz BW) |  |  |  |  |
| Supply voltage | $\pm 24 \mathrm{~V}$ |  |  |  |  |
| Quiescent current (typ) | $15 \mathrm{~mA}(\mathrm{fch} \leq 5 \mathrm{kHz}), 20 \mathrm{~mA}(5 \mathrm{kHz}<\mathrm{fch})$ |  | 20 mA ( $\mathrm{fch} \leq 3 \mathrm{kHz}$ ), $25 \mathrm{~mA}(3 \mathrm{kHz}<\mathrm{fch})$ |  | 30 mA ( (cht $\leq 3 \mathrm{kHz}), 40 \mathrm{~mA}$ (3kHz<cch) |
| Operating temperature | Range: 0 to $+50^{\circ} \mathrm{C}$ |  |  |  |  |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$
*1: fch/fcl
*2: Response hill rolloff: Min. 55 dB Pass-band ripple: $\pm 1 \mathrm{~dB}$
-Dimensions


| fc (Hz) 5 | $\stackrel{100}{1}$ |  | $\stackrel{300}{1}$ | 20k |
| :---: | :---: | :---: | :---: | :---: |
| 6BW | B |  | ML |  |
| 8FW | B |  | ML |  |


| Type | Dimensions $(\mathrm{mm})$ |
| :---: | :--- |
| L | $30.8 \times 53.7 \times 18.4$ |
| ML | $40.8 \times 70.8 \times 20.2$ |
| B | $53.0 \times 53.0 \times 100.0$ |

Note 1: Dimensions are determined with protrusions such as a connect pin excluded.
Note 2: Type B is applicable to US sockets, and all DV series filters are to be made with Type B

## Characteristics



Normalized frequency [f/cc]

Wide-band pass filters


Wide-band pass filters


Normalized frequency [f/fcl] Normalized frequency [f/ch]

Band elimination filters
(Specifications of model and center frequency (fo) are required for order.)

| Model |  | DV-1BE-DC | DV-2BE-DC | DV-3BE-DC | DV-4BE-DC |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Order |  | 2 (1-pole pair) | 4 (2-pole pair) | 6 (3-pole pair) | 8 (4-pole pair) |
| Rolloff | Specified fo | Min. 26dB | Min. 40dB | Min. 60dB | Min. 70dB |
|  | Measured fo | Min. 40dB | Min. 60dB | Min. | 2 dB |
| Center frequency range |  | 40 Hz to 10 kHz |  |  |  |
| Center frequency accuracy |  | $\pm 1 \%$ (0 to $50^{\circ} \mathrm{C}$ ) |  |  |  |
| Q |  | 5 ( $\pm 10 \%$ ) |  |  |  |
| Maximum input voltage |  | 7Vrms |  |  |  |
| Input impedance |  | Min. $50 \mathrm{k} \Omega$ |  |  |  |
| Output impedance |  | Max. 100 ${ }^{\text {a }}$ |  |  |  |
| Load impedance |  | Min. $10 \mathrm{k} \Omega$ |  |  |  |
| Pass-band gain |  | OdB $\pm 0.5 \mathrm{~dB}$, Max. -1 dB at 30 kHz for upper limit frequency |  |  |  |
| Distortion |  | Max. 0.1\% (7Vrms) |  |  |  |
| Noise |  | Max. $140 \mu \mathrm{Vrms}(10 \mathrm{~Hz}$ to 500 kHz ) |  | Max. $240 \mu \mathrm{Vrms}(10 \mathrm{~Hz}$ to 500 kHz ) |  |
| Supply voltage |  | $\pm 15 \mathrm{~V}$ |  |  |  |
| Quiescent current (typ) |  | $\pm 12 \mathrm{~mA}$ | $\pm 20 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ | $\pm 40 \mathrm{~mA}$ |
| Operating temperature |  | Range: 0 to $+50^{\circ} \mathrm{C}$ |  |  |  |
| Dimensions |  | $40.8 \times 70.8 \times 20.2 \mathrm{~mm}$, Type ML |  | $53.0 \times 53.0 \times 100.0 \mathrm{~mm}$, Type B |  |

## Characteristics



Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$

## Precautions for use

-DC voltage is applied to the input/output terminals in 24 -volt filters and $\pm 15$-volt modified filters, which requires the interruption of DC voltage with a capacitor to use the relevant filters. The capacitor capacity is derived from the following equation with $\min .10 \mathrm{k} \Omega$ of load applied. The proper polarity of the capacitor (see the following figure) and withstand pressure should be assured.

## A. Low pass filters

Note that the lower limit (fL) of the pass band is determined.

$$
\mathrm{C}_{\mathrm{IN}}=\frac{32}{\mathrm{f}_{\mathrm{L}}}(\mu \mathrm{~F}) \quad \text { Cout }=5 \times \operatorname{CiN}(\mu \mathrm{F})
$$

A reduction in $\mathrm{fL}_{\mathrm{L}}$ level is limited to 0.1 dB if the above value is assigned, and $12 \mathrm{~dB} /$ oct is obtained for the reduced rolloff.

## B. High pass filters

Cut-off frequency determined by the capacitor should be $1 / 20$ of $\mathrm{f}_{\mathrm{c}}$ at the maximum. The following equation is used to obtain the value.

$$
\mathrm{C}_{\mathrm{IN}}=\frac{64}{\mathrm{f}_{\mathrm{L}}}(\mu \mathrm{~F}) \quad \text { Cout }=5 \times \operatorname{CiN}(\mu \mathrm{F})
$$

High pass frequency characteristics: Max. 1 MHz at output of 2 Vrms

## C. Band pass filters

The equation at "B. High pass filters" is also applied to derive wide band pass filters (with $\mathrm{f}_{\mathrm{CL}}$ and $\mathrm{f}_{\mathrm{CH}}$ specified).

- Be sure to use a stable power that is small in ripple and noise. Potential degradation in filter characteristics and distortion and potential reduction in the maximum input level may be concerned if the voltage that is out of the specifications (Max. 2 mV p-p for $\pm 15 \mathrm{~V}$ filters, and Max. 0.5 mV p-p for +24 V filters and $\pm 15 \mathrm{~V}$ modified filters) is applied.



## Multichannel filter DV-04/04B



This case, which is outfitted with DC power for DV filter drive, is designed to use DV filters on the desktop. It is capable of storing up to 4 DV filters*. DV-04 is designed for Types L/ML/NL, and DV-04B supports Type B ( $\pm 15 \mathrm{~V}$, DC-coupled filters). CF series filters (see Page 36) can also be embedded in DV-04 with the use of the CF/DV conversion adapter.

* The maximum quiescent current may impose limits on the number of filters to be stored.

| Available filters | Max. 4 DV filters*, Types L/ML/NL, CF series <br> (CF/DV conversion adapter used): DV-04 <br> for Type B: DV-04B |
| :--- | :--- |
| Max. quiescent current | $40 \mathrm{~mA} / 1$ channel: (DV-04) |
|  | $140 \mathrm{~mA} / 4$ channels: (DV-04B) |
| I/O terminals | BNC-R |
| Supply voltage | AC100V $\pm 10 \% 50 / 60 \mathrm{~Hz}$ |
| Dimensions | $225(\mathrm{~W}) \times 67(\mathrm{H}) \times 250(\mathrm{D}) \mathrm{mm}$ |
|  | (protrusion not included) |

[^2] us for further information.

## FIXED FREQUENCY FILTER

Fixed Frequency Filter


## CF series

CF series filters are semi-custom-designed fixed frequency filters that allow customers to select desirable characteristics from our various existing standard filter characteristics. These filters can be customized to your specifications including the cut-off frequency, center frequency, and selectivity ( $Q$ ), which requires no external components. Customization is also enabled in accordance with the relevant characteristics plot and transfer functions other than filter characteristics.
Not only the prominent downsizing but the weight reduction of filters has been actualized by capitalizing on surface mount technology, as compared with the current DV series filters.
CF series filters can also be embedded in DV-04 (see Page 35) with the use of the CF/ DV conversion adapter. Integration of CF series filters carrying 40 mA or more of quiescent current is disabled due to limits on the DV-04 current capacity.

## CF series model and order specifications

■Model

*1 Not standard characteristics.

| Absolute maximum rating | Supply voltage ( $\pm \mathrm{Vs}$ ) $\pm 18 \mathrm{~V}$ Input voltage $\pm$ Vs |
| :---: | :---: |
| Input characteristics | Input impedance: Min. $50 \mathrm{k} \Omega$ |
|  | Maximum input voltage: $\pm 10 \mathrm{~V}$ (linear) |
| Output characteristics | Output impedance: Max. $100 \Omega$ |
|  | Maximum output voltage: $\pm 10 \mathrm{~V}$ (in the pass band) |
|  | Load resistance: Min. 10k $\Omega$ |
| DC offset voltage | Max. $\pm 5 \mathrm{mV}$ |
| DC offset adjustment | Enabled |
| Supply voltage | $\pm 15 \mathrm{~V}$ ( $\pm 5$ to $\pm 18 \mathrm{~V}$ ) |
| Temperature/humidity range | $\text { Operation: } \begin{array}{r} -20^{\circ} \mathrm{C} \text { to } 70^{\circ} \mathrm{C}, \\ 10 \text { to } 95 \% \mathrm{RH} \end{array}$ |
|  | Storage: $\quad-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}$, |


-High pass filters
-Butterworth

| Model | CF-3BH | CF-4BH | CF-5BH | CF-6BH | CF-7BH | CF-8BH |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Order | 3 | 4 | 5 | 6 | 7 | 8 |
| Rolloff | 18dB/oct | $24 \mathrm{~dB} /$ oct | 30dB/oct | 36dB/oct | 42dB/oct | 48dB/oct |
| Attenuation characteristics | Butterworth |  |  |  |  |  |
| Cut-off frequency range | 1 Hz to 50 kHz |  |  |  |  |  |
| Cut-off frequency accuracy | $\pm 2 \%\left(23 \pm 5^{\circ} \mathrm{C}\right)$ |  |  |  |  |  |
| Pass-band gain | $0 \mathrm{~dB} \pm 0.5 \mathrm{~dB}$ |  |  |  |  |  |
| Maximum input voltage | $\pm 10 \mathrm{~V}$ |  |  |  |  |  |
| Distortion (7Vrms) | 0.01\% (typ) |  |  |  |  |  |
| Noise | Max. $140 \mu \mathrm{Vrms}(10 \mathrm{~Hz}$ to 500 kHz BW$)$ |  |  |  |  |  |
| Quiescent $\mathrm{fc}<20 \mathrm{kHz}$ | $\pm 8 \mathrm{~mA}$ |  | $\pm 12 \mathrm{~mA}$ |  | $\pm 16 \mathrm{~mA}$ |  |
| current (typ) $20 \mathrm{kHz} \geq \mathrm{fc}$ | $\pm 16 \mathrm{~mA}$ |  | $\pm 24 \mathrm{~mA}$ |  | $\pm 32 \mathrm{~mA}$ |  |
| Type | Type EB: 10 Hz to 50 kHz , Type HB: 1 Hz to 50 kHz |  |  |  |  |  |

## Characteristics



- Elliptic

| Model | CF-6FH |  |  | CF-7FH |  |  | CF-8FH |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Amplitude characteristics | Type A | Type B | Type C | Type A | Type B | Type C | Type A | Type B | Type C |
| Order | 6 |  |  | 7 |  |  | 8 |  |  |
| Rolloff (equivalent) | 60dB/oct | 80dB/oct | $100 \mathrm{~dB} /$ oct | 84dB/oct | 128dB/oct | 260dB/oct | 135dB/oct | $100 \mathrm{~dB} /$ oct | $274 \mathrm{~dB} /$ oct |
| Filter characteristic | Elliptic |  |  |  |  |  |  |  |  |
| Cut-off frequency range ${ }^{41}$ | 10 Hz to 50 kHz |  |  |  |  |  |  |  |  |
| Ripple 100 kHz to 1.1fc | $\pm 0.3 \mathrm{~dB}$ | $\pm 0.5 \mathrm{~dB}$ | $\pm 0.7 \mathrm{~dB}$ | $\pm 0.3 \mathrm{~dB}$ | $\pm 0.5 \mathrm{~dB}$ | $\pm 0.7 \mathrm{~dB}$ | $\pm 0.5 \mathrm{~dB}$ | $\pm 0.3 \mathrm{~dB}$ | $\pm 0.7 \mathrm{~dB}$ |
| 1.1 fc to fc | $\pm 0.7 \mathrm{~dB}$ | $\pm 1.0 \mathrm{~dB}$ | $\pm 1.2 \mathrm{~dB}$ | $\pm 0.7 \mathrm{~dB}$ | $\pm 1.0 \mathrm{~dB}$ | $\pm 1.2 \mathrm{~dB}$ | $\pm 1.0 \mathrm{~dB}$ | $\pm 0.7 \mathrm{~dB}$ | $\pm 1.2 \mathrm{~dB}$ |
| Attenuation characteristics | 80dB (typ) 0.38fc | 60dB (typ) 0.58fc | 60dB (typ) 0.66fc | 82dB (typ) 0.51fc | 62dB (typ) 0.71fc | 50dB (typ) 0.87fc | 86dB (typ) 0.64ic | 100dB (typ) 0.50fc | 64dB (typ) 0.85fc |
| Low frequency attenuation (DC to 0.1fc) | 76dB | 56dB | 55 dB | 77 dB | 57dB | 45 dB | 80dB | 95 dB | 59 dB |
| Pass-band gain | $0 \pm 0.5 \mathrm{~dB}$ |  |  |  |  |  |  |  |  |
| Distortion (7Vrms) | 0.01\% (typ) |  |  |  |  |  |  |  |  |
| Noise | Max. $140 \mu \mathrm{Vrms}$ (fc<20kHz), Max. $240 \mu \mathrm{Vrms}$ (fc $\geq 20 \mathrm{kHz}$ ) in the 10 Hz to 500 kHz BW |  |  |  |  |  |  |  |  |
| Quiescent current (typ) | $\pm 24 \mathrm{~mA}$ |  |  | $\pm 32 \mathrm{~mA}$ |  |  | $\pm 32 \mathrm{~mA}$ |  |  |
|  | $\pm 40 \mathrm{~mA}$ |  |  | $\pm 48 \mathrm{~mA}^{*}$ |  |  | $\pm 48 \mathrm{~mA}^{*}$ |  |  |
| Type | Type HB or Type EB |  |  | Type HB |  |  |  |  |  |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}, \mathrm{Vs}= \pm 15 \mathrm{~V}$
${ }^{*} 1$. $\mathrm{fc}=\mathrm{A}$ point passing 0 dB (applied to simultaneous Chebyshev filters only) *2. Integration into DV-04 is disabled due to excessive quiescent current

## Characteristics

## Low pass filters

## -Butterworth



| Elliptic |  |  |  |  | *fc: A point passing 0dB |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Model |  | CF-6FL |  |  | CF-7FL |  |  | CF-8FL |  |  |  |
| Amplitude characteristics |  | Type A | Type B | Type C | Type A | Type B | Type C | Type A | Type B | Type C |  |
| Order |  | 6 |  |  | 7 |  |  | 8 |  |  |  |
| Rolloff (equivalent) |  | 60dB/oct | 80dB/oct | $100 \mathrm{~dB} / \mathrm{oct}$ | 84dB/oct | 128dB/oct | 260dB/oct | 135dB/oct | $100 \mathrm{~dB} / \mathrm{oct}$ | $274 \mathrm{~dB} /$ oct |  |
| Filter characteristic |  | Elliptic |  |  |  |  |  |  |  |  |  |
| Cut-off frequency range |  | 10 Hz to 1MHz |  |  |  |  | 10 Hz to 100kHz | 10 Hz to 1MHz | 10 Hz to 100 kHz |  |  |
| Ripple | DC to 0.9fc | $\pm 0.3 \mathrm{~dB}$ | $\pm 0.5 \mathrm{~dB}$ | $\pm 0.7 \mathrm{~dB}$ | $\pm 0.3 \mathrm{~dB}$ | $\pm 0.5 \mathrm{~dB}$ | $\pm 0.7 \mathrm{~dB}$ | $\pm 0.5 \mathrm{~dB}$ | $\pm 0.3 \mathrm{~dB}$ | $\pm 0.7 \mathrm{~dB}$ |  |
|  | 0.9 fc to fc | $\pm 0.7 \mathrm{~dB}$ | $\pm 1.0 \mathrm{~dB}$ | $\pm 1.2 \mathrm{~dB}$ | $\pm 0.7 \mathrm{~dB}$ | $\pm 1.0 \mathrm{~dB}$ | $\pm 1.2 \mathrm{~dB}$ | $\pm 1.0 \mathrm{~dB}$ | $\pm 0.7 \mathrm{~dB}$ | $\pm 1.2 \mathrm{~dB}$ |  |
| Attenuation characteristics |  | 80dB (typ) 2.64fc | 60dB (typ) 1.71fc | 60dB (typ) 1.51fc | 82dB (typ) 1.96fc | 62dB (typ) 1.40fc | 50dB (typ) 1.15fc | 86dB (typ) 1.56fc | 100dB (typ) 2.00fc | 64 dB (typ) 1.175fc |  |
| High frequency $\mathrm{fc} \leq 100 \mathrm{kHz}$ attenuation fc>100kHz |  | Min. $76 \mathrm{~dB}^{* 1}$ | Min. 56dB*1 | Min. 55dB ${ }^{+1}$ | Min. $77 \mathrm{~dB}^{* 1}$ | Min. $57 \mathrm{~dB}^{+1}$ | Min. 45dB*1 | Min. 80dB*1 | Min. 86dB*1 | Min. 59dB*1 |  |
|  |  | Min. 64dB*2 | Min. 56dB*2 | Min. 55dB ${ }^{2}$ | Min. 60dB*2 | Min. $54 \mathrm{~dB}^{+2}$ | - | Min. 60dB*2 | - | - |  |
| Pass-band gain |  | $0 \pm 0.5 \mathrm{~dB}$ |  |  |  |  |  |  |  |  |  |
| Distortion (7Vrms) |  | 0.01\% (typ), fc $\leq 100 \mathrm{kHz}$ |  |  |  |  |  |  |  |  |  |
| Noise |  | Max. $100 \mu \mathrm{Vrms}(\mathrm{fc}<50 \mathrm{kHz})$, Max. $200 \mu \mathrm{Vrms}(50 \mathrm{kHz} \leq \mathrm{fc} \leq 100 \mathrm{kHz})(10 \mathrm{~Hz}$ to 500 kHz BW) Max. $700 \mu \mathrm{Vrms}(100 \mathrm{kHz}<\mathrm{fc} \leq 1 \mathrm{MHz})(10 \mathrm{~Hz}$ to 20 MHz BW) |  |  |  |  |  |  |  |  |  |
| Quiescent current (typ) | fc<20kHz | $\pm 24 \mathrm{~mA}$ | $\pm 24 \mathrm{~mA}$ | $\pm 24 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ |  |
|  |  | $\pm 40 \mathrm{~mA}$ | $\pm 40 \mathrm{~mA}$ | $\pm 40 \mathrm{~mA}$ | $\pm 48 \mathrm{~mA}^{*}$ | $\pm 48 \mathrm{~mA}^{*}$ | $\pm 48 \mathrm{~mA}^{*}$ | $\pm 48 \mathrm{~mA}^{*}$ | $\pm 48 \mathrm{~mA}^{*}$ | $\pm 48 \mathrm{~mA}^{*}$ |  |
|  | 100kHz¢ic 1 MHz | $\pm 45 \mathrm{~mA}^{*}$ | $\pm 45 \mathrm{~mA}^{*}$ | $\pm 45 \mathrm{~mA}^{*}$ | $\pm 50 \mathrm{~mA}^{*}$ | $\pm 50 \mathrm{~mA}^{*}$ | - | $\pm 50 \mathrm{~mA}^{*}$ | - | - |  |
| Type |  | Type EB (fc 10 Hz to 100 kHz ), Type HB (fc 10 Hz to 1 MHz ) |  |  | Type HB |  |  |  |  |  |  |

*1. Frequency range: 10 fc to 1 MHz *2. Frequency range: 2 MHz to 10 MHz *3. Integration into DV-04 is disabled due to excessive quiescent current.

## Characteristics



Normalized frequency [fffc]

CF-7FL


Normalized frequency [f/fc]

CF-8FL


Normalized frequency [f/fc]

Band pass filters

| Model | CF-2BP | CF-3BP | CF-4BP | CF-5BP | CF-6BP |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Order | 4 (2-pole pair) | 6 (3-pole pair) | 8 (4-pole pair) | 10 (5-pole pair) | 12 (6-pole pair) |
| Attenuation characteristics | 12dB/oct BW | 18dB/oct BW | 24dB/oct BW | 30dB/oct BW | 36dB/oct BW |
| Center frequency range | 1 Hz to 100 kHz |  |  |  |  |
| Center frequency accuracy | $\pm 1 \%$ (23 $\pm 5^{\circ} \mathrm{C}$ ) |  |  |  |  |
| Q | 1 to 10 (Accuracy: $\pm 5 \%$ ) |  |  |  |  |
| Pass-band gain | $0 \mathrm{~dB} \pm 0.5 \mathrm{~dB}$ |  |  |  |  |
| Maximum input voltage | $\pm 10 \mathrm{~V}$ |  |  |  |  |
| Distortion (7Vrms) | 0.01\% (typ) |  |  |  |  |
| Noise | Max. $100 \mu \mathrm{Vrms}$ (fc<50kHz), <br> Max. $200 \mu \mathrm{Vrms}$ (fc $\geq 50 \mathrm{kHz}$ ) 10 Hz to 500 kHz BW |  |  |  |  |
| Quiescent $\quad$ fc $<20 \mathrm{kHz}$ | $\pm 12 \mathrm{~mA}$ | $\pm 16 \mathrm{~mA}$ | $\pm 24 \mathrm{~mA}$ | $\pm 28 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ |
| current (typ) 20kHz<fc | $\pm 24 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ | $\pm 48 \mathrm{~mA}^{* 3}$ | $\pm 56 \mathrm{~mA}^{*}$ | $\pm 64 \mathrm{~mA}^{*}$ |
| Type | Type EB | Type HB |  |  |  |

■Band elimination filters

| Model |  | CF-1BE | CF-2BE | CF-3BE | CF-4BE |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Order |  | 2 (1-pole pair) | 4 (2-pole pair) | 6 (3-pole pair) | 8 (4-pole pair) |
| Rolloff | Specified fo | Min. 26dB | Min. 40dB | Min. 60dB | Min. 70dB |
|  | Measured fo | Min. 40dB | Min. 60dB | Min. 72dB |  |
| Center frequency range |  | 1 Hz to 50 kHz |  |  |  |
| Center frequency accuracy |  | $\pm 1 \%\left(23 \pm 5^{\circ} \mathrm{C}\right)$ |  |  |  |
| Q |  | 5 (Accuracy: $\pm 5 \%$ ) |  |  |  |
| Pass-band gain |  | $0 \mathrm{~dB} \pm 0.5 \mathrm{~dB}$ |  |  |  |
| Maximum input voltage |  | $\pm 10 \mathrm{~V}$ |  |  |  |
| Distortion (7Vrms) |  | 0.01\% (typ) |  |  |  |
| Noise |  | Max. 140 $\mu \mathrm{Vrms}(10 \mathrm{~Hz}$ to 500 kHz BW$)$ |  | Max. 240 ${ }^{\text {VVrms ( }}$ (10Hz to 500kHz BW) |  |
| Quiescent current (typ) | fc<20kHz | $\pm 8 \mathrm{~mA}$ | $\pm 16 \mathrm{~mA}$ | $\pm 24 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ |
|  | 20kHz $\leq f \mathrm{c}$ | $\pm 16 \mathrm{~mA}$ | $\pm 32 \mathrm{~mA}$ | $\pm 48 \mathrm{~mA}^{*}$ | $\pm 64 \mathrm{~mA}^{*}$ |
| Type |  | Type EB |  | Type HB |  |

*3. Integration into DV-04 is disabled due to excessive quiescent current.

## Characteristics



## Characteristics



Customization is also enabled in accordance with the relevant characteristics plot and transfer functions if no intended characteristics are observed in standard filter characteristics or no specific model is provided.

## Band Elimination Filter



- Absolute maximum ratings

| Model | Band elimination filter |
| :--- | :--- |
| Order | 1-pole pair |
| Mode | FILT mode, THRU mode |
| Setting | TTL or C-MOS, negative logic <br> Pulled up to +5 V at $100 \mathrm{k} \Omega$ |

Transfer characteristics (FILT mode)

| Center frequency (fo) | 50 Hz or 60 Hz |
| :--- | :--- |
| Setting | TTL or $\mathrm{C}-\mathrm{MOS}$, negative logic <br> Pulled up to +5 V at $100 \mathrm{k} \Omega$ |
| Q | $2.0(\mathrm{fo}=50 \mathrm{~Hz}), 2.4(\mathrm{fo}=60 \mathrm{~Hz})$ |
| Maximum attenuation | Min. $24 \mathrm{~dB}(\mathrm{fo} \pm 1 \%)$ |
| VFILT/THRU mode common characteristics |  |


| Pass-band gain | $0 \pm 0.3 \mathrm{~dB}(0.1 \mathrm{fo})$ |
| :--- | :--- |
| Upper limit frequency | 50 kHz, Max. $0 \pm 1 \mathrm{~dB}$ (small signal) |

VI/O characteristics

| Input impedance |  | Max. 60k $\Omega \pm 5 \%$ |  |
| :---: | :---: | :---: | :---: |
| Max. input voltage (linear) |  | $\pm 10 \mathrm{~V}$ |  |
| Output impedance |  | Max. $50 \Omega \pm 5 \%$ |  |
| Max. output voltage (linear) |  | $\pm 10 \mathrm{~V}$ |  |
| Offset voltage |  | Max. $\pm 10 \mathrm{mV}$ (Zero adjustment available) |  |
| Noise |  | $140 \mu \mathrm{Vrms} \mathrm{typ} \mathrm{(BW:} 10 \mathrm{~Hz}$ to 500 kHz ) |  |
| Distortion |  | Max. 0.01\% (at 1kHz, $\pm 3 \mathrm{~V}$ applied) |  |
| $\nabla$ Others |  |  |  |
| Supply voltage (Vs), <br> Quiescent current |  | $\pm 5$ to $\pm 16 \mathrm{~V}$ $\pm 0.7 \mathrm{~mA}$ (typ) <br>  -1.0 mA (typ) <br> $\pm 5 \mathrm{~V}$ $+300 \mu \mathrm{~A}$ (typ) |  |
|  |  |  |  |
| Temperature/ humidity range | Operation | $-10^{\circ} \mathrm{C}$ to $60^{\circ} \mathrm{C}, 10$ to $90 \% \mathrm{RH}$ |  |
|  | Storage | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |  |
| Dimensions |  | $51.5 \times 14.0 \times 4.0 \mathrm{~mm}$, Type S20 |  |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}, \pm 15 \mathrm{~V},+5 \mathrm{~V}$ of power

## SD-1BE

SD-1BE filter is a low-powered hybrid IC 1-pole pair band elimination filter. Mode selection is available under digital control, FILT mode or THRU mode. FILT mode can be placed at 50 Hz or 60 Hz of center frequency, and the rolloff is controlled to remain 24 dB or more even if $\pm 1 \%$ is shifted from center frequency.
The offset voltage is internally adjusted to 10 mV or lower in both THRU and FILT modes. The downsizing has been achieved to actualize a 20-pin single-inline package (SIP).

## Characteristics




## 200B/S Band Pass Filter



## CF-4FPA

CF-4FPA filter is a band pass filter designed for a 200B/S modem. This filter possesses frequencies falling into the following six types: 800, 1200, 1600, 2000, 2400, and 2800 Hz . The downsizing has been achieved to actualize a 40-pin dual-inline package in dimensions of $54.4 \times 33.7 \times 6.5 \mathrm{~mm}$.

- Absolute maximum ratings

Characteristics Frequency characteristics



CF-4FPA 1200 Hz


CF-4FPA 1600 Hz


CF-4FPA 2000 Hz


CF-4FPA 2400 Hz


Low Pass Filter for Wide Band Speech Signals


## SF-8FLC-1 compliant with CCITT Rec.G. 722

SF-8FLC-1 filter is a low pass filter intended for anti-aliasing of terminal equipment in a $64 \mathrm{kbit} / \mathrm{sec}$ of wideband transmission network. This filter possesses steep attenuation characteristics such as -25 dB at $8 \mathrm{kHz},-50 \mathrm{~dB}$ at 9 kHz , and -70 dB at 14 kHz despite 7 kHz of cut-off frequency.
Not only the prominent downsizing but the weight reduction of the filter has been realized to achieve a 20 -pin single-inline package in dimensions of $51.5 \times 14.0 \times 5.5 \mathrm{~mm}$.

## FFilter characteristics

| Filter characteristics | Compliant with CCITT Rec.G.722. |
| :--- | :--- |
| Pass-band gain | $\pm 0.5 \mathrm{~dB}(1 \mathrm{kHz}, 10 \mathrm{k} \Omega$ of load) |
| Amplitude | $+0 /-1.5 \mathrm{~dB}(50 \mathrm{~Hz}), \pm 0.5 \mathrm{~dB}(100 \mathrm{~Hz})$, |
| characteristics | $+0.5 \mathrm{~dB}(6.4 \mathrm{kHz}),+0.5 /-1.5 \mathrm{~dB}(7 \mathrm{kHz})$, |
| $\mathbf{( 1 \mathrm { kHz } = 0 \mathrm { dB } )}$ | Max. $-25 \mathrm{~dB}(8 \mathrm{kHz})$, Max. $-50 \mathrm{~dB}(9 \mathrm{kHz})$, |
|  | Max. $-70 \mathrm{~dB}(14 \mathrm{kHz})$ |
| Fixed delay | Max. $2 \mathrm{~ms}($ minimum pass band $)$ |
| Group delay response | Max. $1 \mathrm{~ms}(50 \mathrm{~Hz})$, Max. $500 \mu \mathrm{~s}(100 \mathrm{~Hz})$, |
|  | Max. $125 \mu \mathrm{~s}(200 \mathrm{~Hz})$, Max. $125 \mu \mathrm{~s}(4 \mathrm{kHz})$, |
|  | Max. $500 \mu \mathrm{~s}(6.4 \mathrm{kHz})$, Max. $1 \mathrm{~ms}(7 \mathrm{kHz})$ |
|  | (Fixed delay =0s) |

VInput characteristics

| Input impedance | Min. $50 \mathrm{k} \Omega$ |
| :--- | :--- |
| Maximum voltage | $\pm 10 \mathrm{~V}$ |

## OOutput characteristics

| Output impedance | Max. $100 \Omega$ |
| :--- | :--- |
| Max. output voltage | $\pm 10 \mathrm{~V}$ |
| Load impedance | Max. $10 \mathrm{k} \Omega$ |
| Noise | Max. $140 \mu \mathrm{Vrms}(\mathrm{BW}: 10 \mathrm{~Hz}$ to 500 kHz ) |
| Offset voltage | $\pm 30 \mathrm{mV}$ (typ) |
| Others |  |


| Supply voltage/current |  | $\pm 15 \mathrm{~V} \pm 10 \%, \pm 32 \mathrm{~mA}$ (typ) |
| :--- | :--- | :--- |
| Temperature/ |  | Operation |
| humidity range | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |  |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions | $51.5 \times 14.0 \times 5.5 \mathrm{~mm}$, Type S20 |  |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}, \mathrm{Vs}= \pm 15 \mathrm{~V}$


## Characteristics



Measured group delay response


Measured frequency response

Low Noise Amplifier


## CA-251F4

CA-251F4 amplifier is a low noise amplifier allocated with bandwidth of DC to 10 MHz . With substantially low noise maintained, outstanding DC characteristics and frequency characteristics have been actualized through the adoption of the noise reduction circuit that is the application of the negative feedback technology. Our original 6-surfaceshielded single-inline package is a great contributor to the implementation of high precision signal processing and high density mounting. CA-251F4 is powered by $\pm 15 \mathrm{~V}$, and its gain is 40 dB .

## - Absolute maximum ratings

| Supply voltage ( $\pm$ Vs) |  | $\pm 16.5 \mathrm{~V}$ |
| :---: | :---: | :---: |
| Signal input voltage |  | $\pm 1 \mathrm{~V}, \pm 0.5 \mathrm{~V}$ (with no power supplied) |
| Offset input voltage |  | $\pm \mathrm{Vs}$ |
| $\nabla$ Input |  |  |
| Input form |  | DC coupling, unbalanced single ended input |
| Input impedance |  | $1 \mathrm{M} \Omega \pm 5 \%$ (DC, Pins (1) and (2) connected, Shunt capacitance: 55pF (typ)) |
| Linear maximum input voltage |  | $\pm 100 \mathrm{mV}$ (at 1kHz) |
| Input bias current |  | $\pm 30 \mathrm{pA}$ (typ) |
| Input voltage noise density |  | Max. $1.8 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ (at 10 kHz , short-circuit in input terminal) $1.4 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ (typ) (at 10 kHz , short-circuit in input terminal) |
| Input current noise density |  | $150 f \mathrm{~A} \sqrt{\mathrm{~Hz}}$ (typ) (at 1 kHz ) |
| Input offset voltage |  | $\pm 50 \mu \mathrm{~V}$ (typ) (short-circuit in input terminal) Zero adjustment available with an external trimmer potentiometer. |
| Input DC drift |  | $\pm 2 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) (short-circuit input terminal) 0 to $40^{\circ} \mathrm{C}$ |
| $\nabla$ Output |  |  |
| Output form |  | DC coupling, unbalanced single ended output |
| Maximum output voltage |  | $\pm 10 \mathrm{~V}$ (at 1 kHz , load resistance $\geq 1 \mathrm{k} \Omega$ ) |
| Maximum output current |  | $\pm 10 \mathrm{~mA}$ |
| Slew rate |  | 110V/ $\mu \mathrm{s}$ (typ) |
| Output impedance |  | $50 \Omega \pm 5 \%$ (DC) |
| $\nabla$ Amplifier |  |  |
| Voltage gain |  | 40 $\pm 0.2 \mathrm{~dB}$ (at 1 kHz ) |
| Voltage gain frequency characteristics |  | DC to $10 \mathrm{MHz}(+0.5 /-3 \mathrm{~dB})$ |
| I/O phase |  | In-phase |
| Harmonics distortion |  | 0.006\% (typ) (at $1 \mathrm{kHz}, \pm 10 \mathrm{~V}$ output) |
| $\nabla$ Power supply |  |  |
| Recommended power supply voltage range |  | $\pm 15 \mathrm{~V} \pm 1 \mathrm{~V}$ |
| Quiescent current |  | $\pm 30 \mathrm{~mA}, \pm 25 \mathrm{~mA}$ (typ) |
| $\nabla$ Environment |  |  |
| Specified temperature range |  | $23 \pm 5^{\circ} \mathrm{C}$ |
| Temperature/ humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $90 \% \mathrm{RH}$ |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| $\nabla$ Dimensions |  |  |
| Type |  | Type SS20 (20-pin shielded SIP) |
| Dimensions |  | $67 \times 10.5 \times 20 \mathrm{~mm}$ (protrusion not included) |
| Weight (NET) |  | Approx. 20g |
| Note: The following Supply voltag | specificatio ge: $\pm 15 \mathrm{~V}$, | s are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$, resistance: $1 \mathrm{M} \Omega$ |

## Basic connection diagram




## Characteristics



## Notes

Proper connection between the case ground and the GND potential should always be assured. No sufficient shielding effect is produced if disregarded.
No signal traces should be assigned on the maximum visible outline of the component mounting surface. Possible contact between the metal case and the board is observed around the maximum visible outline, which triggers the establishment of a short circuit between the signal and case. A ground plane pattern is recommended to incorporate into the maximum visible outline and the inside of the case to enhance shielding effect.

- The maximum input voltage is $\pm 0.5 \mathrm{~V}$ when the module is not in action (no power being supplied). Potential damage to the module may be concerned if the maximum voltage is violated. If a voltage of $\pm 0.5 \mathrm{~V}$ or more is input, a protective circuit is inserted into the input terminal.
- The maximum input amplitude is $\pm 1 \mathrm{~V}$ when the module is in action. If signal amplitude of $\pm 1 \mathrm{~V}$ or more is input, a protective circuit is inserted into the input terminal.
- The series regulator type power supply is required to ensure low noise. Switching noise lies in the switching regulator type power supply such as a DC-DC converter, which impairs low noise in the module.


Pattern dimensions


Maximum outer dimensions

## Evaluation board

A module-mounted evaluation board is available for easy evaluation of this module. Contact us for further information.


## LOW NOISE DIFFERENTIAL AMPLIFIER

Low Noise FET Differential Amplifier

$\boldsymbol{\nabla}$ Absolute maximum ratings

| Supply voltage ( $\pm$ Vs) | $\pm 16.5 \mathrm{~V}$ |  |
| :--- | :--- | :--- |
| Signal input <br> voltage | Differential <br> input | $\pm 1 \mathrm{~V}, \pm 0.7 \mathrm{~V}$ (with no power supplied) |
|  | Common <br> mode input | $\pm \mathrm{Vs}, \pm 0.7 \mathrm{~V}$ (with no power supplied) |
| Offset input voltage |  | $\pm \mathrm{Vs}$ |

$\nabla$ Input

| Input form | DC coupling, differential input |
| :---: | :---: |
| Differential input impedance | 2G $\Omega$ (typ) (DC, single ended) |
|  | Shunt capacitance: 22 pF (typ) |
| Common mode input impedance | $1 \mathrm{G} \Omega$ (typ) (DC) |
|  | Shunt capacitance: 44pF (typ) |
| Linear maximum differential input voltage | $\pm 100 \mathrm{mV}$ (at 1kHz) |
| Linear maximum common mode input voltage | $\pm 5 \mathrm{~V}$ (at 1kHz) |
| Input bias current | $\pm 50 \mathrm{pA}$ (typ) |
| Input offset current | $\pm 10 \mathrm{pA}$ (typ) |
| CMRR (RTI) | 110 dB (at 60 Hz ) |
|  | 70dB (typ) (at 1MHz) |
| Input voltage noise density | Max. $3 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ (at 10kHz) |
|  | $2.5 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ (typ) (at 10kHz) |
| Input current noise density | $100 \mathrm{fA} / \sqrt{\mathrm{Hz}}$ (typ) (at 1kHz) |
| Input offset voltage | $\pm 50 \mu \mathrm{~V}$ (typ) (short-circuit in input terminal) |
|  | Zero adjustment available with an external trimmer potentiometer. |
| Input DC drift | $\pm 2 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) (short-circuit in input terminal) 0 to $40^{\circ} \mathrm{C}$ |

## $\nabla$ Output

| Output form | DC coupling, single ended output |
| :---: | :---: |
| Maximum output voltage | $\pm 10 \mathrm{~V}$ (at 1kHz, load resistance $\geq 1 \mathrm{k} \Omega$ ) |
| Maximum output current | $\pm 10 \mathrm{~mA}$ |
| Slew rate | 110V/ $/$ s (typ) |
| Output impedance | $50 \Omega \pm 5 \%$ (DC) |
| VAmplifier |  |
| Voltage gain | $40 \pm 0.2 \mathrm{~dB}$ (at 1 kHz ) |
| Voltage gain frequency characteristics | DC to $10 \mathrm{MHz}(+0.5 /-3 \mathrm{~dB})$ |
| Harmonics distortion | 0.008\% (typ) (at 1kHz, $\pm 1 \mathrm{~V}$ output) |

- Power supply

| Recommended power <br> supply voltage range | $\pm 15 \mathrm{~V} \pm 1 \mathrm{~V}$ |
| :--- | :--- |
| Quiescent current | $\pm 40 \mathrm{~mA}, \pm 32 \mathrm{~mA}$ (typ) |

VEnvironment

| Temperature/ <br> humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $90 \% \mathrm{RH}$ |
| :--- | :--- | :--- |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |

## $\nabla$ Dimensions

| Type | Type SS20 (20-pin shielded SIP) |
| :--- | :--- |
| Dimensions | $67 \times 10.5 \times 20 \mathrm{~mm}$ (protrusion not included) |
| Weight (NET) | Approx. 20 g |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$, Supply voltage: $\pm 15 \mathrm{~V}$, Load resistance: $1 \mathrm{M} \Omega$

## CA-451F4

CA-451F4 amplifier is a FET input low noise differential amplifier, which ensures not only $2.5 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ of input voltage noise density but $\times 100$ voltage gain. With substantially low noise maintained, outstanding DC characteristics $\left(2 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}\right)$ and frequency characteristics (DC to 10 MHz ) have been actualized through the adoption of the noise reduction circuit that is the application of the negative feedback technology. FET input is incorporated into CA-451F4 amplifier, which delivers low noise characteristics up to high signal source impedance ( $100 \mathrm{k} \Omega$ ). Our original 6-surface-shielded 20 -pin singleinline package is a great contributor to the implementation of high precision signal processing and high density mounting.

## Basic connection diagram



## Block diagram



## Characteristics



## Notes

Proper connection between the case ground and the GND potential should always be assured. No sufficient shielding effect is produced if disregarded.
No signal traces should be assigned on the maximum visible outline of the component mounting surface. Possible contact between the metal case and the board is observed around the maximum visible outline, which triggers the establishment of a short circuit between the signal and case. A ground plane pattern is recommended to incorporate into the maximum visible outline and the inside of the case to enhance shielding effect.

- The maximum input voltage is $\pm 0.5 \mathrm{~V}$ when the module is not in action (no power being supplied). Potential damage to the module may be concerned if the maximum voltage is violated. If a voltage of $\pm 0.5 \mathrm{~V}$ or more is input, a protective circuit is inserted into the input terminal.
- The maximum input amplitude is $\pm 1 \mathrm{~V}$ when the module is in action. If signal amplitude of $\pm 1 \mathrm{~V}$ or more is input, a protective circuit is inserted into the input terminal.
- The series regulator type power supply is required to ensure low noise. Switching noise lies in the switching regulator type power supply such as a DC-DC converter, which impairs low noise in the module.


Pattern dimensions


## Maximum outer dimensions

## Evaluation board

A module-mounted evaluation board is available for easy evaluation of this module. Contact us for further information.


Low Noise Amplifier


## CA-261F2

CA-261F2 amplifier is a low noise amplifier allocated with bandwidth of DC to 10 MHz . With substantially low noise maintained, outstanding DC characteristics and frequency characteristics have been actualized through the adoption of the noise reduction circuit that is the application of the negative feedback technology. Our original 6-surfaceshielded single inline package is a great contributor to the implementation of high precision signal processing and high density mounting. CA-261F2 is powered by $\pm 15 \mathrm{~V}$, and its gain is 40 dB .

## VAbsolute maximum ratings

| Supply voltage ( $\pm$ Vs) |  | $\pm 16.5 \mathrm{~V}$ |
| :---: | :---: | :---: |
| Signal input voltage |  | $\pm 1 \mathrm{~V}, \pm 0.5 \mathrm{~V}$ (with no power supplied) |
| Offset input voltage |  | $\pm \mathrm{Vs}$ |
| VInput |  |  |
| Input form |  | DC coupling, unbalanced single ended input |
| Input impedance |  | $100 \mathrm{k} \Omega \pm 5 \%$ (DC, Pins (1) and (2) connected, Shunt capacitance: 80pF (typ)) |
| Linear maximum input voltage |  | $\pm 100 \mathrm{mV}$ (at 1kHz) |
| Input bias current |  | $\pm 20 \mathrm{nA}$ (typ) |
| Input voltage noise density |  | Max. $0.9 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ (at 1 kHz , short-circuit in input terminal $0.8 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ (at 1 kHz , short-circuit in input terminal) |
| Input current noise density |  | $1.5 \mathrm{pA} / \sqrt{\mathrm{Hz}}$ (typ) (at 10kHz) |
| Input offset voltage |  | $\pm 20 \mu \mathrm{~V}$ (typ) (short-circuit in input terminal) Zero adjustment available with an external trimmer potentiometer. |
| Input DC drift |  | $\pm 0.3 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) (short-circuit input terminal) 0 to $40^{\circ} \mathrm{C}$ |
| $\nabla$ Output |  |  |
| Output form |  | DC coupling, unbalanced single ended output |
| Maximum output voltage |  | $\pm 10 \mathrm{~V}$ (at 1 kHz , load resistance $\geq 1 \mathrm{k} \Omega$ ) |
| Maximum output current |  | Min. $\pm 10 \mathrm{~mA}$ |
| Slew rate |  | 10V/ $\mu \mathrm{s}$ (typ) |
| Output impedance |  | $50 \Omega \pm 5 \%$ (DC) |
| $\nabla$ Amplifier |  |  |
| Voltage gain |  | $40 \pm 0.2 \mathrm{~dB}$ (at 1 kHz ) |
| Voltage gain frequency characteristics |  | DC to $200 \mathrm{kHz}(+0.5 /-3 \mathrm{~dB})$ |
| I/O phase |  | In-phase |
| Harmonics distortion |  | 0.006\% (typ) (at $1 \mathrm{kHz}, \pm 10 \mathrm{~V}$ output) |
| $\nabla$ Power supply |  |  |
| Recommended power supply voltage range |  | $\pm 15 \mathrm{~V} \pm 1 \mathrm{~V}$ |
| Quiescent current |  | Max. $\pm 30 \mathrm{~mA}, \pm 22 \mathrm{~mA}$ (typ) |
| $\nabla$ Environment |  |  |
| Specified temperature range |  | $23 \pm 5^{\circ} \mathrm{C}$ |
| Temperature/ humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $90 \% \mathrm{RH}$ |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| $\nabla$ Dimensions |  |  |
| Type |  | Type SS20 (20-pin shielded SIP) |
| Dimensions |  | $67 \times 10.5 \times 20 \mathrm{~mm}$ (protrusion not included) |
| Weight (NET) |  | Approx. 20 g |

## Basic connection diagram



## Block diagram



Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$, Supply voltage: $\pm 15 \mathrm{~V}$, Load resistance: $1 \mathrm{M} \Omega$

## Characteristics

Frequency


Input voltage noise density


Noise figure


Offset drift


Gain drift


## Notes

Proper connection between the case ground and the GND potential should always be assured. No sufficient shielding effect is produced if disregarded.
No signal traces should be assigned on the maximum visible outline of the component mounting surface. Possible contact between the metal case and the board is observed around the maximum visible outline, which triggers the establishment of a short circuit between the signal and case. A ground plane pattern is recommended to incorporate into the maximum visible outline and the inside of the case to enhance shielding effect.

- The maximum input voltage is $\pm 0.5 \mathrm{~V}$ when the module is not in action (no power being supplied). Potential damage to the module may be concerned if the maximum voltage is violated. If a voltage of $\pm 0.5 \mathrm{~V}$ or more is input, a protective circuit is inserted into the input terminal.
- The maximum input amplitude is $\pm 1 \mathrm{~V}$ when the module is in action. If signal amplitude of $\pm 1 \mathrm{~V}$ or more is input, a protective circuit is inserted into the input terminal.
- The series regulator type power supply is required to ensure low noise. Switching noise lies in the switching regulator type power supply such as a DC-DC converter, which impairs low noise in the module.


Pattern dimensions


Maximum outer dimensions

## Evaluation board

A module-mounted evaluation board is available for easy evaluation of this module. Contact us for further information.


## LOW NOISE DIFFERENTIAL AMPLIFIER

Low Noise Differential Amplifier

$\operatorname{FAbsolute~maximum~ratings~}$

| Supply voltage ( $\pm$ Vs) |  | $\pm 16.5 \mathrm{~V}$ |
| :---: | :---: | :---: |
| Signal input voltage | Differential input | $\pm 1 \mathrm{~V}, \pm 0.7 \mathrm{~V}$ (with no power supplied) |
|  | Common mode input | $\pm \mathrm{Vs}, \pm 0.7 \mathrm{~V}$ (with no power supplied) |
| Offset input voltage |  | $\pm$ Vs |
| $\nabla$ Input |  |  |
| Input form |  | DC coupling, balanced differential input |
| Differential input impedance |  | $100 \mathrm{k} \Omega$, Max. $5 \%$ (DC, single ended) Shunt capacitance: 80pF (typ) |
| Common mode input impedance |  | $\begin{aligned} & \text { 500k } \Omega \text { (typ) (DC) } \\ & \text { Shunt capacitance: } 130 \mathrm{pF} \text { (typ) } \end{aligned}$ |
| Linear maximum differential input voltage |  | $\pm 100 \mathrm{mV}$ (at 1kHz) |
| Linear maximum common mode input voltage |  | $\pm 10 \mathrm{~V}$ (at 1kHz) |
| Input bias current |  | $\pm 30 \mathrm{nA}$ (typ) |
| Input offset current |  | $\pm 7 \mathrm{nA}$ (typ) |
| CMRR (RTI) |  | Min. 100dB, 120dB (typ) (at 60Hz) |
| Input voltage noise density |  | Max. $1.8 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ (at 1 kHz , short circuit in input terminal) $1.5 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ (typ) (at 1 kHz , short circuit in input terminal) |
| Input current noise density |  | $2.5 \mathrm{pA} / \sqrt{\mathrm{Hz}}$ (typ) (at 10kHz) |
| Input offset voltage |  | $\pm 40 \mu \mathrm{~V}$ (typ) (short-circuit in input terminal) Zero adjustment available with an external trimmer potentiometer. |
| Input DC drift |  | $\pm 0.3 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) (short-circuit in input terminal) 0 to $40^{\circ} \mathrm{C}$ |

## VOutput

| Output form | Unbalanced single ended output |
| :--- | :--- |
| Maximum output voltage | $\pm 10 \mathrm{~V}$ (at 1 kHz , load resistance $\geq 1 \mathrm{k} \Omega$ ) |
| Maximum output current | $\pm 10 \mathrm{~mA}$ |
| Slew rate | $10 \mathrm{~V} / \mu \mathrm{s}$ (typ) |
| Output impedance | $50 \Omega \pm 5 \%$ (DC) |
| Amplifier | $40 \pm 0.2 \mathrm{~dB}$ (at 1 kHz$)$ |
| Voltage gain DC to $200 \mathrm{kHz}(+0.5 /-3 \mathrm{~dB})$ <br> Voltage gain frequency <br> characteristics $0.006 \%$ (typ) (at $1 \mathrm{kHz}, \pm 10 \mathrm{~V}$ output) |  |
| Harmonics distortion |  |

- Power supply

| Recommended power <br> supply voltage range | $\pm 15 \mathrm{~V} \pm 1 \mathrm{~V}$ |
| :--- | :--- |
| Quiescent current | $\pm 30 \mathrm{~mA}, \pm 22 \mathrm{~mA}$ (typ) |

VEnvironment

| Temperature/ <br> humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $90 \% \mathrm{RH}$ |
| :--- | :--- | :--- |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |

$\nabla$ Dimensions

| Type | Type SS20 (20-pin shielded SIP) |
| :--- | :--- |
| Dimensions | $67 \times 10.5 \times 20 \mathrm{~mm}$ (protrusion not included) |
| Weight (NET) | Approx. 20 g |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$, Supply voltage: $\pm 15 \mathrm{~V}$, Load resistance: $1 \mathrm{M} \Omega$

## CA-461F2

CA-461F2 amplifier is a low noise differential amplifier, which ensures not only $1.5 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ of input voltage noise density but $\times 100$ voltage gain.
With substantially low noise maintained, outstanding DC characteristics $\left(0.3 \mu /{ }^{\circ} \mathrm{C}\right)$ and frequency characteristics ( DC to 200 kHz ) have been actualized through the adoption of the noise reduction circuit that is the application of the negative feedback technology. Bipolar input is incorporated into CA-461F2 amplifier, which delivers low noise characteristics up to low signal source impedance ( $500 \Omega$ or less).
Our original 6-surface-shielded 20-pin single-inline package is a great contributor to the implementation of high precision signal processing and high density mounting.

## Basic connection diagram




## Characteristics



## Notes

Proper connection between the case ground and the GND potential should always be assured. No sufficient shielding effect is produced if disregarded.
No signal traces should be assigned on the maximum visible outline of the component mounting surface. Possible contact between the metal case and the board is observed around the maximum visible outline, which triggers the establishment of a short circuit between the signal and case. A ground plane pattern is recommended to incorporate into the maximum visible outline and the inside of the case to enhance shielding effect.

- The maximum input voltage is $\pm 0.5 \mathrm{~V}$ when the module is not in action (no power being supplied). Potential damage to the module may be concerned if the maximum voltage is violated. If a voltage of $\pm 0.5 \mathrm{~V}$ or more is input, a protective circuit is inserted into the input terminal.
- The maximum input amplitude is $\pm 1 \mathrm{~V}$ when the module is in action. If signal amplitude of $\pm 1 \mathrm{~V}$ or more is input, a protective circuit is inserted into the input terminal.
- The series regulator type power supply is required to ensure low noise. Switching noise lies in the switching regulator type power supply such as a DC-DC converter, which impairs low noise in the module.



## DIFFERENTIAL AMPLIFIER

## Differential Amplifier



VAbsolute maximum ratings

| Supply voltage ( $\pm \mathrm{Vs}$ ) | $\pm 18 \mathrm{~V}$ |
| :--- | :--- |
| Signal input voltage | $\pm \mathrm{Vs}$ |
| Control voltage | $+5.5 \mathrm{~V},-0.5 \mathrm{~V}$ |
| $\boldsymbol{V}$ Gain | $1,2,5,10,20,50, \times 100$ |
| Gain | $\pm 0.1 \%$ (typ) |
| Accuracy |  |

VInput characteristics

| Input form | Differential input |
| :--- | :--- |
| Impedance | $3 \times 10^{10} \Omega / / 8 \mathrm{pF}$ (typ) (for differential and common input) |
| Common mode voltage | $\pm 10 \mathrm{~V}$ |
| Common mode rejection ratio | Min. 90 dB (DC to $60 \mathrm{~Hz}: \mathrm{G}=100$ ) |
| Offset voltage | $\pm 2 \mathrm{mV}$ (typ) (RTI, G = 100, input grounding) |
| Offset drift | $\pm 25 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) (RTI, G $=100$, input grounding) |
| Voltage noise density | $27 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ (typ) (RTI, G $=100$, input grounding) |

## VFrequency characteristics

| $\pm 3 \mathrm{~dB}$ flat (small signal) | Min. DC to 200 kHz |
| :--- | :--- |
| $\pm 1 \%$ flat (small signal) | DC to 50 kHz (typ) |
| Full power bandwidth | DC to 100 kHz (typ) |
| Slew rate | $20 \mathrm{~V} / \mu$ s (typ) |

## Basic connection diagram



## VOutput characteristics

| Maximum voltage | $\pm 10 \mathrm{~V}$ |
| :--- | :--- |
| Maximum current | $\pm 5 \mathrm{~mA}$ |
| Impedance | Max. $5 \mathrm{k} \Omega$ |
| Load resistance | Min. $2 \mathrm{k} \Omega$ |

-Control characteristics

| Control line |  | $1,2,5,10, \times 1, \times 10$ |
| :---: | :---: | :---: |
| Level |  | TTL or CMOS negative logic |
| Level input process |  | Pulled up to +5 V (internal) at $100 \mathrm{k} \Omega$ |
| $\nabla$ Others |  |  |
| Supply voltage |  | $\pm 15 \mathrm{~V}$ ( $\pm 14$ to 16 V ) |
| Quiescent current |  | +15mA, -12mA (typ) |
| Temperature/ humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions |  | $51.5 \times 14 \times 6.5 \mathrm{~mm}$, Type S20 |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$, $\mathrm{Vs}= \pm 15 \mathrm{~V}$

## Block diagram



## Characteristics




## Characteristics





Harmonics distortion


Common mode rejection ratio (CMRR)


Input voltage noise density


Programmable Gain Amplifier


## CA-206L2

CA-206L2 amplifier is a low noise DC amplifier capable of logical setting of gains ( $\times 1$ to $\times 100$ ) in accordance with 1 -, 2-, or 5-sequence. Frequency characteristics are expressed in a flat response till 100 kHz . A gain error is limited to $0.1 \%$ (typ) that denotes high accuracy. Gain setting is completed by controlling the 6 control terminals ( $1,2,5$, $10, \times 1, \times 10$ ) according to TTL or CMOS IC negative logic.
Easy gain setting with the use of 3-bit binary signal or binary code switch is assured if the amplifier is connected with the binary latch adapter CA-903N. The latch function enables direct connection to CPU. CA-206L2/CA-903N amplifiers are 20-pin single inline package, which enables high density mounting.

VAbsolute maximum ratings

| Supply voltage ( $\pm \mathrm{Vs}$ ) | $\pm 18 \mathrm{~V}$ |
| :--- | :--- |
| Signal input voltage | $\pm \mathrm{Vs}$ |
| Control voltage | $+5.5 \mathrm{~V},-0.5 \mathrm{~V}$ |
| Gain |  |


| Gain (G) | $\begin{aligned} & 1,2,5,10,20,50, \times 100 \\ & \text { Error: } \pm 0.1 \% \text { (typ), Max. } \pm 0.4 \%(1 \mathrm{kHz}) \end{aligned}$ |
| :---: | :---: |
| Setting | 6 control terminals ( $1,2,5,10, \times 1, \times 10$ ) used |
| VInput characteristics |  |
| Input form | Unbalanced |
| Input impedance | $1 \mathrm{M} \Omega \pm 2 \%$ (1kHz) |
| Max. input voltage (linear) | $\pm 10 \mathrm{~V}(\mathrm{G}=1)$ |
| Offset voltage | $\pm 1 \mathrm{mV}$ (typ) (RTI, G = 100, input grounding) <br> Offset voltages of the input/output amplifiers: <br> Adjustable with external trimmer potentiometer (2 pcs.). |
| Offset drift | $\pm 20 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ (typ) (RTI, G = 100, input grounding) |
| Voltage noise density | $7 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ (typ) (RTI, G = 100, input grounding) |

- Frequency characteristics

| $\pm 3 \mathrm{~dB}$ flat (small signal) | Min. DC to 500 kHz |
| :--- | :--- |
| $\pm 1 \%$ flat (small signal) | DC to 100 kHz (typ) |
| Full power bandwidth | DC to 100 kHz (typ) |
| Slew rate | $10 \mathrm{~V} / \mu \mathrm{s}$ (typ) |

Output characteristics

| Maximum output | Voltage: $\pm 10 \mathrm{~V}$, Current: $\pm 5 \mathrm{~mA}$ |
| :--- | :--- |
| Load resistance | Min. $2 \mathrm{k} \Omega$ |
| Output impedance | Max. $5 \mathrm{k} \Omega$ |

Control characteristics

| Control line | $1,2,5,10, \times 1, \times 10$ |
| :--- | :--- |
| Level | TTL or CMOS negative logic |
| Level input process | Pulled up to +5 V (internal) at $100 \mathrm{k} \Omega$ |

## $\nabla$ Others

| Supply voltage |  | $\pm 15 \mathrm{~V}( \pm 14$ to 16 V$)$ |
| :--- | :--- | :--- |
| Quiescent current |  | +15 mA (typ), -20 mA (max) |
| Temperature/ <br> humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions |  | $51.5 \times 14 \times 6.5 \mathrm{~mm}$, Type S20 |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$, $\mathrm{Vs}= \pm 15 \mathrm{~V}$

## Basic connection diagram



Establish a ground for INPUT, and obtain a gain of 100 . Adjust OFFSET1 to obtain 0V of AıOUT.
Adjust OFFSET2 to obtain 0V of OUTPUT.

## Block diagram

OFFSET1 A1 OUT OFFSET2


## Characteristics





## Characteristics



Amplitude-Phase ( $\times 50$ gain)


Distortion ( $\times 1$ gain)


Distortion ( $\times 100$ gain)


Amplitude-Phase ( $\times 20$ gain)


Amplitude-Phase ( $\times 100$ gain)


Distortion ( $\times 10$ gain)


Input voltage noise density ( $\times 100$ gain)


## Binary Latch Adapter



## CA-903N

CA-903N adapter is endowed with outstanding features including gain setting by binary code that is available if connected to CA-206L2 programmable gain amplifier or CA-406L2 differential amplifier. This adapter is also capable of actuating the latch function.
Gain setting binary code input becomes valid if the latch control input terminal is open or set at +5 V , which allows gain setting by the negative logic level signal. Latch control input needs to be set at OV to exert the latch function, which enables data on gain setting binary code input to be latched at the edge of 0 V .
Connection with CA-206L2 and CA-406L2 amplifiers is established with the samenumbered pins ( 8 pcs.), and the power is supplied.

- Absolute maximum ratings

| Supply voltage ( $\pm \mathrm{Vs}$ ) | +5.5 V |
| :--- | :--- |
| Control voltage | $+5.5 \mathrm{~V},-0.5 \mathrm{~V}$ |

VI/O characteristics (truth table)

| Gain | Input |  |  |  | Output |  |  |  |  |  | Function |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | G | C | B | A | 1 | 2 | 5 | 10 | $\times 1$ | $\times 10$ | - |
| 1 | H | H | H | H | L | H | H | H | L | H | - |
| 2 | H | H | H | L | H | L | H | H | L | H | - |
| 5 | H | H | L | H | H | H | L | H | L | H | - |
| 10 | H | H | L | L | H | H | H | L | L | H | - |
| 20 | H | L | H | H | H | L | H | H | H | L | - |
| 50 | H | L | H | L | H | H | L | H | H | L | - |
| 100 | H | L | L | H | H | H | H | L | H | L | - |
| 100 | H | L | L | L | H | H | H | L | H | L | - |
|  | L | $\times$ | $\times$ | $\times$ | Data is latched at the falling edge of G. | Latch |  |  |  |  |  |


| Level | TTL or CMOS negative logic |
| :--- | :--- |
| Level input process | Pulled up to +5 V (internal) at $100 \mathrm{k} \Omega$ |
| Latch function | A variation in control input make its presence <br> at the output if Trigger terminal 17 is open or <br> set at "Hi". If the terminal is set at "Lo", data <br> on control input at the falling edge is latched. |

## Others

| Supply voltage |  | $+5 \mathrm{~V} \pm 10 \%$ |
| :--- | :--- | :--- |
| Quiescent current |  | $150 \mu \mathrm{~A}$ (typ), 1 mA (max) |
| Temperature/ <br> humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions |  | $51.5 \times 14 \times 4.0 \mathrm{~mm}$, Type S20 |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}, \mathrm{Vs}= \pm 15 \mathrm{~V}$

## Basic connection diagram



## Control signal timing chart


td: Setup delay time: Max. 850ns

High Speed Inverting Amplifier


- Amplification characteristics

| Gain | Gained with external resistors (2 pcs.). (Rin, RNF) <br> GAIN $=\frac{R_{N F}}{R_{\mathrm{IN}}}$ |
| :--- | :--- |
| Frequency <br> characteristics | Full power: DC to 1 MHz <br> Small signal: DC to $10 \mathrm{MHz}( \pm 3 \mathrm{~dB})$ |
| Slew rate | $200 \mathrm{~V} / \mu \mathrm{s}$ (typ) |
| Input characteristics |  |
| Impedance | Rin |
| Max. input voltage |  |
| VOutput characteristics |  |


| Impedance | Max. $5 \Omega$ |
| :--- | :--- |
| Max. output voltage | $\pm 10 \mathrm{~V}$ |
| Max. output current | $\pm 10 \mathrm{~mA}$ |
| Offset voltage | $\pm 7 \mathrm{mV}$ (typ) |
| Others |  |


| Supply voltage |  | $\pm 15 \mathrm{~V}$ ( $\pm 14$ to 16V) |
| :---: | :---: | :---: |
| Quiescent current |  | $\pm 20 \mathrm{~mA}$ (typ) |
| Temperature/ | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
| humidity range | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions |  | $32 \times 13.3 \times 4.0 \mathrm{~mm}$, Type S12 |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$, $\mathrm{Vs}= \pm 15 \mathrm{~V}, \mathrm{RiN}_{\mathrm{IN}}=2 \mathrm{k} \Omega, \mathrm{R}_{\mathrm{NF}}=2 \mathrm{k} \Omega$

## Characteristics



Low Noise Amplifier


## SA-220F5 SA-230F5 SA-430F5 SA-200F3 SA-400F3

SA series amplifiers are preamplifiers for submicro-signal detection, which have been developed to assure noise reduction never before accomplished. 5 types of SA series amplifiers, which vary by a frequency band, input form, and input impedance, are available. Not only the dedicated power supply but the sensor control power supply is offered for outstanding noise reduction.
SA series amplifiers have actualized low noise featuring the following items through the adoption of our original circuit that is the application of the negative feedback technology: $50 \Omega$ of input impedance, 0.6 dB of noise figure (SA-230F5), $0.5 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ of input voltage noise at $100 \mathrm{k} \Omega$ (SA-200F3), and $200 \mathrm{fA} / \sqrt{\mathrm{Hz}}$ of input voltage noise at $1 \mathrm{M} \Omega$ (SA220F5).

| Model | SA-220F5 <br> Low noise FET amplifier | SA-230F5 <br> Low noise amplifier | SA-430F5 <br> Low noise differential amplifier |
| :---: | :---: | :---: | :---: |
| Frequency band (typ) | 300 Hz to 100 MHz | 400 Hz to 140 MHz | 400 Hz to 110 MHz |

VInput

| Input form | AC coupling, unbalanced single ended input (SMA connector) | AC coupling, unbalanced single ended input (SMA connector) | AC coupling, balanced differential input (2 SMA connectors) |
| :---: | :---: | :---: | :---: |
| Input impedance | $1 \mathrm{M} \Omega \pm 5 \%$ ( 5 kHz ) Shunt capacitance: 57 pF (typ) | $50 \Omega \pm 5 \%$ (100kHz) | Differential input: $50 \Omega \pm 5 \%$ ( 100 kHz ) Common mode input: $530 \Omega$ typ ( 100 kHz ) |
| Maximum input voltage (burnout voltage) | $\pm 1.0 \mathrm{~V}$ | $\pm 1.0 \mathrm{~V}$ | $\pm 2.0 \mathrm{~V}$ (differential input/common input) |
| CMRR (RTI) | $\frac{-}{-}$ | - | Min. 80dB (100kHz) 90 dB typ $(100 \mathrm{kHz}), 80 \mathrm{~dB}$ typ $(10 \mathrm{MHz})$ |
| Input voltage noise density (short-circuit in input terminal) | Max. $0.7 \mathrm{nV} / \sqrt{\mathrm{Hz}}(100 \mathrm{kHz})$ $0.5 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ typ ( 10 k to 1 MHz ) | Max. 0.35nV// Hz (100kHz) $0.25 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ typ ( 10 k to 1 MHz ) | Max. 0.45 nV WHz ( 100 kHz ) $0.35 \mathrm{nV} / \sqrt{\mathrm{Hz}}$ typ ( 10 k to 1 MHz ) |
| Input noise current density | $200 f \mathrm{~A} / \sqrt{\mathrm{Hz}}$ typ (100kHz) | $5.0 \mathrm{pA} / \sqrt{\mathrm{Hz}}$ typ (100kHz) | $7.0 \mathrm{pA} / \sqrt{\mathrm{Hz}}$ typ (100kHz) |
| Noise figure (50 2 ) | - | Max. 0.7dB, 0.6dB typ (10MHz) Max. 1.0dB, 0.8 dB typ ( 100 MHz ) | Max. 1.25dB, 1.10 dB typ (10MHz) Max. 1.75dB, 1.40dB typ (100MHz) |

## VOutput characteristics

| Maximum output voltage | $2 \mathrm{Vp}-\mathrm{p}(1 \mathrm{kHz}$ to 20 MHz , load resistance: $50 \Omega$ ) | $2 \mathrm{Vp}-\mathrm{p}(1 \mathrm{kHz}$ to 20 MHz , load resistance: $50 \Omega$ ) | $2 \mathrm{Vp}-\mathrm{p}$ ( 1 kHz to 20 MHz , load resistance: $50 \Omega$ ) |
| :---: | :---: | ---: | :---: |

## - Amplifier

| Voltage gain | $46 \pm 0.5 \mathrm{~dB}$ ( 1 MHz , load resistance: $50 \Omega$ ) | $46 \pm 0.5 \mathrm{~dB}$ (1MHz, load resistance: $50 \Omega$ ) | $46 \pm 0.5 \mathrm{~dB}$ ( 1 MHz , load resistance: $50 \Omega$ ) |
| :---: | :---: | :---: | :---: |
| Voltage gain frequency characteristics | 1 kHz to $80 \mathrm{MHz}+0.5$, Max. -3 dB 300 Hz to $100 \mathrm{MHz}+0.5,-3 \mathrm{~dB}$ typ | 1 kHz to $100 \mathrm{MHz}+0.5$, Max. -3 dB 400 Hz to $140 \mathrm{MHz}+0.5,-3 \mathrm{~dB}$ typ | 1 kHz to $100 \mathrm{MHz}+0.5$, Max. -3 dB 400 Hz to $110 \mathrm{MHz}+0.5,-3 \mathrm{~dB}$ typ |
| Intercept point | - | +30dBm typ ( 68 MHz ) | +28dBm typ (68MHz) |
| $\nabla$ Power supply |  |  |  |
| Recommended power supply voltage range | $\pm 15 \mathrm{~V} \pm 5 \%$ | +15V $\pm 5 \%$ | $\pm 15 \mathrm{~V} \pm 5 \%$ |
| Quiescent current (no signal) | Max. +65mA typ +75mA <br> Max. -10 mA typ -15 mA | Max. +55mA | Max. +55mA typ +65mA <br> Max. $-30 m A$ typ $-45 m A$ |

VEnvironment

| Specified temperature range | $23^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}$ | $23^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}$ | $23^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}$ |
| :--- | :---: | :---: | :---: |
| Storage temperature/ | $-10^{\circ} \mathrm{C}$ to $50^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ | $-10^{\circ} \mathrm{C}$ to $50^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ | $-10^{\circ} \mathrm{C}$ to $50^{\circ} \mathrm{C}, 10$ to $80^{\circ} \% \mathrm{RH}$ |
| humidity range | (no condensation) | (no condensation) | (no condensation) |

## VDimensions

| Dimensions | $68 \times 43 \times 28 \mathrm{~mm}$ (protrusion not included) | $68 \times 43 \times 17.6 \mathrm{~mm}$ (protrusion not included) | $68 \times 43 \times 28 \mathrm{~mm}$ (protrusion not included) |
| :--- | :---: | :---: | :---: |
| Weight (NET) | Approx. 130 g | Approx. 90 g | Approx. 130 g |

Note: Power supply: SA-915D1



SA-230F5: Noise figure


## Main applications

SA series amplifiers are used to foster versatility as sensor head amplifiers or preamplifiers for sensitivity improvement and noise reduction in analyzers and measurement instruments.
-"MCT <Mercury Cadmium Tellurium> sensor" for infrared detection
-"Superconducting SQUID sensor" for micro-magnet detection
-"High-temperature superconducting Josephson device" for microwave detection
-"Electromagnetic sensor" for MRI systems
-Photodetector such as a photomultiplier and phototransistor

| SA-200F3 <br> Low noise amplifier | SA-400F3 <br> Low noise differential amplifier |
| :---: | :---: |
| DC to 800 kHz | DC to 700 kHz |
| $\nabla$ Input |  |
| DC coupling, unbalanced single wire grounded input (SMA connector) | DC coupling, balanced differential input (2 SMA connectors) |
| Selectable among $1 \mathrm{k} / 10 \mathrm{k} / 100 \mathrm{k} \Omega \pm 5 \%$ (DC), Shunt capacitance: Max. 150pF | Selectable among 1k/ 10k/ 100k $\Omega \pm 5 \%$ (DC), Shunt capacitance: Max. 80pF |
| $\pm 0.5 \mathrm{~V}$ | Differential input: $\pm 0.5 \mathrm{~V}$ Common input: $\pm 10 \mathrm{~V}$ |
| $\frac{-}{}$ | $\begin{gathered} \text { Min. 110dB (50Hz) } \\ 120 \mathrm{~dB} \text { typ }(50 \mathrm{~Hz}), 80 \mathrm{~dB} \operatorname{typ}(100 \mathrm{kHz}) \\ \hline \end{gathered}$ |
| Max. 0.7nV/ $\sqrt{\mathrm{Hz}}(1 \mathrm{kHz}$ ) | Max. $0.9 \mathrm{nV} / \mathrm{h} / \mathrm{Hz}(1 \mathrm{kHz})$ |
| Max. $0.5 \mathrm{nV} / \sqrt{\mathrm{Hz}}(1 \mathrm{kHz})$ | Max. $0.75 \mathrm{nV} / 2 / \mathrm{Hz}(1 \mathrm{kHz})$ |
| $2.2 \mathrm{pA} / \sqrt{\mathrm{Hz}}(10 \mathrm{kHz})$ | $3.0 \mathrm{pA} / \sqrt{\mathrm{Hz}}(10 \mathrm{kHz})$ |
| - | - |
| VOutput characteristics |  |
| $\pm 10 \mathrm{~V}$ (1kHz, load resistance $\geq 1 \mathrm{k} \Omega$ ) | $\pm 10 \mathrm{~V}$ ( 1 kHz , load resistance $\geq 1 \mathrm{k} \Omega$ ) |
| $50 \Omega \pm 5 \%$ (DC) | $50 \Omega \pm 5 \%$ (DC) |

## - Amplifier

| $40 \pm 0.5 \mathrm{~dB}(1 \mathrm{kHz})$ | $40 \pm 0.5 \mathrm{~dB}(1 \mathrm{kHz})$ |
| :---: | :---: |
| DC to $800 \mathrm{kHz}+0.5,-3 \mathrm{~dB}$ typ | DC to $700 \mathrm{kHz}+0.5,-3 \mathrm{~dB}$ typ |
| - | - |

PPower supply

| $\pm 15 \mathrm{~V} \pm 5 \%$ | $\pm 15 \mathrm{~V} \pm 5 \%$ |
| :---: | :---: |
| $\pm 50 \mathrm{~mA}$ | $\pm 92 \mathrm{~mA}$ typ $\pm 100 \mathrm{~mA}$ |

## VEnvironment

| $23^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}$ | $23^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}$ |
| :---: | :---: |
| $-10^{\circ} \mathrm{C}$ to $50^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ (no condensation) | $-10^{\circ} \mathrm{C}$ to $50^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ (no condensation) |

## $\nabla$ Dimensions

| $68 \times 43 \times 17.6 \mathrm{~mm}$ (protrusion not included) | $68 \times 67 \times 28 \mathrm{~mm}$ (protrusion not included) |
| :---: | :---: |
| Weight: Approx. 90 g | Approx. 180 g (heat sink included) |

SA-230F5: Transient response (rise)

${ }^{[\mathrm{nv} \sqrt{\mathrm{Hz}}]} \mathrm{SA}$-200F3: Input voltage noise density




## DC power supply: SA-915D1



SA-915D1 power supply is to supply DC power, which is intended for SA series amplifiers, for reductions in noise and ripple. The innovative way to fight the noise has been taken in this power supply. The combination use of a SA series amplifier and SA915D1 power supply is suggested to assure outstanding performance.

## VOutput

| Output form | Mini DIN, 4-pin connector |
| :--- | :--- |
| Output voltage | $\pm 15 \mathrm{~V} \pm 3 \%$ |
| Maximum output current | $\pm 100 \mathrm{~mA}$ |
| Output voltage noise/ripple | Max. $300 \mu \mathrm{Vrms}(\mathrm{BW}: 10 \mathrm{~Hz}$ to 20 MHz ) |
| Output voltage <br> temperature coefficient | $500 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ typ |

## $\nabla$ Others

| Power supply | $\mathrm{AC} 100 \mathrm{~V} \pm 10 \%, 48 \mathrm{~Hz}$ to 62 Hz Approx. 10VA |
| :--- | :--- |
| Dimensions | $120 \times 55 \times 200 \mathrm{~mm}$ (protrusion not included) |
| Weight (NET) | Approx. 1.4 kg |
| Operating temperature/ <br> humidity range | $0^{\circ} \mathrm{C}$ to $40^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ <br> (no condensation) |
| Storage temperature $/$ <br> humidity range | $-0^{\circ} \mathrm{C}$ to $50^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ <br> (no condensation) |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}, \mathrm{AC} 100 \mathrm{~V}$, Load resistance: $150 \Omega$

## DC bias supply: SA-912S1



SA-912S1 power supply is a bias power supply for sensors that process micro-signals.
This power supply is composed of a dual-redundant regulator, special noise filter circuit, dual transformers, and dual shield chassis, which offers excellent noise reduction.

## VOutput

Output form
Mini DIN, 4-pin connector
Output voltage $\pm 12 \mathrm{~V} \pm 3 \%$ (no load)
Maximum output current $\pm 100 \mathrm{~mA}$

| Output voltage noise/ripple | Max. $3 \mu \mathrm{Vrms}$ (BW: 10 Hz to 1 MHz ) |
| :--- | :--- |

Output voltage
$300 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ typ
temperature coefficient

## $\nabla$ Others

| Power supply | $\mathrm{AC} 100 \mathrm{~V} \pm 10 \%, 48 \mathrm{~Hz}$ to 62 Hz Approx. 5 VA |
| :--- | :--- |
| Dimensions | $120 \times 55 \times 200 \mathrm{~mm}$ (protrusion not included) |
| Weight (NET) | Approx. 1.4 kg |
| Operating temperature/ | $0^{\circ} \mathrm{C}$ to $40^{\circ} \mathrm{C}, 10$ to $90 \% \mathrm{RH}$ |
| humidity range | (no condensation) |
| Storage temperature $-10^{\circ} \mathrm{C}$ to $50^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ <br> humidity range: (no condensation) |  |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$, AC100V, Load resistance: $70 \Omega$

## RESISTOR TUNABLE OSCILLATOR

Resistor Tunable Oscillator

## CG-402R1/2



CG-402R series oscillators have achieved reductions in price and size through the simplification of the circuit. Frequencies are allocated with the external resistors (2 pcs.), and CG-402R series oscillators are in 12-pin single-inline package that enhances mounting density.

| Model |  | CG-402R1 | CG-402R2 |
| :---: | :---: | :---: | :---: |
| Frequency range ${ }^{41}$ |  | 20 Hz to 20kHz | 1 kHz to 100 kHz |
| Frequency setting |  | Specified with external resistors (2 pcs.). |  |
| Frequency accuracy ${ }^{\text {+2 }}$ |  | $\pm 5 \%, \pm 2 \%$ (typ) |  |
| Frequency stability |  | $\pm 50 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ (typ) |  |
| Output voltage |  | $2.5 \mathrm{Vrms} \pm 5 \%^{* 3}$ |  |
| Output voltage stability |  | $100 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ (typ) |  |
| Output impedance |  | Max. $5 \Omega$ |  |
| Load impedance |  | Min. 2k $\Omega$ |  |
| Distortion |  | $\begin{aligned} & \text { Max. 0.1\% } \\ & (200 \mathrm{~Hz} \text { to } 10 \mathrm{kHz}) \end{aligned}$ |  |
| Supply voltage |  | $\pm 15 \mathrm{~V} \pm 10 \%$ |  |
| Quiescent current (typ) |  | $\pm 8 \mathrm{~mA}$ | 12 mA |
| Temperature/ humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |  |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |  |
| Dimensions |  |  |  |

## Basic connection diagram



Output voltage
Pins (10) - (11) shorted: $\pm 10 \mathrm{~V}$
Pins (11) - (12) shorted: $\pm 2 \mathrm{~V}$

Note: The following specifications are applied unless otherwise specified:
$23 \pm 5^{\circ} \mathrm{C}, \pm 15 \mathrm{~V}, \mathrm{RF}=15.9 \mathrm{k} \Omega$
*1. Expansion of the lower frequency is enabled.
*2. Errors of external resistors are excluded.
*3. Available at $\pm 2$ to $\pm 10 \mathrm{~V}$. Max. output: $\pm 10 \mathrm{~V}, \pm 5 \mathrm{~mA}$
(402R1: 20 Hz to $10 \mathrm{kHz}, 402 \mathrm{R} 2: 1 \mathrm{kHz}$ to 50 kHz )

## Frequency setting

Equation of external resistor

$$
\begin{array}{ll}
\text { CG-402R1 } & \mathrm{R}_{\mathrm{F}}=\frac{15.9 \times 10^{3}}{\mathrm{fo}}(\mathrm{k} \Omega) \\
\text { CG-402R2 } & \mathrm{R}_{\mathrm{F}}=\frac{159 \times 10^{3}}{\mathrm{fo}}(\mathrm{k} \Omega)
\end{array}
$$

Note: fo: Oscillation frequency
Units: fo in Hz

Frequency setting requires 2 external resistors of the same resistance.
Be sure to use resistors with relative tolerance of $1 \%$ to ensure optimal internal operation.

## Block diagram



Expansion of the lower frequency range (10Hz to 20Hz)

Equation of external resistor

$$
\text { CG-402R1 } \quad \mathrm{RF}=\frac{15.9 \times 10^{3}}{\text { fo }}(\mathrm{k} \Omega)
$$

Note: fo: Oscillation frequency
Units: fo in Hz


## Output voltage setting

2.5 Vrms of output voltage is obtained if Pins (10) and (11) are open, but output voltage varies as follows: $\pm 10 \mathrm{~V}$ (approx. 7 Vrms ) when Pins (10-(11) are shorted and $\pm 2 \mathrm{~V}$ (approx. 1.4 Vrms ) when Pins (11)-(12) are shorted. An external resistor is required for setting output voltage ( 1.4 to 7 Vrms ) other than the above. The graph at the right expresses the standard values of external resistor and output voltage. Adjustment with a variable resistor is required to derive correct voltage.


## When load capacity is large

Potential unstable and abnormal oscillation may be concerned if 100 pF or more of load capacity is observed. With a coaxial cable or shielding wire put under load, overload capacity is detected in some oscillators a load capacity if the cable or

wire reaches a length in excess of 50 cm . In the event of the above, a $50 \Omega$-resistor or buffer amplifier needs to be inserted between the relevant oscillator and load.


## Characteristics



Harmonics distortion - Oscillation frequency


Oscillation frequency accuracy (CG-402R2)



## RESISTOR TUNABLE OSCILLATOR

Resistor Tunable Oscillator

## CG-202R3

CG-202R3 oscillator is a sine-wave oscillator capable of allocating oscillation frequency in the range of 100 kHz to 1 MHz . This oscillator not only possesses superior frequency accuracy and output voltage accuracy but also assures stable sine-wave signals. CG202R3 oscillator is in 24-pin dual-inline package, which enables a low-pass expansion up to 10 Hz with the external capacitors and capacitor.

| Frequency range** |  | 100 kHz to 1 MHz |
| :---: | :---: | :---: |
| Frequency setting |  | Specified with external resistors (2 pcs.). |
| Frequency accuracy ${ }^{*}$ |  | Max. $\pm 5 \%, \pm 2 \%$ (typ) |
| Frequency stability |  | $\pm 50 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ (typ) |
| Output voltage ${ }^{* 3}$ |  | $2.5 \mathrm{Vrms} \pm 3 \%$ |
| Output voltage stability |  | 50ppm/ ${ }^{\circ} \mathrm{C}$ (typ) |
| Output impedance |  | 50תtyp |
| Load impedance |  | Min. 2k , Max. 100pF |
| Harmonics level |  | $\begin{aligned} & \hline-50 \mathrm{~dB}(\operatorname{typ})(1 \mathrm{MHz}) \\ & -60 \mathrm{~dB}(\operatorname{typ})(100 \mathrm{kHz}) \end{aligned}$ |
| Supply voltage |  | $\pm 15 \mathrm{~V} \pm 10 \%$ |
| Quiescent current (typ) |  | $+30 \mathrm{~mA} /-20 \mathrm{~mA}$ |
| Temperature/ humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions |  | $34.5 \times 18.7 \times 7.9 \mathrm{~mm}$, Type KB |

Note: The following specifications are applied unless otherwise specified:
$23 \pm 5^{\circ} \mathrm{C}, \pm 15 \mathrm{~V}, \mathrm{RF}=15.9 \mathrm{k} \Omega$
*1. Expansion of the lower frequency is enabled.
*2. Errors of external resistors are excluded.
Oscillators


## Basic connection diagram



## Block diagram



Output voltage setting


|  |  | Freque | ge ( |  |  | C1 <br> (3)-20 | $\mathrm{C} 2$ | $\begin{gathered} \mathrm{C} 3 \\ (22-23 \end{gathered}$ | Cf <br> (13-(16) (18-21) | $\underset{(16-18)}{R_{F}}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 10 | 100 | 1k | 10k | 100k | 1M |  |  |  |  |  |
|  |  |  |  |  |  |  |  |  |  | $1.59 \times 10^{6}$ |
|  |  |  |  |  |  | - | - | - | - | fo (k ${ }^{\text {a }}$ |
|  |  |  |  |  |  |  |  |  |  | $1.59 \times 10^{6}$ |
|  |  |  |  |  |  |  | $0.2 \mu \mathrm{~F}$ | - | - | fo (k) |
|  |  |  |  |  |  |  |  |  |  | $1.59 \times 10^{6}$ |
|  |  |  |  |  |  | $0.47 \mu \mathrm{~F}$ | $2 \mu \mathrm{~F}$ | - | - | fo (k ${ }^{\text {f }}$ |
|  |  |  |  |  |  |  |  |  |  | $\underline{1.59 \times 10^{5}}$ |
|  |  |  |  |  |  | $4.7 \mu \mathrm{~F}$ | $20 \mu \mathrm{~F}$ | - | 900pF | fo (k 2 ) |
|  |  |  |  |  |  |  |  |  |  | $\underline{1.59 \times 10^{4}}$ |
|  |  |  |  |  |  | $20 \mu \mathrm{~F}$ | $47 \mu \mathrm{~F}$ | $1 \mu$ | 9900pF | fo (k $)^{\text {) }}$ |

## Characteristics

Output voltage - Frequency
100 kHz standard $\mathrm{RI}=1 \mathrm{k} \Omega$


Output voltage - Temperature $\mathrm{fo}=100 \mathrm{kHz}$


Output voltage - Temperature $\quad \mathrm{fo}=1 \mathrm{MHz}$


Oscillation frequency - Temperature


## Harmonics leve



Oscillation frequency - Temperature

$$
\mathrm{fo}=1 \mathrm{MHz}
$$



Oscillation frequency accuracy 100 kHz standard: RF: Calculated value


## CG-102R1/2 CG-302R1/2



| Model |  | CG-102R1 | CG-302R1 | CG-102R2 | CG-302R2 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Frequency range*1 |  | 20 Hz to 20kHz |  | 1 kHz to 100 kHz |  |
| Frequency setting |  | Specified with external resistors (2 pcs.). |  |  |  |
| Frequency accuracy* ${ }^{\text {2 }}$ |  | Max. $\pm 2 \%, \pm 0.5 \%$ (typ) |  |  |  |
| Frequency stability |  | $\pm 15 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ (typ) |  | $\pm 25 \mathrm{ppm} /{ }^{\circ} \mathrm{C}$ (typ) |  |
| Output voltage |  | $2.5 \mathrm{Vrms} \pm 0.5 \%{ }^{* 3}$ |  |  |  |
| Output voltage stability |  | 50ppm/ ${ }^{\circ} \mathrm{C}$ (typ) |  |  |  |
| Output impedance |  | Max. $5 \Omega$ |  |  |  |
| Load impedance |  | Min. 2ת, Max. 100pF |  |  |  |
| Distortion |  | $\begin{aligned} & \text { Max. 0.005\% } \\ & (70 \mathrm{~Hz} \text { to } 10 \mathrm{kHz}) \end{aligned}$ |  | $\begin{gathered} \hline \text { Max. } 0.005 \% \\ (2 \mathrm{kHz} \text { to } 50 \mathrm{kHz}) \\ \text { Max. } 0.01 \% \\ (50 \mathrm{kHz} \text { to } 100 \mathrm{kHz}) \\ \hline \end{gathered}$ |  |
| Supply voltage |  | $\pm 15 \mathrm{~V} \pm 10 \%$ |  |  |  |
| Quiescent current (typ) |  | $\begin{aligned} & +13 \mathrm{~mA} / \\ & -23 \mathrm{~mA} \end{aligned}$ | $\pm 13 \mathrm{~mA}$ | $\begin{aligned} & +28 \mathrm{~mA} / \\ & -38 \mathrm{~mA} \end{aligned}$ | $\pm 28 \mathrm{~mA}$ |
| Temperature/ humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |  |  |  |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |  |  |  |
| Dimensions |  | 34.5x18.777.9mm | 51.5×14.0x5.5mm | 34.5x18.777.9mm | 51.5x14.0x5.5mm |
|  |  | Type KB | Type S20 | Type KB | Type S20 |
|  |  | 24pin DIP | 20pin SIP | 24pin DIP | 20pin SIP |

Note: The following specifications are applied unless otherwise specified:
$23 \pm 5^{\circ} \mathrm{C}, \pm 15 \mathrm{~V}, \mathrm{RF}=15.9 \mathrm{k} \Omega$
*1. Expansion of the lower frequency is enabled.
*2. Errors of external resistors are excluded.
*3. Available at 0.5 to 20 Vp -p. Max. output: $\pm 10 \mathrm{~V}, \pm 5 \mathrm{~mA}$
(CG-102R1/302R1: 20 Hz to $10 \mathrm{kHz}, \mathrm{CG}-102 \mathrm{R} 2 / 302 \mathrm{R} 2: 1 \mathrm{kHz}$ to 50 kHz )

## Basic connection diagram




## Block diagram




Equation of external resistor

| CG-102R1 | $\mathrm{RF}_{\mathrm{F}}=\frac{15.9 \times 10^{3}}{\text { fo }}(\mathrm{k} \Omega)$ |
| :--- | :--- |
| CG-302R1 |  |
| CG-102R2 | $\mathrm{RF}_{\mathrm{F}}=\frac{159 \times 10^{3}}{\text { fo }}(\mathrm{k} \Omega)$ |
| CG-302R2 | Note: fo: Oscillation frequency <br> Units: fo in Hz |

$$
\mathrm{RF}_{\mathrm{F}}=\frac{15.9 \times 10^{3}}{\mathrm{fo}_{\mathrm{o}}}(\mathrm{k} \Omega)
$$

$$
\mathrm{RF}_{\mathrm{F}}=\frac{159 \times 10^{3}}{\mathrm{fo}}(\mathrm{k} \Omega)
$$

$$
\text { Units: fo in } \mathrm{Hz}
$$

Frequency setting requires 2 external resistors of the same resistance.
An accuracy between the external resistors causes fluctuations in output level.
E.g.: Max. $\pm 0.5 \%$ of difference between outputs 1 and 2 if a resistor with tolerance of $1 \%$ is used

## Output voltage adjustment

CG-102 series oscillators are designed to obtain 20Vp-p of output voltage if Pins (12) and (14) are shorted, and CG-302 series oscillators are designed to obtain the same voltage if Pins (10) and (11) are shorted. 2.5 V rms of output voltage is obtained if CG-102/302 oscillators are open.

- To set the voltage at 1.5 Vrms or less




## $\square$ When load capacity is large

Potential unstable and abnormal oscillation may be concerned if 100 pF or more of load capacity is observed. With a coaxial cable or shielding wire put under load, overload capacity is detected in some oscillators a load capacity if the cable or


Note: Output voltage: Max. 2.5Vrms if 10kHz or more is allocated to CG-102R1/302R1 and 50 kHz or more is allocated to CG-102R2/302R2.


An external resistor is required for setting output voltage other than the above. The graphs as shown below express the standard values of external resistor and output voltage. Adjustment with a variable resistor is required to derive correct voltage.

- To set the voltage at 1.5 V rms or more


## Characteristics <br> Characteristics

wire reaches a length in excess of 50 cm . In the event of the above, a $50 \Omega$-resistor or buffer amplifier needs to be inserted between the relevant oscillator and load.



Output voltage deviation - Oscillation frequency


## RESISTOR TUNABLE OSCILLATOR

## Expansion of the lower frequency range



Type $1 \quad \mathrm{R}_{\mathrm{F}}=\frac{159}{(\mathrm{CF}+0.01) \times \mathrm{fo}}[\mathrm{k} \Omega]$
Type $2 \quad \mathrm{RF}_{\mathrm{F}}=\frac{159}{\left(\mathrm{C}_{\mathrm{F}}+0.001\right) \times \mathrm{fo}}[\mathrm{k} \Omega]$
$\mathrm{C}_{\mathrm{F}}:[\mu \mathrm{F}]$, fo : $[\mathrm{Hz}]$
Note: $1 \mathrm{~Hz} \leq$ fo $\leq 1 \mathrm{kHz}$
$800 \Omega \leq \mathrm{RF} \leq 800 \mathrm{k} \Omega$

## ■Sync oscillation

CG-102/302 series oscillators are capable of bringing external synchronization signals into sync with oscillation output produced by the oscillators. The synchronization bandwidth is approx. 1.5\%/Vrms.
The frequency of external synchronization signals is to be determined with great accuracy in advance, and then the oscillation frequency is allocated to agree with the frequency of external synchronization signal. Synchronization between the output frequency and external signal frequency is developed if an external signal ( 1 V to 5 Vrms ) is added to Pin 1.
The graph at the right represents the phase difference between the external signal and oscillation output, besides the external signal level and synchronization band.
Note that potential change in oscillator level and increase in distortion may be concerned depending on the conditions of synchronization.


Ratio between sync input signal frequency and oscillation frequency without sync input

## Oscillator Adapter



## OP-102

OP-102 adapter can be used as a sine-wave oscillator that is capable of setting the frequency by a 3-digit BCD input, in combination with DT-212D series filters.
Performance linked with oscillation frequency (setting method, setting accuracy, temperature coefficient) that DT-212D series filters offer is applied. Performance related to output voltage (accuracy, stability, temperature coefficient), however, is determined by OP-102 adapter. The output voltage has been trimmed to $2.5 \mathrm{Vrms} \pm 0.5 \%$ internally but can be set between 0.5 Vrms and 20 Vp -p with the use of an external resistor.
The oscillation frequency range is 1 Hz to 100 kHz . An external capacitor is required if 100 Hz or less of frequency is obtained.
OP-102 adapter is powered by $\pm 15 \mathrm{~V}$ and a 20 -pin single-inline package in dimensions of $51.5 \times 14.0 \times 4.0 \mathrm{~mm}$.

## - Absolute maximum ratings

| Supply voltage ( $\pm$ Vs) | $\pm 18 \mathrm{~V}$ |
| :--- | :--- |
| Signal input (Pins $(13$ and $(15)$ | $\pm \mathrm{Vs}$ |

## FOutput characteristics

| Output voltage | 2.5 Vrms | $20 \mathrm{Vp}-\mathrm{p}$ |
| :--- | :--- | :---: |
| Accuracy | Max. $\pm 0.5 \%$ |  |
| $0.6 \%$ (typ) |  |  |
| Output voltage range | 500 mVrms to $2.5 \mathrm{Vrms} \leq 100 \mathrm{kHz}$ |  |
|  | 500 mVrms to $20 \mathrm{Vp}-\mathrm{p} \quad \leq 50 \mathrm{kHz}$ <br> Short in a specified pin (20Vp-p) <br> Set with an external resistor. |  |
| Output resistance | Max. $5 \Omega$ (DT-212D) |  |
| Distortion | $0.01 \%$ (typ) |  |
| $90^{\circ}$ output | Output with $90^{\circ}$-phase lag at frequency <br> same as the main output |  |

VOscillation frequency characteristics (DT-212D)

| Frequency range | 1 Hz to 100 kHz <br> An external capacitor required if frequency <br> is 100 Hz or less |
| :--- | :--- |
| Frequency accuracy | $\pm 0.1 \%$ (typ) |
| Frequency setting | $\mathrm{BCD}: 3$ digits |
|  |  |
| Vthers $\pm 15 \mathrm{~V} \pm 10 \%$ <br> Supply voltage $+15 \mathrm{~mA},-25 \mathrm{~mA}$ <br> Quiescent current Operation <br> $-20^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$  <br> Temperature/ humidity range | Storage |
| Dimensions | $-30^{\circ} \mathrm{C}$ to $80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |

Note: The following specifications are applied unless otherwise specified:
$23 \pm 5^{\circ} \mathrm{C}, \mathrm{Vs}= \pm 15 \mathrm{~V}, 1 \mathrm{kHz}, 2.5 \mathrm{Vrms}$

## Basic connection diagram

The following diagram represents the basic connection of oscillation at 2.5 Vrms with this module connected to the DT-212DC2 filter.

The setting of oscillation frequency requires the utilization of digital signal input from the DT-212DC2 filter. Input: TTL/CMOS compatible



## OSCILLATOR ADAPTER

## ■ Expansion of the lower oscillation frequency

Expansion of the lower oscillation frequency is enabled with the external capacitors СЕхт (2 pcs.) connected to the DT-212D filter as shown in Figure 3.
Cext is derived from the following equation.
Cext $=\frac{5 \times 10^{4}}{\text { fo }}[\mathrm{pF}]$
fo: Oscillation frequency [Hz] when set at 001

The oscillation frequency range and Сехт are listed below.

|  | Set resolution | Cext $^{* 1}$ |
| :---: | :---: | :---: |
| 100 to 100 kHz | 100 Hz | 500 pF |
| 10 to 15.99 kHz | 10 Hz | 5000 pF |
| $1^{* 1}$ to 1.599 kHz | 1 Hz | 50000 pF |

*1. Be sure to use an external loop filter to ensure 1 Hz to 10 Hz of oscillation if the adapter is used in the 1 Hz to 1.599 kHz range.
*2. The DT-212DC1 filter is pre-assigned with 50000pF, and the DT-212DC2 filter is pre-assigned with 500pF


Connected with Cext

## Distortion improvement with external Cc

Distortion is improved by establishing the connection between the external capacitor Cc and OP-102 adapter as shown in Figure 4. No effects are considered if the external capacitors Сехт are switched to use.


Connected with Cc



Distortion - Frequency characteristics

## ■Distortion improvement with external loop filter

Deterioration in distortion is observed upon expansion of the lower oscillation frequency, but the connection with an external component contributes to improvement in distortion.
The oscillation frequency range and circuit example are provided below.

## 1. 10 Hz to 15.99 kHz



Connected with loop filter

2. 1 Hz to 1.599 kHz


Connected with loop filter


Distortion - Frequency characteristics

## Expansion of the lower oscillation frequency

The upper limit of oscillation frequency for 2.5 Vrms of output is 100 kHz , and for $20 \mathrm{Vp}-\mathrm{p}$ is 50 kHz . Oscillation up to 159.9 kHz for 2.5 Vrms and 100 kHz for 20 Vp -p is enabled through the connection with the protective circuit embedded in the OP102 adapter.
The protective circuit goes into action when the connection between Pin (11) (DT-212D) and Pin (11) (OP-102) is established. The above, however, results in deterioration in distortion regardless of the range.


## Output voltage setting

The setting of output voltage for the OP-102 adapter requires
Pins (17) to (20). The following procedure should be used to vary output voltage.

## 1. 20Vp-p

The OP-102 adapter is outfitted with a trimmed resistor. Connect Pins (16) to (19).


## 2. 20Vp-p to 500 mVrms

Connect an external resistor as shown below.
The output voltage is derived from the following equation: $R[k \Omega]=1111 / \mathrm{Vo}$ Vo: Output voltage [Vrms]

The standard values are provided above. Adjustment through the partial replacement of the resistor with a trimmer potentiometer is required to derive correct voltage.


Output voltage adjustment

## Sync oscillation

Sync oscillation by external signals is to be ensured if Rsync is added as shown below.
The frequency range that allows synchronization varies with input voltages.
The most stable synchronization is maintained at $90^{\circ}$ in the $0-$ to $180^{\circ}$ - range of synchronization.
If a synchronization input voltage remains the same, the synchronization range can be changed by changing Rsync. Duplation of the input voltage is equivalent to a reduction of Rsync by half.
The following represents the standard input voltage of I/O phase difference to a frequency ratio, as parameter.


Connection of synchronization input

fin : Sync input signal frequency
fo: Oscillation frequency without sync input

## Amplitude and phase difference of 2-phase output

The OP-102 adapter is rated to deliver 2 types of output as follows: main output (Pin (15) and $-90^{\circ}$ (Pin (13). These outputs are equal in oscillation frequency but have slight errors in the output voltage and phase difference.
Examples of errors in amplitude and phase difference at main output and $-90^{\circ}$ output are provided below.


Amplitude


Response for oscillation frequency setting

The output response to changes in the oscillation frequency setting is phase-continuous, which causes 300ns typ-delay.


VER : $500 \mathrm{mV} / \mathrm{div}$
HOR : $10 \mathrm{~ms} / \mathrm{div}$

## RANDOM BINARY GENERATOR

Random Binary Generator


## CG-742N

CG-742N generator is a noise generator that produces false random binary signals with high stability.
The original oscillation frequency setting is completed with the external resistor or external clock, and a frequency demultiplier is embedded in the generator to facilitate the noise bandwidth setting.
The generator is allocated with long periodic noise source through pseudo random $M$ series with the use of a 42-stage shift resistor. The CG-742N generator assures the output falling into TTL level and $\pm 5 \mathrm{~V}$ for analog process. The initialization of pulse trains to be output is enabled with the use of the reset terminal, which can be applicable to reproducibility.
The filtering of outputs delivered by this generator contributes to the acquisition of power spectrum characteristics up to 100 kHz .

| Noise source |  | Pseudo random M series with a 42stage shift resistor <br> Cycle $=\frac{\text { Approx. } 4.398 \times 10^{12}}{\text { fo }}[\mathrm{S}]$ <br> fo: Clock frequency [Hz] <br> 10.18 day at 5 MHz of clock frequency <br> Spectrum intervals <br> $1.136 \mu \mathrm{~Hz}$ at 5 MHz of clock frequency |
| :---: | :---: | :---: |
| Original oscillation frequency |  | Allocated with the external resistor or TTL-level external clock. |
| Original oscillation frequency range |  | 0.5 M to 5 MHz (with external resistor) Max. 5MHz (with external clock) |
| Frequency demultiplier (bypass enabled) |  | $1 / 1,1 / 10,1 / 100,1 / 1000$ <br> Set with the logic signals (TTL level). Latch function assigned |
| Output |  | Random binary output <br> TTL level <br> LSTTL (1 pc.) actuated $\pm 5 \mathrm{~V}$ (no load) <br> Output impedance: Approx. $100 \Omega$ <br> Load resistance: Min. $5 \mathrm{k} \Omega$ <br> (Max. 1mA) <br> Rise/fall time: Max. 200ns |
| Power supply |  | $\pm 15 \mathrm{~V}$ ( $\pm 11$ to $\pm 16 \mathrm{~V}$ ) |
| Maximum input voltage |  | (2)(12)(17)(1)22)23) $+5.5 \mathrm{~V},-0.5$ |
| Dimensions |  | $54.4 \times 33.7 \times 9.4 \mathrm{~mm}$, Type HA |
| Temperature/ humidity range | Operation | $-20^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}, 10$ to $95 \% \mathrm{RH}$ |
|  | Storage | $-30^{\circ} \mathrm{C}$ to $+80^{\circ} \mathrm{C}, 10$ to $80 \% \mathrm{RH}$ |

## Basic connection diagram 1 (Random binary)



## Block diagram




## RANDOM BINARY GENERATOR

## Usage

## 1. Oscillator

Either the built-in oscillator or external clock is available for M
series drive.
a. When a built-in oscillator is used

$\mathrm{fg}_{\mathrm{gen}}=\frac{1}{\mathrm{R}_{\mathrm{F}} \times 2 \times 10^{-10}}$
$\mathrm{fgen}^{\text {: Oscillation frequency }[\mathrm{Hz}]}$
RF : Element resistance [ $\Omega$ ]

## b. When an external clock is used




T : Min. 200ns
Twh: Min. 20ns
Twl : Min. 20ns
Duty Any (enabled if $1: 1$ is not ensured)
c. Reset

Initialization of M series
The initialization of $M$ series takes effect through the addition of pulses to Pin (2) with an open collector or setting at OV by contact signals, as show below. Proper initialization upon supply of the power is assured only if a rise in the supply voltage is at or less 10 ms .

## 2. Frequency demultiplier

Clock derived from the built-in oscillator or external clock is to be divided with the frequency demultiplier into $1 / 10,1 / 100$, and $1 / 1000$. The frequency demultiplier is under control of Pins (21) and (22) and, the setting is latched by Pin (23) signal.
Direct clock input to Pin (17) is required if no built-in frequency demultiplier is necessary. Pin (12) for frequency demultiplier input is connected to GND.

## Control signal timing chart

## 3. M series

Pin ${ }^{(17)}$ is designed for clock input.


T : 200ns (Max. 5MHz) Tw : Min. 20ns

## 4. White noise

Outputs delivered by the CF-742N generator is random binary outputs ( $\pm 5 \mathrm{~V}$-square waves with random cycles), which prompts the CF-742N generator to adopt the amplitude distribution (application of normal analog noise) as Gaussian distribution (normal distribution). Filtering is required to obtain flat frequency characteristics (white noise).
There is a close connection among the following: clock frequency (frequency of Pin (17), low-pass filter cut-off frequency, equivalent noise bandwidth determined by filter order, peak factor of analog noise (filter output) and output voltage.
The filter cut-off frequency and clock frequency are derived from the equations shown at right.
a. Set the output voltage Eo [rms]. (Peak factor: Min. 4)

$$
\mathrm{E}_{0} \leq 1.25 \mathrm{~V} \quad \mathrm{E}_{\mathrm{o}}: \mathrm{RMS} \text { value of output voltage [ } \mathrm{Vrms} \text { ] }
$$

b. Designate an equivalent noise bandwidth (B) and filter order to obtain filter cut-off frequency (fc).

fc: Filter cut-off frequency [Hz]
B: Equivalent noise bandwidth [Hz]
K: Noise bandwidth coefficient (see table 1)

| Table 1: Coefficient of noise |  |
| :---: | :---: |
| bandwidth (Butterworth) |  |
| Order | k |
| 1 | 1.57 |
| 2 | 1.11 |
| 3 | 1.05 |
| 4 | 1.03 |

c. Derive a clock frequency (fo) from the cut-off frequency (fc) and output voltage (Eo).

$$
\mathrm{fo}=\frac{50 \mathrm{~B}}{\mathrm{Eo}^{2}} \quad \text { fo. Clock frequency }[\mathrm{Hz}]
$$

## Characteristics



Voltage (V)

Random binary output power spectrum





Condition: Clock frequency: 10 kHz
Low-pass filter equivalent noise bandwidth: 200 Hz (2-pole Butterworth: $\mathrm{fc}=180 \mathrm{~Hz}$ )

## Technical data

## Noise output characteristics after filtering

Power spectrum of random binary output is derived from the following equation:
$\operatorname{PE}(\mathrm{f})=\frac{25}{\mathrm{f}_{0}}\left[\mathrm{~V}^{2} / \mathrm{Hz}\right]$ $\qquad$
fo: Clock frequency of random binary generator
Power spectrum varies along with filtering as shown below.
$\operatorname{PEO}(\mathrm{f})=\frac{25}{\mathrm{f}_{0}}|\mathrm{H}(\mathrm{j} \omega)|^{2}$ $\qquad$
$\mathrm{H}(\mathrm{j}(\mathrm{t})$ ): Filter transfer function
RMS value is determined from the following equation:
$E O=\sqrt{\frac{2.25}{\mathrm{f}_{0}} \int_{0}^{\infty}|\mathrm{H}(\mathrm{j} \omega)|^{2} \mathrm{~d} \omega}[\mathrm{rms}]$
If it is simplified,
$\mathrm{E}=\sqrt{\frac{50}{\mathrm{f}_{0}}} \mathrm{~A}^{2} \mathrm{~B}[\mathrm{rms}]$ $\qquad$
B: Equivalent noise bandwidth
A: Filter pass-band gain

The equivalent noise bandwidth ( $B$ ) is defined as follows:
$\mathrm{B}=\frac{1}{\mathrm{~A}^{2}} \int_{0}^{\infty}|\mathrm{H}(\mathrm{j} \omega)|^{2} \mathrm{~d} \omega$ $\qquad$
The filter order-equivalent noise bandwidth (B) relationship is provided in Table 1.
Flatness of not only noise bandwidth but noise frequency is of importance to use frequency as white noise
Random binary amplitude characteristics are expressed by the following equation:
$\mathrm{E}(\mathrm{f})=\frac{\sin (\pi \mathrm{f} / \mathrm{f} 0)}{\pi \mathrm{f} / \mathrm{f} 0}$
Amplitude characteristics are obtained as shown in Table 2.


Amplitude characteristics of Butterworth low-pass filters are assigned with " 1 " for a pass-band gain.
$\mathrm{E} 0(\mathrm{f})=\sqrt{\frac{1}{1+(\mathrm{f} / \mathrm{fc})^{2 \mathrm{n}}}}$
Amplitude characteristics are obtained as shown in Table 3.

Table 3: Butterworth filter amplitude characteristics

| f/fc | Amplitude [dB] |  |  |  |
| :---: | :---: | :---: | ---: | :---: |
|  | 1-pole | 2-pole | 3-pole | 4-pole |
| 0.001 | -0.00 | 0.00 | 0.00 | 0.00 |
| 0.01 | -0.00 | -0.00 | 0.00 | 0.00 |
| 0.1 | -0.04 | -0.00 | -0.00 | -0.00 |
| 0.2 | -0.17 | -0.01 | -0.00 | -0.00 |
| 0.3 | -0.37 | -0.04 | -0.00 | -0.00 |
| 0.4 | -0.64 | -0.10 | -0.02 | -0.00 |
| 0.5 | -0.97 | -0.26 | -0.07 | -0.02 |
| 0.6 | -1.34 | -0.53 | -0.20 | -0.07 |
| 0.7 | -1.73 | -0.93 | -0.48 | -0.24 |
| 0.8 | -2.14 | -1.49 | -1.01 | -0.67 |
| 0.9 | -2.57 | -2.19 | -1.85 | -1.55 |

The peak factor (P.F.) is defined as follows:
P.F. $=\frac{E_{p}}{E_{0}}$

$$
\begin{equation*}
\frac{p}{i 0} \ldots \tag{8}
\end{equation*}
$$

$\qquad$
Assign 5 [Vo-p] to peak value ( Ep ) for the generator, and substitute Eo from Equation (4).
P.F. $=\frac{5}{\sqrt{\frac{50}{\mathrm{f}_{0}} \mathrm{~A}^{2} \mathrm{~B}}}$

To use it as Gaussian random noise, determine Eo to maintain 4 in peak factor.

## Phase Detector



VAbsolute maximum ratings

| Supply voltage ( $\pm \mathrm{Vs}$ ) | $\pm 18 \mathrm{~V}$ |
| :--- | :--- |
| Signal input voltage | $\pm \mathrm{Vs}$ |
| Reference signal <br> input voltage | $+5.5 \mathrm{~V},-0.5 \mathrm{~V}$ |
| Logic control voltage | $+5.5 \mathrm{~V},-0.5 \mathrm{~V}$ |

Phase Detectors

|  | Operating frequency range | 1 kHz to 200 kHz | 10 kHz to 2 MHz |
| :---: | :---: | :---: | :---: |
| Phase Detectors | Gain ( $\phi=0$ ) | (sine-wave): Pins (12) and (13) open (sine-wave): Short in Pins (12) and (13) Selectable in the 1 to $10-\mathrm{Vdc} / \mathrm{Vo}-\mathrm{p}$ with the external resistor (Pins (12) and (13) |  |
|  | Gain accuracy | Max. $\pm 3 \%$ |  |
|  | Phase difference (signal system and reference signal system) | $\begin{aligned} & -0.05^{\circ} \text { (typ) at } 1 \mathrm{kHz}, \\ & -8^{\circ}(\text { typ }) \text { at } 200 \mathrm{kHz} \end{aligned}$ | $\begin{aligned} & -0.5^{\circ} \text { (typ) at } 10 \mathrm{kHz}, \\ & +13^{\circ} \text { (typ) at } 2 \mathrm{MHz} \end{aligned}$ |
|  | $\nabla$ Low-pass filter |  |  |
|  | Order | 1-pole (6dB/oct) |  |
|  | Cut-off frequency | Pins (9)-(10) shorted, Low-pass expansion is enabled with an external resistor or capacitor. | Pins (9)-(10) shorted, Low-pass expansion is enabled with an external resistor or capacitor. |
|  | $\nabla$ Detection output |  |  |
|  | Output impedance | Max. $50 \Omega \pm 10 \%$ at 1 kHz | Max. $50 \Omega \pm 10 \%$ at 10 kHz |
|  | Linear maximum input voltage | Min. $\pm 10 \mathrm{~V}$ ( DC, Load resistance $\geq 2 \mathrm{k} \Omega$ ) |  |
|  | Linear maximum input current | Min. $\pm 5 \mathrm{~mA}$ (DC) |  |
|  | Offset voltage | Max. $\pm 15 \mathrm{mV}, \pm 5 \mathrm{mV}$ (typ) <br> Short in input, Gain: 1Vdc/Vo-p |  |
|  | Offset voltage adjustment | Zero adjustment available with external pre-set resistors. (Pin (14) |  |

$\boldsymbol{\nabla}$ Signal system
$\nabla$ Signal input

| Model | CD-552R3 | CD-552R4 |
| :--- | :--- | :--- |
| Input impedance | Max. $10 \mathrm{k} \Omega \pm 5 \%$ <br> at 1 kHz | Max. 2.5k $\Omega+5 \%$ <br> at 10 kHz |
| Linear maximum <br> input voltage | Min. $\pm 10 \mathrm{~V}$ | Max. $130 \mathrm{~V} / \mu \mathrm{s}$ |
| Allowable slew rate | Max. $5 \mathrm{~V} / \mu \mathrm{s}$ |  |


| $\nabla$ Phase detector |  |
| :--- | :--- |
| Detection method | Synchronous rectifying type by square-wave <br> multiplication |

## Detection

 characteristics
## CD-552R3 CD-552R4

CD-552R series detectors are an on-board phase detectors possessing frequencies falling within the range of 1 kHz to 200 kHz for CD-552R3 and frequencies falling within the range of 10 kHz to 2 MHz for CD-552R4.
The signal system is composed of the phase sensitive detector (PSD), low-pass filter (LPF), and output amplifier. A low-pass expansion of output low-pass filter cut-off frequency is available with the addition of one external resistor, and the gain setting ( $\times 1$ to $\times 10$ ) is also enabled. The reference signal system consists of a $0^{\circ}-90^{\circ}$ phase shifter (PAT.P) and $50 \%$-duty circuit (PAT.P), which enables the detection of $A \sin \phi$ or $A \cos \phi$ phase. The phase detection with double frequency is permitted if $2 f$ mode is placed through the connection with the specified pin.
CD-552R series detectors are in a static-shielded 20-pin single inline package.

## $\nabla$ Reference signal system

$\nabla$ Reference signal input

| Model | CD-552R3 | CD-552R4 |
| :---: | :---: | :---: |
| Input circuit | CMOS Schmitt trigger, pulled up at $100 \mathrm{k} \Omega$ Trip point: $+3.5 \mathrm{~V} /+1.5 \mathrm{~V}$ (typ) |  |
| Input voltage | CMOS (0/+5V) level |  |
| Unipolar (1f) mode | A rising or falling edge is regarded as a reference. |  |
| Polarity switch | Pin (17) open or +5 V : Rising edge regarded as a reference 0 V : Falling edge regarded as a reference |  |
| Pulse duration | Min. 50nsec |  |
| Bipolar (2f) mode | Both rising and falling edge are regarded as a reference. |  |
| Mode setting | Connected with the reference signal input (Pin (18) and polarity switch input (Pin (17). |  |
| Input waveform | Duty: 50\% |  |
| Input frequency range | 1 kHz to 100 kHz | 10 kHz to 1 MHz |

## $\nabla 0^{\circ}-90^{\circ}$ phase shifter

| Function |  | This enables the detection of COS or SIN through a $0^{\circ}-90^{\circ}$ phase shift of reference signal input (Pin (18) |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $0^{\circ}-90^{\circ}$ phase d | ifference | $-90 \pm 0.5^{\circ},-90 \pm 0.1^{\circ}$ (typ) |  |  |  |
| Control |  | Pin (16) open or | $\begin{aligned} \hline+5 \mathrm{~V} & : 0^{\circ} \\ 0 \mathrm{~V} & :-90^{\circ} \end{aligned}$ |  | $\begin{aligned} & \text { (COS) } \\ & \text { (SIN) } \end{aligned}$ |
| Control inpu | circuit | CMOS Schmitt trigger, pulled up at $100 \mathrm{k} \Omega$ |  |  |  |
| $\nabla$ Others |  |  |  |  |  |
| Recommended supply voltage |  | $\pm 15 \mathrm{~V} \pm 1 \mathrm{~V}$ |  |  |  |
| Quiescent current |  | $\begin{aligned} & \pm 25 \mathrm{~mA}(\max ), \\ & \pm 20 \mathrm{~mA} \text { (typ) } \end{aligned}$ |  | $\begin{aligned} & \pm 35 \mathrm{~mA}(\max ), \\ & \pm 26 \mathrm{~mA} \text { (typ) } \end{aligned}$ |  |
| Specified temperature range |  | $23^{\circ} \mathrm{C} \pm 5^{\circ} \mathrm{C}$ |  |  |  |
| Temperature/ humidity range | Operation | $-20 \mathrm{C}^{\circ}$ to $70 \mathrm{C}^{\circ}, 10$ to $90 \% \mathrm{RH}$ |  |  |  |
|  | Storage | $-30 \mathrm{C}^{\circ}$ to $80 \mathrm{C}^{\circ}, 10$ to $80 \% \mathrm{RH}$ |  |  |  |
| Dimensions |  | $67 \times 10.5 \times 20 \mathrm{~mm}$ (protrusion not included) Type SS20 (20-pin shielded SIP) |  |  |  |
| Weight (NET) |  | Approx. 20 g |  |  |  |

Note: The following specifications are applied unless otherwise specified:
$23 \pm 5^{\circ} \mathrm{C}$, Supply voltage: $\pm 15 \mathrm{~V}$

## Block diagram



|  | CD-552R3 | CD-552R4 |
| :---: | :---: | :---: |
| Rin | 10 k | 2.5 k |
| CBint | 10000 p | 1000 p |

SIN/COS This is used to switch the internal phase shifter between $0^{\circ}$ and $90^{\circ}$, which enables the switching of detector input/output between $A \sin \phi$ and $A \cos \phi$.
[A: Amplitude (o-p) of input signal, $\phi$ : Phase difference between input signal and reference signal]
HI: A $\cdot \cos \phi$
$\left(0^{\circ}\right)$ (specified when the pin is open)
LO: A•sin $\phi$
(90 ${ }^{\circ}$ )

REF POL This is used to switch the reference polarity of reference signals. An edge specified is a reference phase. With the REF POL terminal connected to the REF IN terminal, the phase detection with double frequency is enabled if $50 \%$ of duty is assigned to the reference signal.

HI :Rising edge regarded as a reference (specified when the pin is open)
LO: Falling edge regarded as a reference
Connected with REF IN terminal :
Both rising and falling edge regarded as a reference
OFFSET This is used to adjust output DC offset. $\pm 15 \mathrm{~V}$ is available for input, which allows both terminals of the pre-set resistor to be connected with $\pm 15 \mathrm{~V}$ input. The sliding terminal is connected to the OFFSET terminal. The signal is transmitted to the REF IN terminal with the SIG IN terminal connected to the ground, which brings the pre-set resistor into action to make offset adjustment.

## Basic connection diagram



## Gain setting

CD-552R3/4 detectors are outfitted with the variable-gain output amplifiers ( $\times 1$ to $\times 10$ ). The maximum output voltage is set at $10 \mathrm{Vo}-\mathrm{p}$ that should not be surpassed when setting proper gain for post processor.

$$
\operatorname{Rg}=\frac{2.9873 \times 10^{4}}{\mathrm{~A}-1}-3.3 \times 10^{3}[\Omega]
$$

A: Gain [times $(\times)$ ]
Example: Set points

| Gain | $\times 1$ | $\times 2$ | $\times 5$ | $\times 10$ |
| :---: | :---: | :---: | :---: | :---: |
| Resistance | $\infty$ | $26.7 \mathrm{k} \Omega$ | $4.12 \mathrm{k} \Omega$ | 0 |

## CD-552R4



Example: Set points

| Cut-off frequency <br> (Equivalent noise bandwidth) | 10 Hz <br> $(15.7 \mathrm{~Hz})$ | 100 Hz <br> $(157 \mathrm{~Hz})$ | 1 kHz <br> $(1.57 \mathrm{~Hz})$ | 10 kHz <br> $(15.7 \mathrm{kHz})$ |
| :---: | :---: | :---: | :---: | :---: |
| Resistance | $140 \mathrm{k} \Omega$ | $1.58 \mathrm{M} \Omega$ | $143 \mathrm{k} \Omega$ | 0 |
| Capacitance | $0.1 \mu \mathrm{~F}$ | - | - | - |

$\mathrm{Rf}=\frac{1}{2 \pi \cdot\left(1 \times 10^{-9}+\mathrm{Cf}[\mathrm{F}]\right) \cdot \mathrm{fc}[\mathrm{Hz}]}-15.9 \times 10^{3}[\Omega]$
fc: Cut-off frequency
Cf: External capacitor

[^3]$R$ should remain at $2 \mathrm{M} \Omega$ or less with the use of the eternal capacitor $\left(\mathrm{C}_{\mathrm{f}}\right)$. Theory holds that a larger value can be assigned, but potential (Cf). Theory holds that a larger value can be assigned, but potential
deterioration in offset, DC drift and noise may be concerned if assigned.

LPF setting
CD-552R3/4 detectors are outfitted with the primary LPF that is capable of setting frequencies of $1 \mathrm{kHz}(10 \mathrm{kHz})$ or less with the use of the external CR. Proper frequency is to be allocated, allowing for the bandwidth, responsibility, and fluctuation for output signals.

CD-552R3

$$
\mathrm{Rf}=\frac{1}{2 \pi \cdot\left(1 \times 10^{-8}+\mathrm{Cf}[\mathrm{~F}]\right) \cdot \mathrm{fc}[\mathrm{~Hz}]}-15.9 \times 10^{3}[\Omega]
$$

fc : Cut-off frequency
Cf: External capacitor
Example: Set points

| Cut-off frequency <br> (Equivalent noise bandwidth) | 1 Hz <br> $(1.57 \mathrm{~Hz})$ | 10 Hz <br> $(15.7 \mathrm{~Hz})$ | 100 Hz <br> $(157 \mathrm{~Hz})$ | 1 kHz <br> $(1.57 \mathrm{kHz})$ |
| :---: | :---: | :---: | :---: | :---: |
| Resistance | $1.43 \mathrm{M} \Omega$ | $1.58 \mathrm{M} \Omega$ | $143 \mathrm{k} \Omega$ | 0 |
| Capacitance | $0.1 \mu \mathrm{~F}$ | - | - | - |



Characteristics CD-552R4

Gain fluctuations
Reference: 10 kHz , Gain: $\times 10$


Offset voltage fluctuations
Reference: 10 kHz , Gain: $\times 10$

$90^{\circ}$ phase shift fluctuations


Phase offset


## Characteristics CD-552R3

Gain accuracy Temperature


Offset voltage Temperature

$90^{\circ}$ phase shift accuracy Temperature


Phase offset Temperature


## Characteristics CD-552R4

Gain accuracy Temperature


Offset voltage Temperature

$90^{\circ}$ phase shift accuracy Temperature


Phase offset Temperature

## Pattern design

Proper connection between the case ground and the GND potential should always be assured. No sufficient shielding effect is produced if disregarded.
No signal traces should be assigned on the maximum visible outline of the component mounting surface. Possible contact between the metal case and the board is observed around the maximum visible outline, which triggers the establishment of a short circuit between the signal and case. A ground plane pattern is recommended to incorporate into the maximum visible outline and the inside of the case to enhance shielding effect.


Pattern dimensions


Maximum outer dimensions

## To assure dynamic range and stability

## Signal pre-processing

If a sufficient $\mathrm{S} / \mathrm{N}$ ratio fails to be obtained by the optimization of detector input level or setting of the output amplifier, a filter needs to be inserted in front of the detector to enhance the $\mathrm{S} / \mathrm{N}$ ratio of input signal.
The filter falls into the four types (low-pass, high-pass, band pass, and band elimination) and becomes a determinant of the following items: asynchronous signal frequency component, amplitude characteristics, filter characteristics, and cut-off frequency.
The band pass filter attenuates all signals other than synchronization signal, which maximizes the improvement of the $\mathrm{S} / \mathrm{N}$ ratio. Relatively large variations in phase around the center frequency, which may lead to detection accuracy if a phase change is made in response to temperature drift. Phase drift is minimized if low-order (1-pole if possible) $Q$ is assigned.
The low-/high-pass filters attenuate low-/high-pass signals, and offer the smaller improvement of the $\mathrm{S} / \mathrm{N}$ ratio as compared with the band pass filter. A phase change at a pass band is curbed, which contributes to a smaller detection accuracy attributed to fluctuations in cut-off frequency.
The band elimination provides large attenuation to signals of specified frequencies. An efficient improvement of the $\mathrm{S} / \mathrm{N}$ ratio is obtained if specified frequency is assigned to the asynchronous signal. The least phase change at a pass band is assured, which minimizes a detection accuracy attributed to fluctuations in cut-off frequency.

## Input signal level

CD-552R3/4 detectors features 10 V 0 -p of the maximum input level. A dynamic range can be assured if a large level of synchronization signal is input by maintaining within $10 \mathrm{~V} o-\mathrm{p}$. The actual input signal contains both asynchronous and synchronization signals, which requires a decrease in the amplitude of 10Vo-p or less.
E.g.: 0.1Vo-p synchronization signal is present in $1 \mathrm{~V}_{0}-\mathrm{p}$ signal that is a total of asynchronous and synchronization signals. CD$552 R 3 / 4$ detectors performs the detection of the signals at 1 Vdc of output despite the $\times 10$-post-stage DC amplifier being designated. The allowable input level enables a $\times 10$-amplifier to be inserted in front of the CD-552R3/4 detectors to input the maximum input voltage of 10 V 0 -p. The detection output obtains 10 Vdc when the $\times 10$-post-stage DC amplifier is designated, which allows the obtainment of the maximum output signal.

## ■Output amplifier

The output amplifier is capitalized on to obtain a proper output level if a small detection output remains despite the optimization of input signals. CD-552R3/4 detectors are outfitted with the vari-able-gain output amplifiers ( $\times 1$ to $\times 10$ ). The maximum output voltage is set at 10 V 0 -p that should not be surpassed when setting gain to assure proper voltage for post processor.
Note that an increase in DC drift, offset voltage and output noise is considered with an increase in gain.

## Phase adjustment

Phase detection with the use of the CD-552R3/4 detectors may require phase adjustment for the optimization of detection sensibility and cancellation of processing phase.
Phase adjustment is conducted in combination with the voltage controlled phase detector CD-951V4. Continuous change in phase shift of the reference signal is enabled through DC voltage.


## Evaluation board

A module-mounted evaluation board is available for easy evaluation of this module. Contact us for further information.


## Phase Shifter



VAbsolute maximum ratings

| Supply voltage ( $\pm$ Vs) | $\pm 18 \mathrm{~V}$ |
| :---: | :---: |
| Phase control | $\pm \mathrm{Vs}$ |
| DC input voltage |  |
| Phase shifter input voltage | $+5.5 \mathrm{~V},-0.5 \mathrm{~V}$ |
| Logic control voltage | $+5.5 \mathrm{~V},-0.5 \mathrm{~V}$ |
| 『50\%-duty output/voltage control phase shifter $\nabla$ Setting |  |
| Setting | Pins (15)-(16) shorted, Pin (17) open |
| I/O characteristics | $50 \%$-duty square wave, which a phase is shifted by voltage control, is output with reference to the edge specified at polarity switch of phase shifter input signal waveform. |
| VFrequency range |  |
| Frequency range | 1 kHz to 2 MHz <br> (2 ranges available: 1 kHz to $200 \mathrm{kHz}, 10 \mathrm{kHz}$ to 2 MHz ) |
| Range switch | Pin (12) open or +5 V : 1 kHz to 200 kHz 0 V : 10 k to 2 MHz |

$\boldsymbol{\nabla}$ Phase shifter input characteristics

| Input circuit | CMOS Schmitt trigger, pulled up at $100 \mathrm{k} \Omega$ |
| :---: | :---: |
| Trip point | +3.5V/+1.5V (typ) |
| Input voltage | CMOS (0/+5V) level |
| Unipolar (1f) mode | A rising or falling edge is regarded as a reference. |
| Polarity switch | Pin (13) open or +5 V : Rising edge regarded as a reference OV: Falling edge regarded as a reference |
| Pulse duration | Min. 50nsc |
| Bipolar(2f) mode | Both rising and falling edge are regarded as a reference. |
| Mode setting | Connected with the phase shifter input (Pin (14) and polarity switch input (Pin (13). |
| Input waveform | Duty : 50\% |
| Input frequency range | 1 kHz to 1 MHz |
| $\nabla$ Voltage control characteristics |  |


| Control method | Phase shift is specified in the proportion to <br> phase control DC input voltage. |
| :--- | :--- |
| Input resistance | $100 \mathrm{k} \Omega \pm 3 \% \quad$ (DC) |
| Linear maximum <br> input voltage | $\pm 5 \mathrm{~V} \leq 1 \mathrm{MHz}$ |
| Linear control range | $\pm 90^{\circ}$ |
| Voltage control sensitivity | $-20^{\circ} \mathrm{V}\left(-100^{\circ} /+5 \mathrm{~V}, 100^{\circ} /-5 \mathrm{~V}\right)$ |
| Sensitivity accuracy | $\pm 1^{\circ} \mathrm{V}$ |


| Output circuit |  | HCMOS output, series resistor at $100 \Omega$ |
| :---: | :---: | :---: |
| Output voltage |  | CMOS (0/+5V) level |
| Duty |  | $\begin{aligned} & 50 \% \pm 0.03 \% \text { (typ) } \quad \text { (at } 200 \mathrm{kHz} \text { ) } \\ & 50 \% \pm 0.3 \% \text { (typ) } \quad \text { (at } 2 \mathrm{MHz} \text { ) } \end{aligned}$ |
| 0/180 ${ }^{\circ}$ switch |  | Pin (20) open or $+5 \mathrm{~V}:-180^{\circ}, 0 \mathrm{~V}: 0^{\circ}$ |
| -180 ${ }^{\circ}$ accuracy |  | $\begin{array}{ll} -180^{\circ} \pm 0.02^{\circ} \text { (typ) } & \text { (at } 200 \mathrm{kHz}) \\ -180^{\circ} \pm 0.2^{\circ} \text { (typ) } & \text { (at } 2 \mathrm{MHz} \text { ) } \\ \hline \end{array}$ |
| Phase offset |  | (1k to 200 kHz$)$ $-0.6^{\circ}$ (typ) (at 1 kHz ) <br>  $-4.5^{\circ}$ (typ) (at 200 kHz ) <br> $(10 \mathrm{kHz}$ to 2 MHz$)$ $-0.9^{\circ}$ (typ) (at 10 kHz$)$ <br>  $-42.0^{\circ}$ (typ) (at 2 MHz ) |
| Phase offset adjustment |  | Adjustment available with a 20k $\Omega$-external potentiometer. (Pin (2)) |
| Adjustment range |  | $\pm 5^{\circ}$ (typ) |
| VReference voltage |  |  |
| Output voltage/accuracy |  | Max. $\pm 5 \mathrm{~V} \pm 2 \%$ |
| Temperature stability |  | 50ppm/ ${ }^{\circ} \mathrm{C}$ (typ) |
| Maximum output current |  | $\pm 1 \mathrm{~mA}$ |
| $\nabla$ Others |  |  |
| Recommended supply voltage |  | $\pm 15 \mathrm{~V} \pm 1 \mathrm{~V}$ |
| Quiescent current |  | $\begin{aligned} & +25 m A(\max ),+18 m A(\text { typ }) \\ & -20 m A(m a x),-12 m A(t y p) \end{aligned}$ |
| Temperature/ humidity range | Operation | $-20 \mathrm{C}^{\circ}$ to $70 \mathrm{C}^{\circ}, 10$ to $90 \% \mathrm{RH}$ |
|  | Storage | $-30 \mathrm{C}^{\circ}$ to $80 \mathrm{C}^{\circ}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions |  | $67 \times 10.5 \times 20 \mathrm{~mm}$ (protrusion not included) Type SS20 (20-pin shielded SIP) |
| Weight (NET) |  | Approx. 20g |
| Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}$, Supply voltage: $\pm 15 \mathrm{~V}$ |  |  |



Block diagram


SHIFTER IN This is used to switch the reference polarity of shifter POL input. The operation at double frequency, as compared with the reference signal, is actualized through the connection between the SHIFTER IN POL terminal and SHIFTER IN terminal if $50 \%$ of duty is assigned to the reference signal.

HI: Rising edge regarded as a reference (specified when the pin is open)
LO: Falling edge regarded as a reference
Connected with SHIFTER IN terminal: Both rising and falling edge regarded as a reference

PHASE OFFSET
$200 \mathrm{k} / \overline{2 \mathrm{M}}$

NOR/INV

DUTY50 IN POL

This is used to cancel phase offset. Zero adjustment of the phase offset for CD-951V4 phase shifter only is enabled in the range of 1 kHz to 200 kHz . Both terminals of a trimmer potentiometer of $20 \mathrm{k} \Omega$ min . are connected with $\pm 5 \mathrm{~V}$ input (Pins (6) and (7), and the center terminal is connected to the PHASE OFFSET terminal.

This is used to switch the operating frequency range between $1 \mathrm{kHz}-200 \mathrm{kHz}$ and $10 \mathrm{kHz}-2 \mathrm{MHz}$ in response to the used frequency.
$\mathrm{HI}: 1 \mathrm{kHz}$ to 200 kHz (The pin is open)
LO: 10 kHz to 200 MHz
This is used to switch the output phase between $0^{\circ}$ and $180^{\circ}$. A $360^{\circ}$-phase shifter is configured in combination with a continuously variable phase shifter $\left( \pm 90^{\circ}\right)$.
$\mathrm{HI}: 0^{\circ}$ (The pin is open)
LO: $180^{\circ}$
This is used to switch the input polarity of the 50\%duty circuit. "HI" (open) should remain on for normal connection.
HI: Rising edge regarded as a reference
(The pin is open)
LO: Falling edge regarded as a reference

## Timing chart



This timing chart presents the operation of the voltage controlled phase shifter CD-951V4.
E.g.: The CD-951V4 phase shifter is set to regard a rising edge of the input signal as a phase reference. This detector produces the signal "LO" (Pin (15)) for the time proportionate to the control voltage (td) if a rise is observed in the input signal (Pin (14)).
Waveform shaping (Pin (18) is performed to assure $50 \%$ in duty (t1 $=\mathrm{t} 2)$ with reference the rising edge in the obtained signal.
td adjustment allows continuous change in input/output rise time (tsft), which denotes phase change.
The same operating principles* are applied to the phase detector CD-552R3 that has realized $90^{\circ}$-phase shift with high accuracy.

* Patent pending


## Usage example 2-phase detector



This example indicates the adoption of this detector to the 2-phase detector. The cos and sin detection outputs are obtained, which allows amplitude and phase of the synchronization signals to be derived from the relevant vector operation.
The settings of GAIN ( $\times 1$ to $\times 10$ ) and LPFfc (max. 1 kHz ) are available in this detector. Offset adjustment is required as necessary. Phase adjustment is available by $90^{\circ}$-continuous phase shift (CD951 V4 R1) or $0 / 180^{\circ}$-switch (S33), which enables $360^{\circ}$-phase change in total.

GAIN setting: Short: $\times 10$
Open: $\times 1$
LPFfc setting (same as R21):
Short: 1kHz

Note: See the CD-552R3/R4 in Page 72 for details in the GAIN setting and LPF setting.

## Characteristics



Control voltage coefficient - Temperature


## Phase Detector



## CD-505R2

CD-505R2 detector is a hybrid phase detector composed of the following units: input differential amplifier, two post amplifiers, band-pass filter, phase shifter, phase detector, and low-pass filter. This detector possessing the frequency range of 10 Hz to 10 kHz enables the setting of center frequency with the use of the resistors (2 pcs.). Not only gain setting for the post amplifiers with the resistors (2 pcs.) but phase setting with the resistor and trimmer potentiometer is also available.
The reference signal is designed to apply square wave with $1: 1$ of a duty factor, and the phase shifter assures its phase adjustment in the range of $\pm 45^{\circ}$. The post amplifier can be utilized as a $90^{\circ}$-phase shifter and inverting amplifier that actualizes $\pm 360^{\circ}$-adjustment with the combined use of the switch.
The 2-pole low-pass filter $(Q=0.5)$ is allocated, which facilitates the setting of the equivalent noise bandwidth with the use of resistors and capacitors.

| Supply voltage ( $\pm$ Vs) | $\pm 18 \mathrm{~V}$ |
| :---: | :---: |
| Signal input voltage | $\pm \mathrm{Vs} \mathrm{(1)}, \mathrm{(3)}, \mathrm{(5)}, \mathrm{(11)}, \mathrm{(33)}, \mathrm{(39)}$ |
| Reference signal input voltage | +5.5V (11) |
| FInput amplifier |  |
| Input form | Differential input |
| Input impedance | Differential input $200 \mathrm{k} \Omega$ <br> Inverting input $100 \mathrm{k} \Omega$ Non-inverting input $200 \mathrm{k} \Omega$ |
| Gain | $\times 1$ |
| Frequency characteristics | DC to 10 kHz |
| Maximum input voltage (linear) | $\pm 10 \mathrm{~V}$ |
| VPost amplifier |  |
| Gain | $\times 1$ to $\times 100$ ( 2 -stage amplifier, $\times 10 \times 2$ ) <br> Setting : Specified with external resistors (2 pcs.). |
| I/O phase | In-phase |
| Frequency characteristics | DC to 10kHz |
| V Band pass filter |  |
| Characteristics | 1-pole pair band pass filter |
| Q | 5 |
| Center frequency (fo) Setting method | Range: 10 Hz to 10 kHz <br> Setting: <br> Specified with external resistors (2 pcs.). $\mathrm{RBP} \leq 1.59 \mathrm{M} \Omega$ Combined use of external capacitor is also available if 100 Hz or less is obtained. |
| Gain | $0 \mathrm{~dB} \pm 0.5 \mathrm{~dB}$ |


| Frequency range |  | 10 Hz to 10 kHz |
| :---: | :---: | :---: |
| Phase shift |  | Range: $90^{\circ} \pm 45^{\circ}$ <br> Setting: Specified with an external resistor and a trimmer potentiometer <br> Combined use of an external capacitor is also available if 100 Hz or less is obtained. |
| Gain |  | $\times 1$ |
| -Phase detector |  |  |
| Frequency range |  | 10 Hz to 10 kHz |
| Type |  | Synchronous detection (with reference signal) |
| Reference signal <br> Input processing (internal) |  | TTL level, Duty factor: 1:1 Pulled down at $100 \mathrm{k} \Omega$ |
| Offset |  | $\phi 1 / \phi 2$ balanced, Output offset adjustment available with an external trimmer potentiometer. |
| $\nabla$ Low-pass filter |  |  |
| Characteristics |  | 2-pole low-pass filter |
| Equivalent noise bandwidth |  | Range: 30 Hz to 1 kHz (with 2 external resistors) Any setting is available with 2 external resistors (RLP) and 2 capacitors (Clp). |
| $\nabla$ Others |  |  |
| Supply voltage |  | $\pm 15 \mathrm{~V}( \pm 14$ to $\pm 16 \mathrm{~V})$ |
| Quiescent current |  | $\pm 30 \mathrm{~mA}$ (typ) |
| Temperature/ humidity range | Operation | $-20 \mathrm{C}^{\circ}$ to $70 \mathrm{C}^{\circ}, 10$ to $95 \% \mathrm{RH}$ |
|  | Storage | $-30 \mathrm{C}^{\circ}$ to $80 \mathrm{C}^{\circ}, 10$ to $80 \% \mathrm{RH}$ |
| Dimensions |  | $54.4 \times 33.7 \times 6.5 \mathrm{~mm}$ Type H |

Note: The following specifications are applied unless otherwise specified: $23 \pm 5^{\circ} \mathrm{C}, \pm 15 \mathrm{~V}$

## Basic connection diagram

## -Usage example of post amplifier



1. Signal system amplifier for the detection of micro-input signals
2. Instrumentation amplifier to obtain a large CMRR at high input impedance
3. Phase shifter to assure the $360^{\circ}$-range for phase adjustment

## -Calculation of constant

1. To determine the center frequency
$\Rightarrow$ Band pass filter: Rep1, 2 (Cbp1,2)
2. To determine the phase shift
$\Rightarrow$ Phase shifter: $\mathrm{C} \phi, \mathrm{R} \phi, \mathrm{RV} \phi$
3. To determine the equivalent noise bandwidth
$\Rightarrow$ Low-pass filter: RLp1, 2 (CLp1,2)

## Block diagram



## (1)-(4): Input amplifier

The input amplifier is a differential amplifier carrying Pin (1) for non-inverting input and Pin (2) for inverting input.
The basic usage of the input amplifier is shown below.

## 1. Differential amplifier


2. Inverting amplifier

3. Non-inverting amplifier


## (11)-(14): Phase shifter

The phase shifter is used to adjust signal system phase in the $90^{\circ}$-range.
Phase adjustment exceeding the above range requires the $360^{\circ}$-phase shifter application.
Signal monitor terminal: Pins (29) and (30)

## (5)-(10: Band pass filter

This band pass filter enables the measurement of fundamental waves with harmonics eliminated.
By using external components, it is possible to configure a 1-pole pair band pass filter ( $\mathrm{Q}=5$ ).
This band pass filter is capable of providing an attenuation of 20 dB to 3 -order harmonics and of 26 dB to 5 -order harmonics. With RBP used, center frequency adjustment for the band pass filter is performed to keep a phase difference " 0 " or " $180^{\circ}$ " through a comparison between the input signal and BpF OUT terminal 8 signal.

## PSD

This is used for the phase detection in 2-phase signals by the reference signal.

## (16)-(20: LPF

This is a low-pass filter capable of determining the equivalent noise bandwidth. The configuration of a 2-pole LPF is enabled with the use of the external resistors ( 2 pcs.) or the combined use of the external capacitor according to frequency. Some use applications may require the use of a 1-pole pair low-pass filter. See Page 83 for details.

## PHASE DETECTOR

## ©Offset adjustment

Offset adjustment is required for 2 places.
Use the following procedures for offset adjustment.

## 1. BALANCE (RV1)

Establish a ground for +/- inputs. The PSD OUT terminal (16) is to be monitored at the maximum sensitivity of the oscilloscope. Input the reference signal at the used frequency, and adjust the BALANCE RV1 to minimize p-p of the square wave.

(Before adjustment)

(After adjustment)

Oscilloscope waveform

## 2. OFFSET (RV2)

Use the same steps to connect the DC OUT (20) to the DC voltmeter. Adjust the OFFSET volume to obtain "0" in output DC voltage.

Note: Offset voltage contains frequency characteristics, which requires re-adjustment if a change is made in the signal frequency.

Band pass filter setting

| Table 1: RBP constants |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :--- | :---: |
|  | No CBP used |  | CBP used |  |  |
| fo | RBP | CBP | RBP | CBP |  |
| 10 kHz | $1.58 \mathrm{k} \Omega$ | - |  | - |  |
| 1 kHz | $15.8 \mathrm{k} \Omega$ | - |  | - |  |
| 100 Hz | $158 \mathrm{k} \Omega$ | - | 14.3 k | $0.1 \mu$ |  |
| 10 Hz | $1.58 \mathrm{M} \Omega$ | - | 143 k | $0.1 \mu$ |  |

## fo fine adjustment

Fine adjustment of center frequency requires a trimmer potentiometer to be assigned to either of RBPS as series.


If a frequency is out of the constants listed in Table 1, RBP and CBP need to be derived from the following equations.

$$
\begin{aligned}
& \text { fo } \geq 100 \mathrm{~Hz} \\
& \text { RBP }=\frac{15915}{\text { fo }}[\mathrm{k} \Omega] \quad \text { fo: }[\mathrm{Hz}]
\end{aligned}
$$

fo $<100 \mathrm{~Hz}$

$$
\begin{gathered}
\mathrm{P}_{\mathrm{BP}}=\frac{1.5915 \times 10^{5}}{\left(0.01+\mathrm{C}_{\mathrm{BP}}\right) \cdot \mathrm{fo}}[\Omega] \quad \text { fo: }[\mathrm{Hz}], \mathrm{C}_{\mathrm{BP}}:[\mu \mathrm{F}] \\
1.59 \mathrm{k} \Omega \leq \mathrm{R}_{\mathrm{BP}} \leq 1.59 \mathrm{M} \Omega
\end{gathered}
$$

## ■ Phase shifter setting

## 1) When any frequency is allocated: 1



Connection diagram 1
$1 k \Omega \leq R \phi \leq 100 k \Omega$
Determine $\mathrm{R} \phi$ and $\mathrm{C} \phi$ from the following equation to agree with the above values.

$$
\mathrm{R} \phi=\frac{1}{2 \pi \cdot\left(\mathrm{C} \phi+3.9 \times 10^{-9}\right) \cdot 2.72 \mathrm{f}}[\Omega]
$$

$$
\begin{aligned}
& \mathrm{f}:[\mathrm{Hz}] \\
& \mathrm{C} \phi:[\mathrm{F}]
\end{aligned}
$$

Derive RV $\phi$ in accordance with the conditions of the determined $R \phi$ and $R V \phi \geq 6.67 R \phi$.
E.g.: 400 Hz
$\mathrm{C} \phi=1700 \mathrm{pF}$
$216.1 \mathrm{k} \Omega$ is derived for $R \phi$ from the above equation.
$R V \phi>174 \mathrm{k} \Omega$ leads to $R V \phi=200 \mathrm{k} \Omega$.
2) When any frequency is allocated: 2

$R V \phi=100 k \Omega$

$$
\begin{aligned}
& \mathrm{C} \phi=\frac{1}{2 \pi \cdot \mathrm{f} \cdot 40.8 \times 10^{3}}[\mathrm{~F}] \\
& \mathrm{f}:[\mathrm{Hz}]
\end{aligned}
$$

Determine $\mathrm{C} \phi$ from the following equation.

## Equivalent noise bandwidth setting

1) When 2-pole low-pass filter is used

| Equivalent noise bandwidth | Time constant (TC) Connection diagram | RLP1, 2 | CLP1, 2 |  |
| :---: | :---: | :---: | :--- | :---: |
| 100 Hz | 1.25 msec | 1 | $124 \mathrm{k} \Omega$ | - |
| 30 Hz | 4.17 msec | 1 | $412 \mathrm{k} \Omega$ | - |
| 10 Hz | 12.5 msec | 1 | $1.24 \mathrm{M} \Omega$ | - |
| 3 Hz | 41.7 msec | 2 | $41.2 \mathrm{k} \Omega$ | $1 \mu \mathrm{~F}$ |
| 1 Hz | 125 msec | 2 | $124 \mathrm{k} \Omega$ | $1 \mu \mathrm{~F}$ |
| 0.3 Hz | 417 msec | 2 | $412 \mathrm{k} \Omega$ | $1 \mu \mathrm{~F}$ |
| 0.1 Hz | 1.25 sec | 2 | $1.24 \mathrm{M} \Omega$ | $1 \mu \mathrm{~F}$ |
| 0.03 Hz | 4.17 sec | 2 | $412 \mathrm{k} \Omega$ | $10 \mu \mathrm{~F}$ |
| 0.01 Hz | 12.5 sec | 2 | $1.24 \mathrm{M} \Omega$ | $10 \mu \mathrm{~F}$ |

Time constant (TC)=RLP•CLP
Equivalent noise bandwidth=1/8TC
\{ Any RLp and Cடp available according to $10 \mathrm{k} \Omega \leq \operatorname{RLp} \leq 1.59 \mathrm{M} \Omega$.

(19)
<Figure 1> Equivalent noise bandwidth: 10 Hz to 100 Hz

<Figure $2>$ Equivalent noise bandwidth $<10 \mathrm{~Hz}$

A settling time for output voltage is 6 - to 7 -times time constant.
2) When 1-pole low-pass filter is used

| Equivalent noise bandwidth | Time constant (TC) Connection diagram | RLp1, 2 | CLP1, 2 |  |
| :---: | :---: | :---: | :---: | :---: |
| 100 Hz | 2.5 msec | 1 | $249 \mathrm{k} \Omega$ | - |
| 30 Hz | 8.33 msec | 1 | $825 \mathrm{k} \Omega$ | - |
| 10 Hz | 25 msec | 2 | $226 \mathrm{k} \Omega$ | $0.1 \mu \mathrm{~F}$ |
| 3 Hz | 83.3 msec | 2 | $750 \mathrm{k} \Omega$ | $0.1 \mu \mathrm{~F}$ |
| 1 Hz | 250 msec | 2 | $249 \mathrm{k} \Omega$ | $1 \mu \mathrm{~F}$ |
| 0.3 Hz | 833 msec | 2 | $825 \mathrm{k} \Omega$ | $1 \mu \mathrm{~F}$ |
| 0.1 Hz | 2.5 sec | 2 | $249 \mathrm{k} \Omega$ | $10 \mu \mathrm{~F}$ |
| 0.03 Hz | 8.33 sec | 2 | $825 \mathrm{k} \Omega$ | $10 \mu \mathrm{~F}$ |
| 0.01 Hz | 25.0 sec | 2 | $1.13 \mathrm{M} \Omega$ | $22 \mu \mathrm{~F}$ |

Time constant (TC) $=$ RLP $\bullet C_{L P} \quad\{$ Any RLP and CLP available Equivalent noise bandwidth=1/4TC $\quad$ according to $10 \mathrm{k} \leq \mathrm{R}_{\mathrm{L} P} \leq 1.59 \mathrm{M} \Omega$.

$\leq$ Figure 2> Equivalent noise bandwidth $<30 \mathrm{~Hz}$

A settling time for output voltage is 4 - to 5 -times time constant.

## Application of post amplifier


-360응 phase shifter

- Instrumentation amplifier

Gain between input and Pin (4)
 Gain $=\frac{R_{G} 3+20 \times 10^{3}}{R_{\mathrm{G}} 3} \mathrm{R}_{\mathrm{G}} 3:[\Omega]$


Determine C 2 2 [F].

$$
\begin{aligned}
& \mathrm{R} \phi 2=\frac{1}{2 \pi \cdot \mathrm{f} \cdot \mathrm{C} \phi 2}[\Omega] \\
& \mathrm{f}:[\mathrm{Hz}] \\
& * 1.59 \mathrm{k} \leq \mathrm{R} \phi 2 \leq 1.59 \mathrm{M}
\end{aligned}
$$

## Characteristics



Output step response (1-pole low-pass filter)


I/O waveform (Phase difference: $0^{\circ}$ )


I/O waveform (Phase difference: $180^{\circ}$ )
Input signal

Ref signal $\rceil \square \square \square \square$
PSD output

| +FS |  |
| ---: | ---: |
| DC output |  |
|  | $\square$ |
| -FS | $\square$ |

Output step response (2-pole low-pass filter)


I/O waveform (Phase difference: $90^{\circ}$ )



PSD output $\sqrt[A]{ }$


I/O waveform (Phase difference: $270^{\circ}$ )
Input signal



## [Type S15]


[Type S12]

[Type SS20]


## Dimensional Outline Drawing

-Dual-inline package (DIP)
[Type EB]


Cross section of pin $0.5 \times 0.25 \mathrm{~mm}$
[Type H] [Type HA] [Type HB]


Cross section of pin $0.5 \times 0.25^{\mathrm{mm}}$

## [Type KB]



Cross section of pin $\quad 0.5 \times 0.25 \mathrm{~mm}$

## [Type KC]



Material: Body: 90\% alumina ceramics (black) Cover:Kovar ( $\mathrm{Fe}, \mathrm{Ni}, \mathrm{Co}$ )

Surface treatment: Gold-plated Pin: Alloy 42 ( $\mathrm{Fe}, \mathrm{Ni}$ )
Surface treatment: Nickel-plated + Gold-plated

## [Type ID]




Cross section of pin $\quad 0.5 \times 0.25 \mathrm{~mm}$

## [Type B]



|  | $\pm 15 \mathrm{~V}$ | -24 V |
| :---: | :---: | :---: |
| 1 | +B | +B |
| 2 | OUTPUT | OUTPUT |
| 3 | -B | GND |
| 4 | CASE GND | CASE GND |
| 5 | INPUT | INPUT |
| 6 | GND | NC |
| 7 | NC | NC |
| 8 | NC | NC |

* US socket-compliant plug
[Type L]



## [Type NL]



UOUU


|  | $\pm 15 \mathrm{~V}$ | +24 V |
| :---: | :---: | :---: |
| 1 | INPUT | INPUT |
| 2 | GND | GND |
| 3 | + B | + B |
| 4 | OUTPUT | OUTPUT |
| 5 | $-B$ | GND |

## FUNCTION MODULE DATA BOOK

- The description given in this data book is based on the information as of April 1, 2005 - Some appearance and specifications may change without notice.
- Please check the latest specifications before purchasing

NF Corporation

## NF Corporation

- Head Office

OREPRESENTATIVE
6-3-20 Tsunashima Higashi, Kohoku-ku, Yokohama 223-8508, Japan
Phone : +81-45-545-8128 Fax : +81-45-545-8187

## OShanghai Representative Office

Room5E, Modern Mansion, 218 Xiangyang South Road, Xuhui District
Shanghai 200031, China
Phone : +86-21-6473-5735 Fax : +86-21-6415-6576
OShenzhen Representative Office
Room1701, East, Aidi Building, No. 5003 Binhe Road, Futian District,
Shenzhen 518045, China
Phone : +86-755-8355-1866 Fax : +86-755-8355-1214


[^0]:    *1 Types are determined by the frequency range. E.g.: SR-4FL2 (Type 2) $\rightarrow 400 \mathrm{~Hz}$ to 20 kHz
    Some models allow frequency expansion with the adoption of external components.
    *2 These filters can be customized to your specifications including the cut-off (center) frequency and filter characteristics that you select from our existing filter characteristics.

[^1]:    Note: The following specifications are applied unless otherwise specified:

[^2]:    A multichannel filter with DC power supply is also available by special order. Contact

[^3]:    Capaciance

